# 0.9 V, Rail-to-Rail, Single **Operational Amplifier**

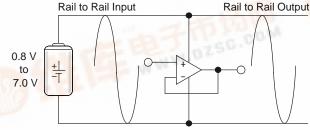
The NCS2001 is an industry first sub-one voltage operational amplifier that features a rail-to-rail common mode input voltage range, along with rail-to-rail output drive capability. This amplifier is guaranteed to be fully operational down to 0.9 V, providing an ideal solution for powering applications from a single cell Nickel Cadmium (NiCd) or Nickel Metal Hydride (NiMH) battery. Additional features include no output phase reversal with overdriven inputs, trimmed input offset voltage of 0.5 mV, extremely low input bias current of 40 pA, and a unity gain bandwidth of 1.4 MHz at 5.0 V. The tiny NCS2001 is the ideal solution for small portable electronic applications and is available in the space saving SOT23-5 and SC70-5 packages with two industry standard pinouts.

#### **Features**

- 0.9 V Guaranteed Operation
- Rail-to-Rail Common Mode Input Voltage Range
- Rail-to-Rail Output Drive Capability
- No Output Phase Reversal for Over–Driven Input Signals
- 0.5 mV Trimmed Input Offset
- 10 pA Input Bias Current
- 1.4 MHz Unity Gain Bandwidth at  $\pm 2.5$  V, 1.1 MHz at  $\pm 0.5$  V
- Tiny SC70-5 and SOT23-5 Packages
- Pb-Free Package is Available

#### **Typical Applications**

- Single Cell NiCd/NiMH Battery Powered Applications
- Cellular Telephones
- Personal Digital Assistants WIMM DESGREE
   Electronic Games
- Digital Cameras
- Camcorders
- Hand–Held Instruments



This device contains 63 active transistors.

Figure 1. Typical Application



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#### **MARKING DIAGRAMS**



SOT23-5 **SN SUFFIX CASE 483** 





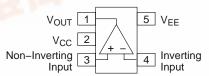
SC70-5 **SQ SUFFIX CASE 419A** 



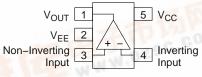
x = G for SN1H for SN2 I for SQ1 J for SQ2

Y = Year W = Work Week M = Date Code

#### **PIN CONNECTIONS**



Style 1 Pinout (SN1T1, SQ1T1)



Style 2 Pinout (SN2T1, SQ2T1)

#### ORDERING INFORMATION

See detailed ordering and shipping information in the dimensions section on page 14 of this data sheet.

#### **MAXIMUM RATINGS**

Rating	Symbol	Value	Unit
Supply Voltage (V <sub>CC</sub> to V <sub>EE</sub> )	Vs	7.0	V
Input Differential Voltage Range (Note 1)	$V_{IDR}$	$V_{\text{EE}}$ –300 mV to 7.0 V	V
Input Common Mode Voltage Range (Note 1)	V <sub>ICR</sub>	$V_{\text{EE}}$ –300 mV to 7.0 V	V
Output Short Circuit Duration (Note 2)	t <sub>Sc</sub>	Indefinite	sec
Junction Temperature	$T_J$	150	°C
Power Dissipation and Thermal Characteristics SOT23–5 Package Thermal Resistance, Junction–to–Air Power Dissipation @ T <sub>A</sub> = 70°C SC70–5 Package Thermal Resistance, Junction–to–Air Power Dissipation @ T <sub>A</sub> = 70°C	$egin{array}{c} R_{ heta JA} \ P_D \ \end{array}$ $egin{array}{c} R_{ heta JA} \ P_D \ \end{array}$	235 340 280 286	°C/W mW °C/W mW
Storage Temperature Range	T <sub>stg</sub>	-65 to 150	°C
ESD Protection at any Pin Human Body Model (Note 3)	V <sub>ESD</sub>	2000	V

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.

- 1. Either or both inputs should not exceed the range of  $V_{EE}$  –300 mV to  $V_{EE}$  +7.0 V.
- 2. Maximum package power dissipation limits must be observed to ensure that the maximum junction temperature is not exceeded.  $T_J = T_A + (P_D R_{\theta JA}).$ 3. ESD data available upon request.

