



National Semiconductor

December 1996

LMC272

CMOS Dual Low Cost Rail to Rail Output Operational Amplifier

General Description

The LMC272 is a CMOS dual operational amplifier with rail-to-rail output swing and an input common voltage range that extends below the negative supply. Other performance characteristics include low voltage operation, low bias current, excellent channel-to-channel isolation, good bandwidth performance and a competitive price.

These devices are available in MSOP package which is about half the size of a SO-8 device. This enables the designer to fit the device in extremely small applications.

The LMC272C is a direct replacement for TLC272C with performance which meets or exceeds the TLC272C's guaranteed limits in the commercial temperature range when operating from a supply of 2.7V to 15V (see Electrical Characteristics table for details).

These features make this cost effective device ideal for new designs as well as for upgrading existing designs. Applications include hand-held analytic instruments, transducer amplifiers, sample and hold circuits, etc.

Features

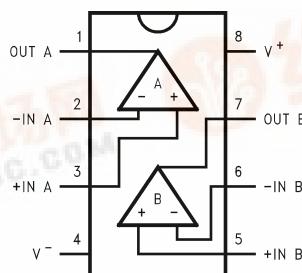
(Typical unless otherwise noted) $V_S = 5V$, $T_A = 25^\circ C$

- Output Swing to within 60 mV of supply rail (10 k Ω load)
- High voltage gain: 90 dB
- Unity gain-bandwidth: 2.0 MHz
- Wide supply voltage: 2.7V to 15V
- Characterized for: 2.7V, 5V, 10V
- Low supply current: 0.975 mA/amplifier
- Input voltage range: -0.3V to 4.2V

Applications

- Portable instruments
- Upgrade for TLC272C and TS272C
- Photodetector preamplifiers
- D/A converters
- Filters

Connection Diagram



Top View

Ordering Information

Package	Ordering Information	NSC Drawing Number	Package Marking	Supplied as
8-pin Molded DIP	LMC272CN	N08E	LMC272CN	Rails
8-pin SO-8	LMC272CM	M08A	LMC272CM	Rails
	LMC272CMX	M08A	LMC272CM	2.5k Tape and Reel
MSOP	LMC272CMM	MUA08A	A07	Rails
	LMC272CMMX	MUA08A	A07	3k Tape and Reel

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

ESD Tolerance (Note 2)	2 kV
Differential Input Voltage	\pm Supply Voltages
Voltage at Input/Output Pin	(V ⁺)+0.3V, (V ⁻)-0.3V
Supply Voltage (V ⁺ – V ⁻):	16V
Current at Input Pin (Note 10)	\pm 5 mA
Current at Output Pin (Note 3) (Note 7)	\pm 30 mA
Lead Temperature (soldering, 10 sec.)	260°C

Storage Temp. Range –65°C to +150°C
Junction Temperature (Note 4) 150°C

Operating Ratings (Note 1)

Supply Voltage	2.5V \leq V _S \leq 15V
Junction Temperature Range	0°C \leq T _J \leq +70°C
LMC272C	115° C/W
Thermal Resistance (θ_{JA})	177° C/W
N Package, 8-pin Molded DIP	235° C/W
M Package, 8-pin Surface Mount	
MSOP Package	

2.7V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for T_J = 25°C, V⁺ = 2.7V, V⁻ = 0V, V_{CM} = V_O = V⁺/2, R_L to ground, and R_L > 1 MΩ. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
V _{OS}	Input Offset Voltage	V _O = 1.4V, R _S = 50, V _{CM} = 0V, R _L = 10k	1.40	7 9	mV max
TCV _{OS}	Temp. Coefficient of Input Offset Voltage	T _A = 0°C to 70°C	3.9		µV/°C
I _B	Input Bias Current		1	64	pA max
I _{OS}	Input Offset Current		0.5	32	pA max
CMRR	Common Mode Rejection Ratio	V _{CM} = –0.2V to 1.2V	77	65 60	dB min
PSRR	Power Supply Rejection Ratio	V ⁺ = 2.7V to 5V, V _O = 1.4V	75	65 60	dB min
V _{CM}	Input Common-Mode Voltage Range	CMRR \geq 50 dB	1.7	1.5 1.2	V min
			–0.3	–0.2 –0.2	V max
A _V	Large Signal Voltage Gain	V _O = 0.25V to 2.45V, R _L = 10k	88		dB
V _O	Output Swing	R _L = 10 kΩ, V _{ID} = 100 mV (Note 11)	2.64	2.55	V min
		V _{ID} = –100 mV (Note 11)	0	20 25	mV max
I _{SC}	Output Short Circuit Current	Sourcing, V _{ID} = 100 mV (Note 11)	3.7		mA
		Sinking, V _{ID} = –100 mV (Note 11)	2.5		mA
I _S	Total Supply Current		1.60	2.5 3.0	mA max