### DC ELECTRICAL CHARACTERISTICS

(V<sub>CC</sub> = 2.5 V, V<sub>EE</sub> = -2.5 V, V<sub>CM</sub> = V<sub>O</sub> = 0 V, R<sub>L</sub> to GND,  $T_A$  = 25°C unless otherwise noted.)

Characteristics	Symbol	Min	Тур	Max	Unit
Input Offset Voltage V <sub>CC</sub> = 0.45 V, V <sub>FF</sub> = -0.45 V	V <sub>IO</sub>				mV
T <sub>A</sub> = 25°C		-6.0	0.5	6.0	
$T_A = 0$ °C to $70$ °C		-8.5	_	8.5	
$T_A = -40$ °C to 105°C		-9.5	_	9.5	
$V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}$					
$T_A = 25^{\circ}C$		-6.0	0.5	6.0	
$T_A = 0$ °C to 70°C		-7.0	_	7.0	
$T_A = -40^{\circ}\text{C} \text{ to } 105^{\circ}\text{C}$		-7.5	_	7.5	
$V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$ $T_{\Delta} = 25^{\circ}\text{C}$		-6.0	0.5	6.0	
$T_A = 25 \text{ C}$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$		-0.0 -7.5	0.5	7.5	
$T_{\Delta} = -40^{\circ}\text{C to } 105^{\circ}\text{C}$		-7.5 -7.5	_	7.5 7.5	
Th.					
Input Offset Voltage Temperature Coefficient ( $R_S = 50$ ) $T_A = -40$ °C to 105°C	ΔV <sub>IO</sub> /ΔT	_	8.0	_	μV/°C
Input Bias Current (V <sub>CC</sub> = 1.0 V to 5.0 V)	I <sub>IB</sub>	-	10	-	pA
Input Common Mode Voltage Range	V <sub>ICR</sub>	-	$V_{EE}$ to $V_{CC}$	-	V
Large Signal Voltage Gain	A <sub>VOL</sub>				kV/V
$V_{CC} = 0.45 \text{ V}, V_{EE} = -0.45 \text{ V}$					
$R_L = 10 \text{ k}$		_	40	_	
$R_{L} = 2.0 \text{ k}$		_	20	_	
$V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}$ $R_{I} = 10 \text{ k}$			40		
$R_1 = 10 \text{ k}$ $R_1 = 2.0 \text{ k}$		] _	40 40		
$V_{CC} = 2.5 \text{ V}, V_{FF} = -2.5 \text{ V}$		_	40		
$R_1 = 10 \text{ k}$		20	40	_	
R <sub>L</sub> = 2.0 k		15	40	-	

DC ELECTRICAL CHARACTERISTICS (continued)  $(V_{CC}=2.5 \text{ V}, \text{ V}_{EE}=-2.5 \text{ V}, \text{ V}_{CM}=\text{V}_{O}=0 \text{ V}, \text{ R}_{L} \text{ to GND, T}_{A}=25^{\circ}\text{C} \text{ unless otherwise noted.} )$ 

Characteristics	Symbol	Min	Тур	Max	Unit
Output Voltage Swing, High State Output (V <sub>ID</sub> = +0.5 V)	V <sub>OH</sub>				V
$V_{CC} = 0.45 \text{ V}, V_{EE} = -0.45 \text{ V}$					
$T_A = 25^{\circ}C$ $R_L = 10 \text{ k}$		0.40	0.494	_	
$R_L = 10 \text{ k}$ $R_L = 2.0 \text{ k}$		0.40	0.466	_	
$T_A = 0$ °C to 70°C					
$R_L = 10 \text{ k}$		0.40	_	-	
$R_L = 2.0 \text{ k}$		0.35	_	_	
$T_A = -40^{\circ}\text{C} \text{ to } 105^{\circ}\text{C}$ $R_L = 10 \text{ k}$		0.40	_	_	
R <sub>L</sub> = 2.0 k		0.35	_	_	
$V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}$					
$T_A = 25^{\circ}C$		4.45	1 400		
R <sub>L</sub> = 10 k R <sub>L</sub> = 2.0 k		1.45 1.40	1.498 1.480	_	
$T_A = 0$ °C to 70°C					
R <sub>L</sub> = 10 k		1.45	_	_	
R <sub>L</sub> = 2.0 k		1.40	_	_	
$T_A = -40^{\circ}C$ to $105^{\circ}C$ $R_L = 10 \text{ k}$		1.45	_	_	
$R_L = 2.0 \text{ k}$		1.40	_	_	
$V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$					
$T_A = 25^{\circ}C$					
R <sub>L</sub> = 10 k R <sub>L</sub> = 2.0 k		2.45 2.40	2.498 2.475	_	
$T_A = 0^{\circ}C$ to $70^{\circ}C$		2.40	2.475	_	
R <sub>L</sub> = 10 k		2.45	_	_	
$R_L = 2.0 \text{ k}$		2.40	_	-	
$T_A = -40^{\circ}\text{C to } 105^{\circ}\text{C}$		0.45			
$R_L = 10 \text{ k}$ $R_L = 2.0 \text{ k}$		2.45 2.40	_	_	
Output Voltage Swing, Low State Output ( $V_{ID} = -0.5 \text{ V}$ )	V <sub>OL</sub>				V
V <sub>CC</sub> = 0.45 V, V <sub>EE</sub> = $-0.45$ V	V OL				v
T <sub>A</sub> = 25°C					
$R_L = 10 \text{ k}$		-	-0.494	-0.40	
$R_L = 2.0 \text{ k}$ $T_A = 0^{\circ}\text{C to } 70^{\circ}\text{C}$		_	-0.480	-0.35	
$R_L = 10 \text{ k}$		_	_	-0.40	
$R_L = 2.0 \text{ k}$		_	-	-0.35	
$T_A = -40^{\circ}$ C to 105°C					
R <sub>L</sub> = 10 k R <sub>L</sub> = 2.0 k		_	_	-0.40	
V <sub>CC</sub> = 1.5 V, V <sub>EE</sub> = -1.5 V		_	_	-0.35	
T <sub>A</sub> = 25°C					
$R_L = 10 \text{ k}$		-	-1.493	-1.45	
$R_L = 2.0 \text{ k}$		_	-1.480	-1.40	
T <sub>A</sub> = 0°C to 70°C R <sub>L</sub> = 10 k		_	_	-1.45	
$R_L = 2.0 \text{ k}$		_	_	-1.40	
$T_A = -40$ °C to 105°C					
$R_L = 10 \text{ k}$		-	-	-1.45	
$R_L = 2.0 \text{ k}$ $V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$		_	_	-1.40	
$V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$ $T_A = 25^{\circ}\text{C}$		_			
$R_L = 10 \text{ k}$		-	-2.492	-2.45	
$R_L = 2.0 \text{ k}$			-2.479	-2.40	
$T_A = 0$ °C to 70°C $R_L = 10 \text{ k}$		_	_	-2.45	
$R_L = 10 \text{ k}$ $R_L = 2.0 \text{ k}$		_	_	-2.45 -2.40	
$T_A = -40$ °C to 105°C					
$R_L = 10 \text{ k}$		-	-	-2.45	
R <sub>L</sub> = 2.0 k		_	_	-2.40	

## DC ELECTRICAL CHARACTERISTICS (continued)

( $V_{CC}$  = 2.5 V,  $V_{EE}$  = -2.5 V,  $V_{CM}$  =  $V_{O}$  = 0 V,  $R_{L}$  to GND,  $T_{A}$  = 25°C unless otherwise noted.)