2.7V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^\circ\text{C}$, $V^+ = 2.7\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, R_L to ground and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
SR	Slew Rate (Note 8)	$A_V = +1$, $R_L = 10 \text{k}\Omega$, $V_I = 1 \text{V}_{PP}$, $C_L = 20 \text{ pF}$ (Note 12)	1.7		$\text{V}/\mu\text{s}$
GBW	Unity Gain Frequency	$V_I = 10 \text{ mV}_{PP}$, $C_L = 20 \text{ pF}$ (Note 12)	1.9		MHz
ϕ_m	Phase Margin	$V_I = 10 \text{ mV}_{PP}$, $C_L = 20 \text{ pF}$ (Note 12)	39		Deg
e_n	Input-Referred Voltage Noise	$f = 1 \text{ kHz}$, $R_S = 20\Omega$	27		$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
i_n	Input-Referred Current Noise	$f = 1 \text{ kHz}$	0.0015		$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
f_{max}	Full Power Bandwidth	$V_S = 10\text{V}$, $C_L = 20 \text{ pF}$, $R_L = 20 \text{k}\Omega$	120		kHz
	Amp-to-Amp Isolation	(Note 9)	150		dB
THD	Total Harmonic Distortion	$A_V = +1$, $V_{IN} = 0.7V_{PP}$ $f = 1 \text{ kHz}$	0.035		%

5V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^\circ\text{C}$, $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, R_L to ground and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
V_{OS}	Input Offset Voltage	$V_O = 1.4\text{V}$, $R_S = 50$, $R_L = 10\text{k}$, $V_{CM} = 0\text{V}$	1.75	7 9	mV max
TCV_{OS}	Temp. Coefficient of Input Offset Voltage	$T_A = 0^\circ\text{C}$ to 70°C	3.3		$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current		1	64	pA max
I_{OS}	Input Offset Current		0.5	32	pA max
CMRR	Common Mode Rejection Ratio	$V_{CM} = -0.2\text{V}$ to 3.5V	77	65 60	dB min
PSRR	Power Supply Rejection Ratio	$V^+ = 5\text{V}$ to 10V , $V_O = 1.4\text{V}$	88	65 60	dB min
V_{CM}	Input Common-Mode Voltage Range	CMRR ≥ 50 dB	4.2	4 3.5	V min
			-0.3	-0.2 -0.2	V max
A_V	Large Signal Voltage Gain	$V_O = 0.25\text{V}$ to 2V , $R_L = 10\text{k}$	90	80 72	dB min
V_O	Output Swing	$R_L = 10\text{k}\Omega$, $V_{ID} = 100\text{ mV}$ (Note 11)	4.94	4.85 4.75	V min
		$V_{ID} = -100\text{ mV}$ (Note 11)	0	20 25	mV max
I_{SC}	Output Short Circuit Current	Sourcing, $V_{ID} = 100\text{ mV}$ (Note 11)	16		mA
		Sinking, $V_{ID} = -100\text{ mV}$ (Note 11)	16		mA
I_S			1.95	3.2 3.6	mA max

5V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^\circ\text{C}$, $V^+ = 5\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, R_L to ground and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
SR	Slew Rate (Note 8)	$A_V = +1$, $R_L = 10\text{ k}\Omega$, $VI = 1\text{ V}_{PP}$, $C_L = 20\text{ pF}$ (Note 12)	2.5		$\text{V}/\mu\text{s}$
		$A_V = +1$, $R_L = 10\text{ k}\Omega$, $VI = 2.5\text{ V}_{PP}$, $C_L = 20\text{ pF}$ (Note 12)	2.5		
GBW	Unity Gain Frequency	$VI = 10\text{ mV}$, $C_L = 20\text{ pF}$ (Note 12)	2.0		MHz
ϕ_m	Phase Margin	$VI = 10\text{ mV}$, $C_L = 20\text{ pF}$ (Note 12)	43		Deg
e_n	Input-Referred Voltage Noise	$f = 1\text{ kHz}$, $R_S = 20\Omega$	25		$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$	0.0015		$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
f_{max}	Full Power Bandwidth	$V_S = 10\text{V}$, $C_L = 20\text{ pF}$, $R_L = 20\text{ k}\Omega$	120		kHz
	Amp-to-Amp Isolation	(Note 9)	150		dB
THD	Total Harmonic Distortion	$A_V = +1$, $V_{IN} = 2.5\text{ V}_{PP}$ $f = 1\text{ kHz}$	0.015		%