Characteristics	Symbol	Min	Тур	Max	Unit
Common Mode Rejection Ratio (V <sub>in</sub> = 0 to 5.0 V)	CMRR	60	70	-	dB
Power Supply Rejection Ratio ( $V_{CC} = 0.5 \text{ V}$ to 2.5 V, $V_{EE} = -2.5 \text{ V}$ )	PSRR	55	65	-	dB
Output Short Circuit Current	I <sub>SC</sub>				mA
$V_{CC} = 0.45 \text{ V}, V_{EE} = -0.45 \text{ V}, V_{ID} = \pm 0.4 \text{ V}$		0.5	4.0		
Source Current High Output State		0.5	1.2	-	
Sink Current Low Output State $V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}, V_{ID} = \pm 0.5 \text{ V}$		_	-3.0	-1.5	
Source Current High Output State		15	29	_	
Sink Current Low Output State		_	-40	-20	
$V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}, V_{ID} = \pm 0.5 \text{ V}$					
Source Current High Output State		40	76	_	
Sink Current Low Output State		-	-96	-50	
Power Supply Current (Per Amplifier, V <sub>O</sub> = 0 V)	I <sub>D</sub>				mA
$V_{CC} = 0.45 \text{ V}, V_{EE} = -0.45 \text{ V}$					
$T_A = 25$ °C		_	0.51	1.10	
$T_A = 0$ °C to $70$ °C		-	-	1.10	
$T_A = -40$ °C to $105$ °C		_	_	1.10	
$V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}$			0.70	4.40	
$T_A = 25^{\circ}C$		_	0.72	1.40	
$T_A = 0$ °C to 70°C $T_A = -40$ °C to 105°C		_	_	1.40 1.40	
$V_{CC} = 2.5 \text{ V}, V_{FF} = -2.5 \text{ V}$		_	_	1.40	
$V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$ $T_{\Delta} = 25 \text{ °C}$		_	0.82	1.50	
$T_A = 0^{\circ}$ C to $70^{\circ}$ C		_	-	1.50	
$T_A = -40^{\circ}\text{C to } 105^{\circ}\text{C}$		_	_	1.50	

## AC ELECTRICAL CHARACTERISTICS

(V<sub>CC</sub> = 2.5 V, V<sub>EE</sub> = -2.5 V, V<sub>CM</sub> = V<sub>O</sub> = 0 V, R<sub>L</sub> to GND,  $T_A$  = 25°C unless otherwise noted.)

Characteristics	Symbol	Min	Тур	Max	Unit
Differential Input Resistance (V <sub>CM</sub> = 0 V)	R <sub>in</sub>	_	>1.0	_	tera Ω
Differential Input Capacitance (V <sub>CM</sub> = 0 V)	C <sub>in</sub>	_	3.0	_	pF
Equivalent Input Noise Voltage (f = 1.0 kHz)	e <sub>n</sub>	_	100	_	nV/√Hz
Gain Bandwidth Product (f = 100 kHz) $V_{CC} = 0.45 \text{ V}, V_{EE} = -0.45 \text{ V}$ $V_{CC} = 1.5 \text{ V}, V_{EE} = -1.5 \text{ V}$ $V_{CC} = 2.5 \text{ V}, V_{EE} = -2.5 \text{ V}$	GBW	- - 0.5	1.1 1.3 1.4	- - -	MHz
Gain Margin ( $R_L = 10 \text{ k}$ , $C_L = 5.0 \text{ pf}$ )	Am	_	6.5	_	dB
Phase Margin (R <sub>L</sub> = 10 k, C <sub>L</sub> = 5.0 pf)	φm	_	60	_	0
Power Bandwidth ( $V_O = 4.0 V_{pp}$ , $R_L = 2.0 k$ , THD = 1.0%, $A_V = 1.0$ )	BW <sub>P</sub>	_	80	_	kHz
Total Harmonic Distortion ( $V_O$ = 4.0 $V_{pp}$ , $R_L$ = 2.0 k, $A_V$ = 1.0) f = 1.0 kHz f = 10 kHz	THD	_ _	0.008 0.08	- -	%
Slew Rate (V <sub>S</sub> = $\pm 2.5$ V, V <sub>O</sub> = $-2.0$ V to 2.0 V, R <sub>L</sub> = 2.0 k, A <sub>V</sub> = 1.0) Positive Slope Negative Slope	SR	1.0 1.0	1.6 1.6	6.0 6.0	V/μs

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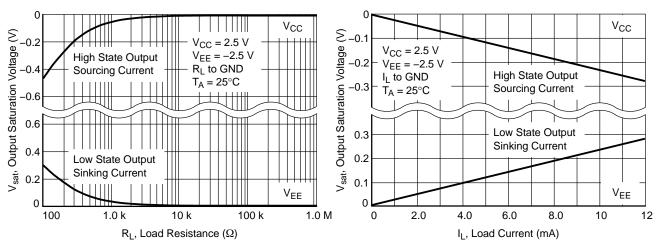


Figure 2. Split Supply Output Saturation vs. Load Resistance

Figure 3. Split Supply Output Saturation vs. Load Current

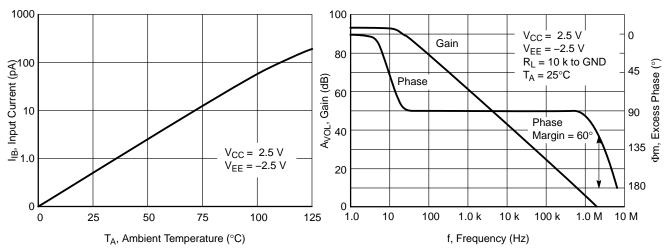


Figure 4. Input Bias Current vs. Temperature

Figure 5. Gain and Phase vs. Frequency

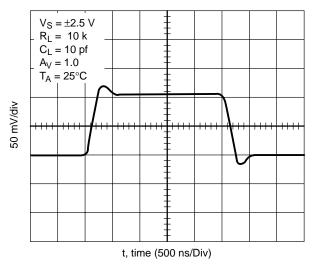


Figure 6. Transient Response

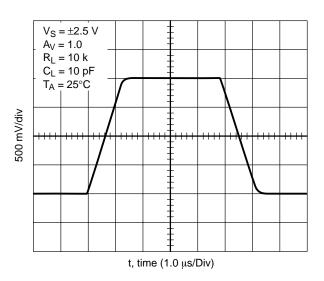


Figure 7. Slew Rate

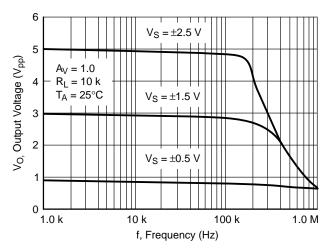


Figure 8. Output Voltage vs. Frequency

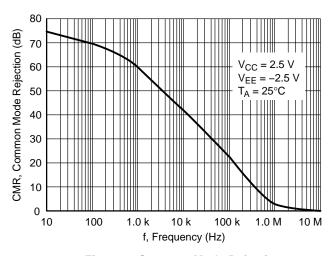


Figure 9. Common Mode Rejection vs. Frequency

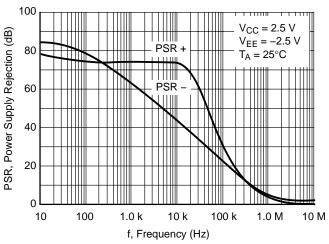


Figure 10. Power Supply Rejection vs. Frequency

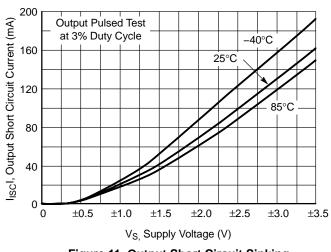


Figure 11. Output Short Circuit Sinking Current vs. Supply Voltage

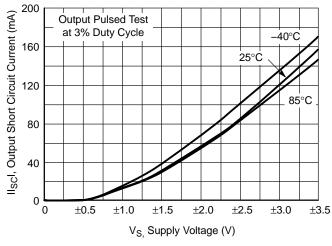


Figure 12. Output Short Circuit Sourcing Current vs. Supply Voltage

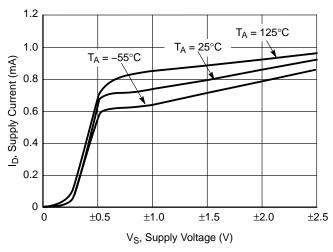
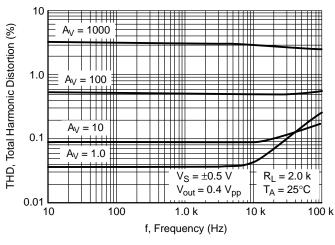