10V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^\circ\text{C}$, $V^+ = 10\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, R_L to ground and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
V_{OS}	Input Offset Voltage	$V_O = 1.4\text{V}$, $R_S = 50\text{k}\Omega$, $R_L = 10\text{k}\Omega$, $V_{CM} = 0\text{V}$	2.1	7 9	mV max
TCV_{OS}	Temp. Coefficient of Input Offset Voltage	$T_A = 0^\circ\text{C}$ to 70°C	3.6		$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current		1	64	pA max
I_{OS}	Input Offset Current		0.5	32	pA max
CMRR	Common Mode Rejection Ratio	$V_{CM} = -0.2\text{V}$ to 8.5V	77	65 60	dB min
PSRR	Power Supply Rejection Ratio	$V^+ = 5\text{V}$ to 10V , $V_O = 1.4\text{V}$	88	65 60	dB min
V_{CM}	Input Common-Mode Voltage Range	CMRR ≥ 50 dB	9.2	9 8.5	V min
			-0.3	-0.2 -0.2	V max
A_V	Large Signal Voltage Gain	$V_O = 1\text{V}$ to 6V , $R_L = 10\text{k}\Omega$	95	85 78	dB min
V_O	Output Swing	$R_L = 10\text{ k}\Omega$, $V_{ID} = 100\text{ mV}$ (Note 11)	9.93	9.85 9.75	V min
		$V_{ID} = -100\text{ mV}$ (Note 11)	33	45 50	mV max
I_{SC}	Output Short Circuit Current	Sourcing, $V_{ID} = 100\text{ mV}$ (Note 11)	55		mA
		Sinking, $V_{ID} = -100\text{ mV}$ (Note 11)	25		mA
I_S	Total Supply Current		2.25	3.6 4.0	mA max

10V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for $T_J = 25^\circ\text{C}$, $V^+ = 10\text{V}$, $V^- = 0\text{V}$, $V_{CM} = V_O = V^+/2$, R_L to ground and $R_L > 1\text{ M}\Omega$. **Boldface** limits apply at the temperature extremes

Symbol	Parameter	Conditions	Typ (Note 5)	LMC272C Limit (Note 6)	Units
SR	Slew Rate (Note 8)	$A_V = +1$, $R_L = 10\text{k}\Omega$, $V_I = 1\text{V}_{PP}$, $C_L = 20\text{ pF}$ (Note 12)	2.65		$\text{V}/\mu\text{s}$
		$A_V = +1$, $R_L = 10\text{k}\Omega$, $V_I = 5.5\text{V}_{PP}$, $C_L = 20\text{ pF}$ (Note 12)	2.65		
GBW	Unity Gain Frequency	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$ (Note 12)	2.1		MHz
ϕ_m	Phase Margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$ (Note 12)	44		Deg
e_n	Input-Referred Voltage Noise	$f = 1\text{ kHz}$, $R_S = 20\Omega$	25		$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
i_n	Input-Referred Current Noise	$f = 1\text{ kHz}$	0.0015		$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
f_{max}	Full Power Bandwidth	$C_L = 20\text{ pF}$, $R_L = 20\text{k}\Omega$	120		kHz
	Amp-to-Amp Isolation	(Note 9)	150		dB
THD	Total Harmonic Distortion	$A_V = +1$, $V_{IN} = 5\text{V}_{PP}$ $f = 1\text{ kHz}$	0.005		%

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical characteristics.

Note 2: Human body model, $1.5\text{k}\Omega$ in series with 100 pF .

Note 3: Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C . Output currents in excess of $\pm 30\text{ mA}$ over long term may adversely affect reliability.

Note 4: The maximum power dissipation is a function of $T_{J(max)}$, θ_{JA} , and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(max)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

Note 7: Do not short circuit output to V^+ , when V^+ is greater than 13V or reliability will be adversely affected.

Note 8: Slew rate is the slower of the rising and falling slew rates.

Note 9: Input referred, $V^+ = 10\text{V}$ and $R_L = 100\text{k}\Omega$ connected to 5V . Each amp excited in turn with 1 kHz to produce about 10V_{PP} output.

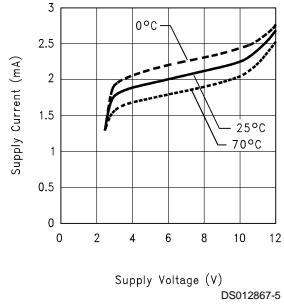
Note 10: Limiting input pin current is only necessary for input voltages that exceed absolute maximum input voltage ratings.

Note 11: V_{ID} is the differential voltage on the non-inverting input with respect to the inverting input.

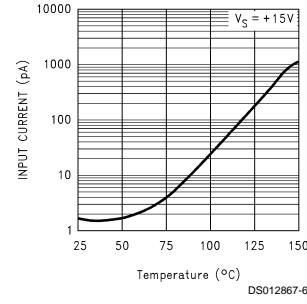
Note 12: V_I is the input voltage.

Typical Performance Characteristics ($V_S = +5V$, single supply, $T_A = 25^\circ C$, and R_L to ground unless otherwise specified)

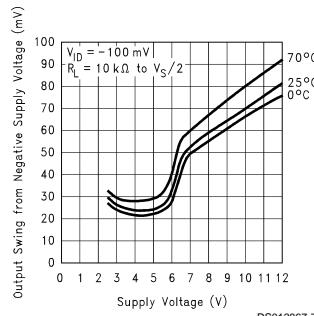
Supply Current vs Supply Voltage



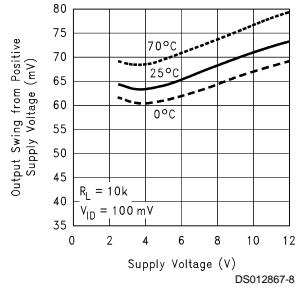
Input Current vs Temperature



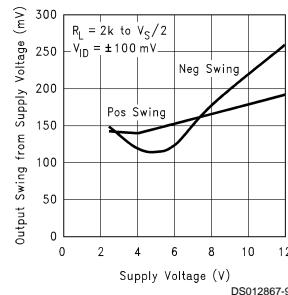
Negative Output Voltage Swing vs Supply Voltage



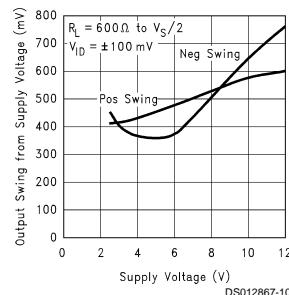
Positive Output Voltage Swing vs Supply Voltage



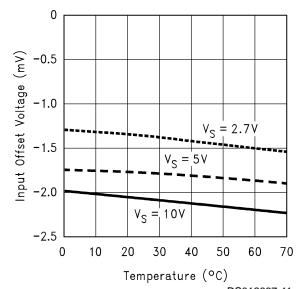
Output Voltage Swing vs Supply Voltage



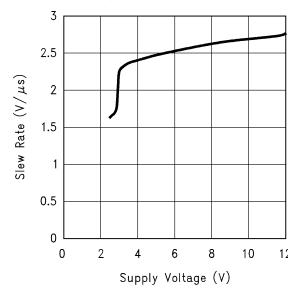
Output Voltage Swing vs Supply Voltage



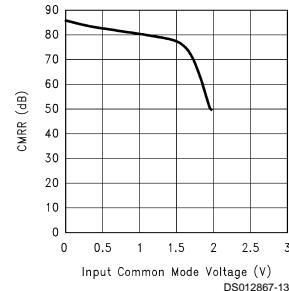
Input Offset Voltage vs Temperature



Slew Rate vs Supply Voltage

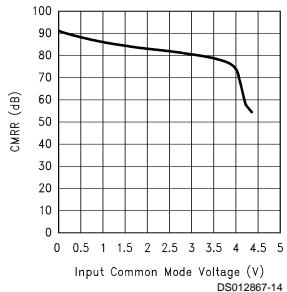


CMRR vs Input Common Mode Voltage ($V_S = 2.7V$)

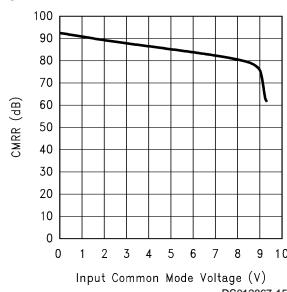


Typical Performance Characteristics ($V_S = +5V$, single supply, $T_A = 25^\circ C$, and R_L to ground unless otherwise specified) (Continued)

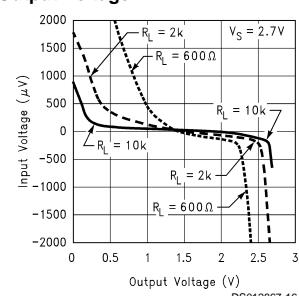
CMRR vs Input Common Mode Voltage ($V_S = 5V$)



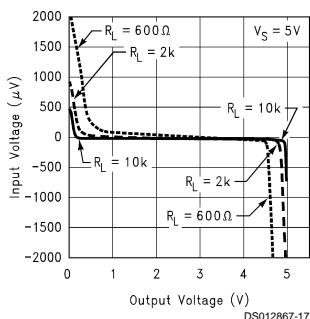
CMRR vs Input Common Mode Voltage ($V_S = 10V$)



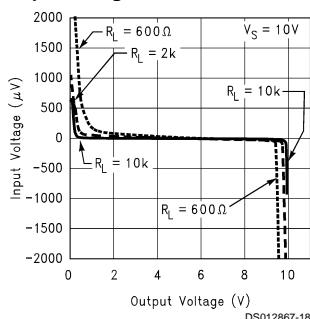
Input Voltage vs Output Voltage



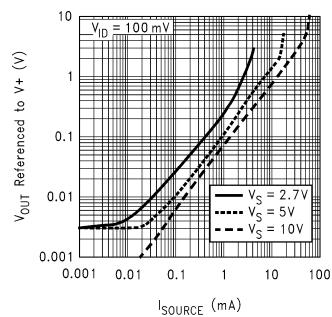
Input Voltage vs Output Voltage



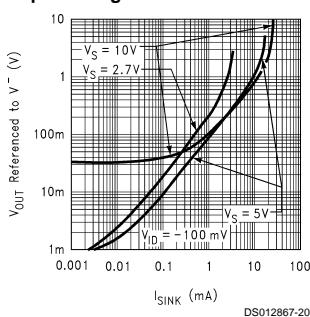
Input Voltage vs Output Voltage



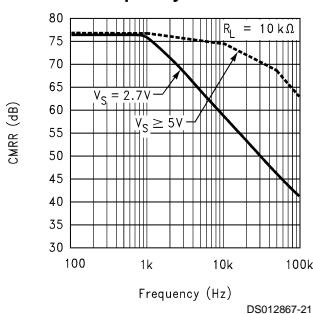
Sourcing Current vs Output Voltage



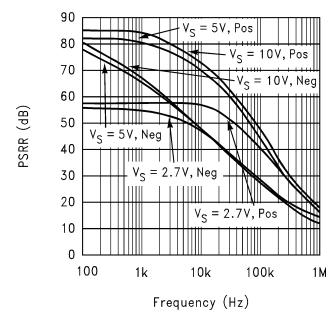
Sinking Current vs Output Voltage



CMRR vs Frequency

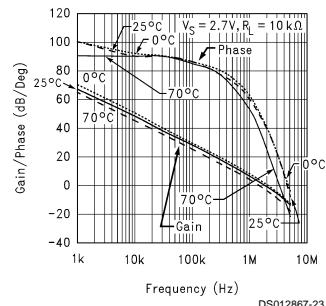


PSRR vs Frequency

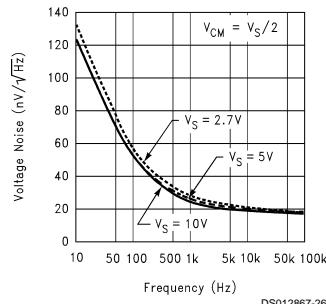


Typical Performance Characteristics ($V_S = +5V$, single supply, $T_A = 25^\circ C$, and R_L to ground unless otherwise specified) (Continued)

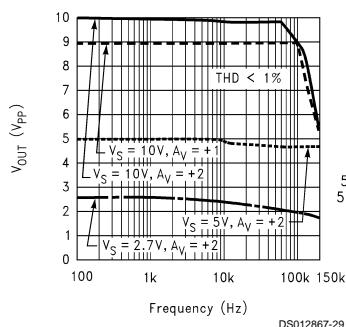
Gain/Phase Response vs Temperature



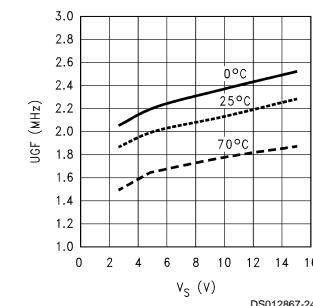
Input Voltage Noise vs Frequency



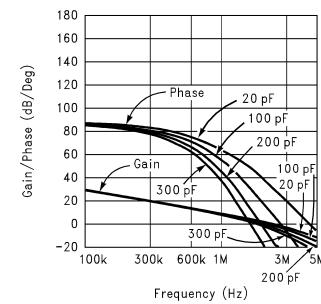
Output Swing vs Frequency



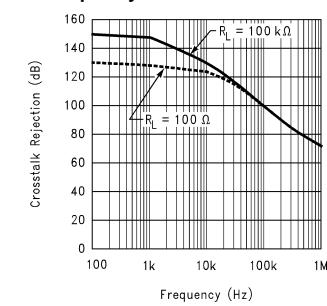
Unity Gain Frequency vs Temperature



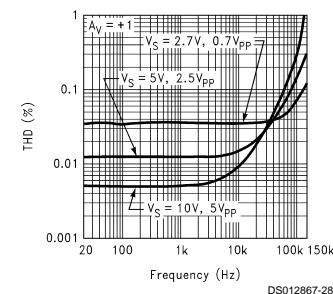
Gain/Phase vs Capacitive Load



Crosstalk Rejection vs Frequency

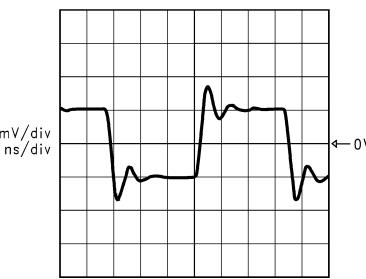


THD vs Frequency



Small Signal Step Response

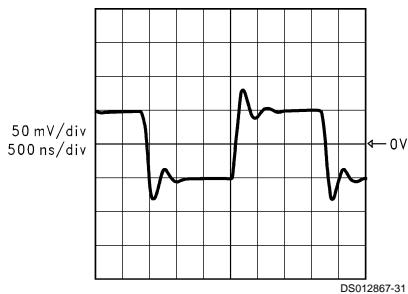
$V_S = \pm 1.35V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 0.1 V_{PP}$



Typical Performance Characteristics ($V_S = +5V$, single supply, $T_A = 25^\circ C$, and R_L to ground unless otherwise specified) (Continued)

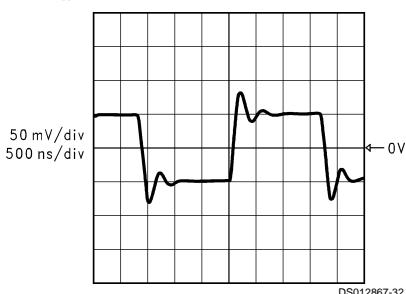
Small Signal Step Response

$V_S = \pm 2.5V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 0.1 V_{PP}$



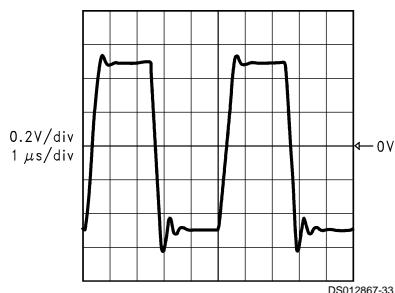
Small Signal Step Response

$V_S = \pm 5V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 0.1 V_{PP}$



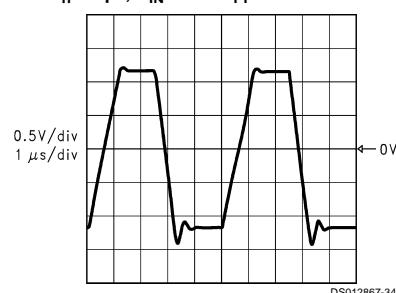
Large Signal Step Response

$V_S = \pm 1.35V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 1 V_{PP}$



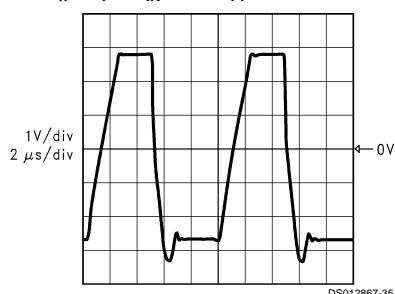
Large Signal Step Response

$V_S = \pm 2.5V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 2.4 V_{PP}$



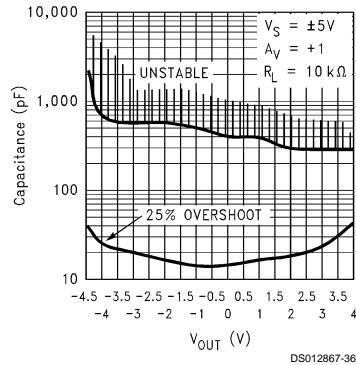
Large Signal Step Response

$V_S = \pm 5V$, $A_V = +1$,
 $Z_L = 10 k\Omega \parallel 20 pF$, $V_{IN} = 5.5 V_{PP}$

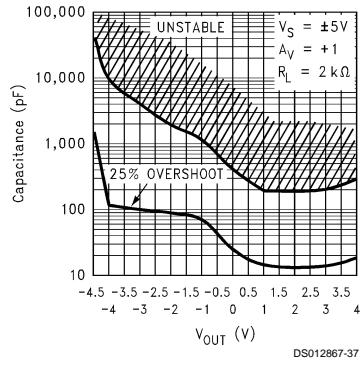


Typical Performance Characteristics ($V_S = \pm 5V$, single supply, $T_A = 25^\circ C$, and R_L to ground unless otherwise specified) (Continued)

Stability vs Capacitive Load

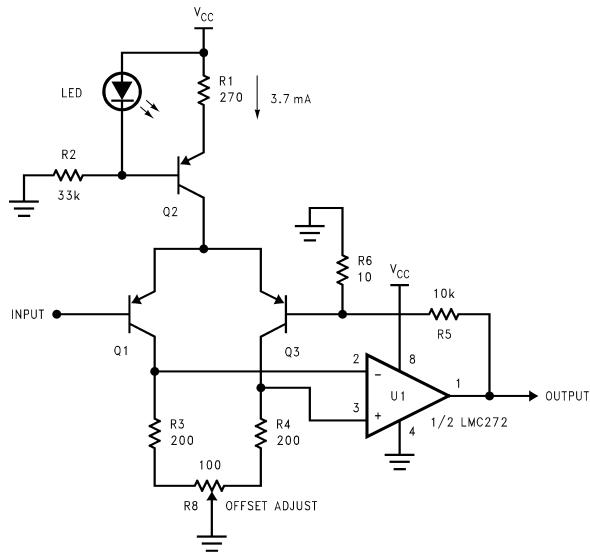


Stability vs Capacitive Load



Application Information

Low Noise Single Supply Preamplifier



It is generally difficult to find already existing solutions in the market which are single supply and low noise. The circuit above is a low noise single supply preamp using the LMC272. It utilizes the feature of input common mode voltage range to ground to achieve zero-volt-in zero-volt-out performance and uses the RR output swing to achieve maximum dynamic range. By introducing a differential pair operating at high bias current as the front end, the equivalent input noise voltage, e_n , is reduced. The gain is $1 + R5/R6$

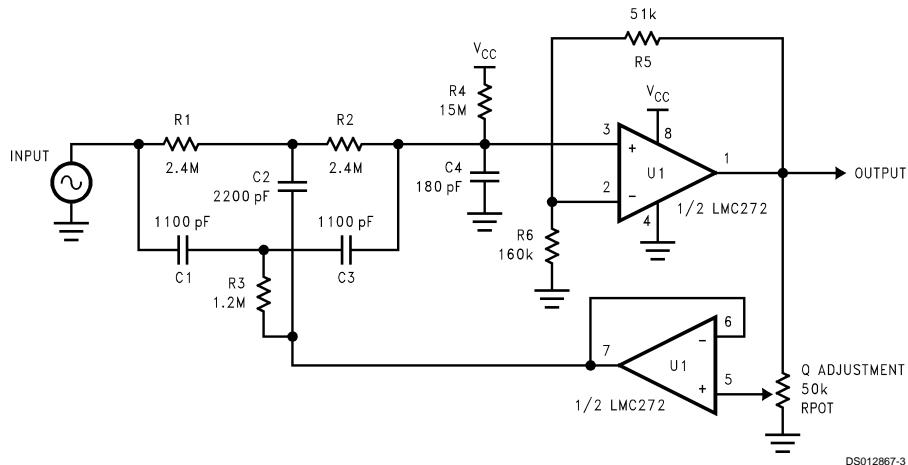
which is a 1000 in this case. There is an inherent trade off between noise voltage and power consumption, input bias current, and input noise current. Input equivalent noise current is inconsequential if the source impedance is small. $R1$ can be adjusted to vary bias current. To avoid saturation, $R3$ and $R4$ should be set such that $Q1$ and $Q3$ collector voltages do not exceed 0.5V. *Table 1* shows typical noise data for two different $R1$ settings:

Application Information (Continued)

TABLE 1. Equivalent Input Noise Voltage, e_n , for Two Different Values of R1

Ω	mA	nV/ $\sqrt{\text{Hz}}$		
R1	$I_C(Q1, 3)$	e_n (100 Hz)	e_n (1 kHz)	e_n (10 kHz)
270	1.85	3.2	2.0	1.7
1000	0.50	5.3	2.4	1.9

Single Supply Twin-T Notch Filter with "Q" Adjustment



Here is another application for the LMC272. This is a single supply notch filter set for 60 Hz using the component values shown, but the frequency can be changed using the equations below. The main feature of this circuit is its ability to adjust the filter selectivity (Q) using RPOT. You can trade off notch depth for Q. Table 2 shows data for two different settings. The LMC272 lends itself nicely to general purpose applications like this because it is very well behaved and easy to use. This filter can operate from 2.7V to 15V supplies. Component value matching is important to achieve good results. Here R4 is used to set the input to within the common mode range of the device to allow maximum swing on the non-inverting input (pin 3). Since R1, R2, and R4 form a voltage divider at low frequencies, C4 is added to introduce a high frequency attenuation in conjunction with C1, and C3. R5 and R6 were picked to set the pass band gain to 0 dB.

$$R = R1 = R2 = 2R3$$

$$C = C1 = C3 = C2/2$$

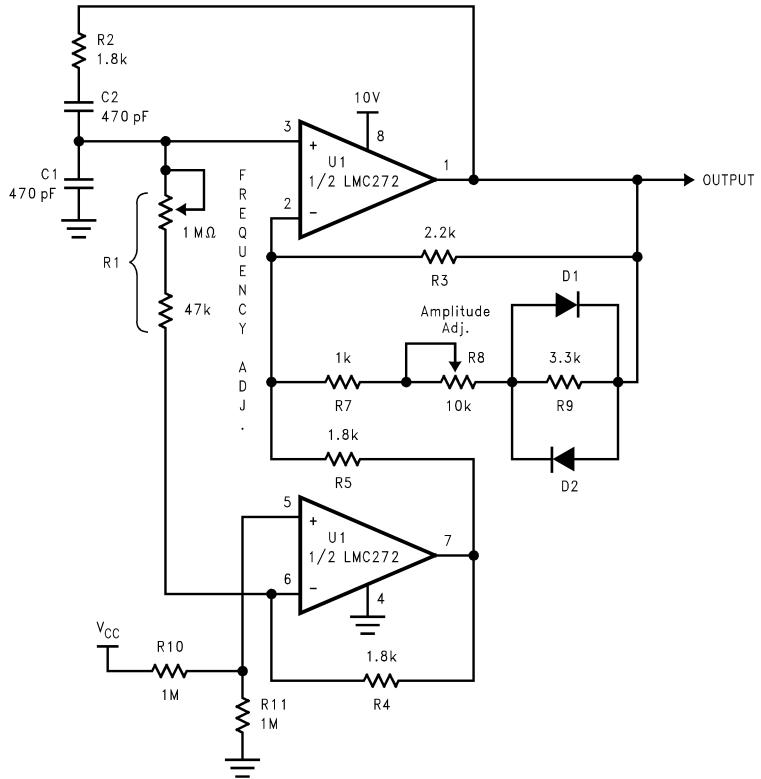
$$f(\text{notch}) = \frac{1}{2\pi RC}; C4 = \frac{R \cdot C}{R4}, Q = \frac{f(\text{notch})}{BW}$$

TABLE 2. Filter Selectivity (Q) vs Notch Depth

Q	(dB)
Notch Depth	
0.3	40
6	17

Application Information (Continued)

Single Supply Wein_Bridge Oscillator with Amplitude and Frequency Adjustment



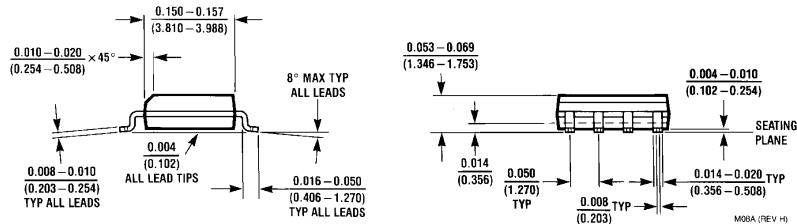
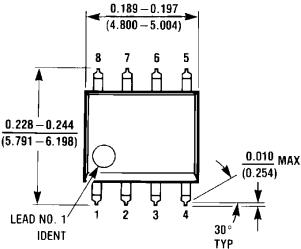
DS012867-4

$$f_{osc} \approx \frac{1}{2\pi\sqrt{C_1 C_2 R_1 R_2}}$$

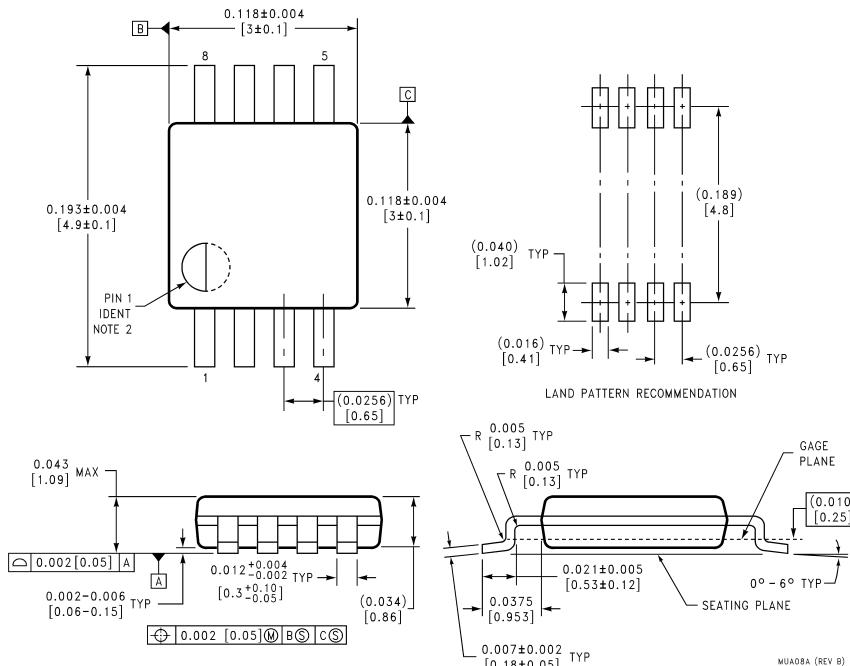
f(range) = 6.4 kHz to 30 kHz

Amplitude Adjustment (range) = 2.8 V_{PP} to 8.6 V_{PP}

Physical Dimensions inches (millimeters) unless otherwise noted



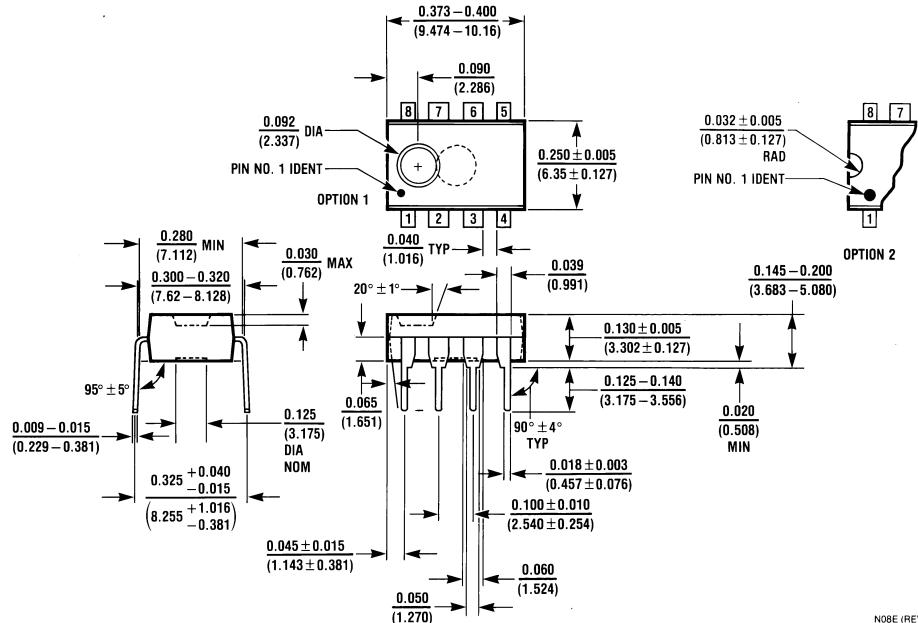
**8-Lead (0.150" Wide) Molded Small Outline Package, JEDEC
Order Number LMC272CM or LMC272CMX
NS Package Number M08A**



**8-Lead (0.118" Wide) Molded Mini Small Outline Package
Order Number LMC272CMM or LMC272CMMX
NS Package Number MUA08A**

LMC272 CMOS Dual Low Cost Rail to Rail Output Operational Amplifier

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



N08E (REV F)

**8-Lead (0.300" Wide) Molded Dual-In-Line Package
Order Number LMC272CN
NS Package Number N08E**

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