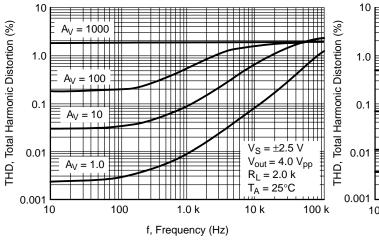
Figure 13. Supply Current vs. Supply Voltage



 $A_{V} = 1000$ THD, Total Harmonic Distortion (%) 1.0  $A_{V} = 100$  $A_{V} = 10$ 0.1  $A_V = 1.0$  $V_S = \pm 0.5 \text{ V}$  $R_L = 10 k$ T<sub>A</sub> = 25°C  $V_{out} = 0.4 V_{pp}$ 0.01 10 100 1.0 k 10 k 100 k f, Frequency (Hz)

Figure 14. Total Harmonic Distortion vs. Frequency with 1.0 V Supply

Figure 15. Total Harmonic Distortion vs. Frequency with 1.0 V Supply



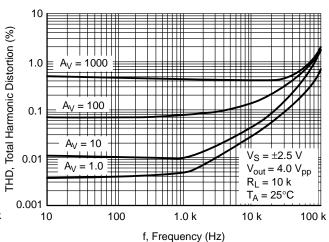
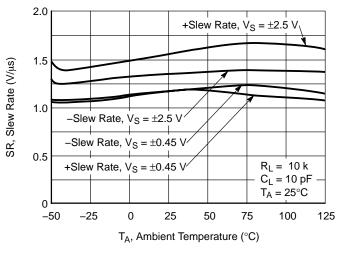


Figure 16. Total Harmonic Distortion vs. Frequency with 5.0 V Supply

Figure 17. Total Harmonic Distortion vs. Frequency with 5.0 V Supply



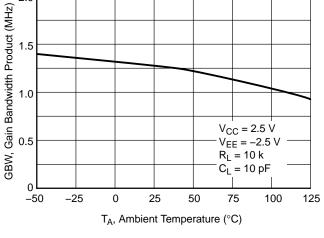


Figure 18. Slew Rate vs. Temperature

Figure 19. Gain Bandwidth Product vs. Temperature

2.0

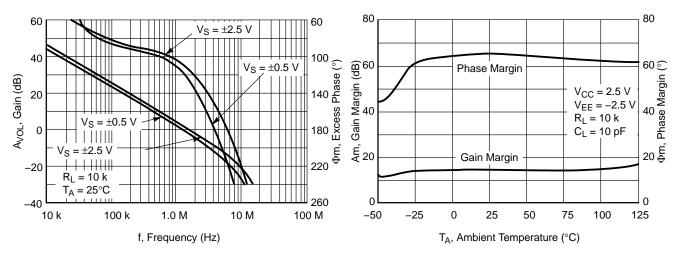


Figure 20. Voltage Gain and Phase vs. Frequency

Figure 21. Gain and Phase Margin vs. Temperature

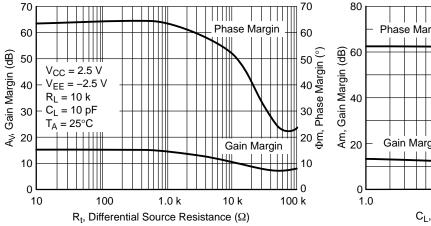


Figure 22. Gain and Phase Margin vs.
Differential Source Resistance

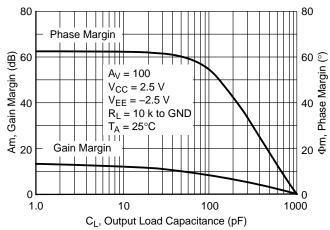


Figure 23. Gain and Phase Margin vs.
Output Load Capacitance

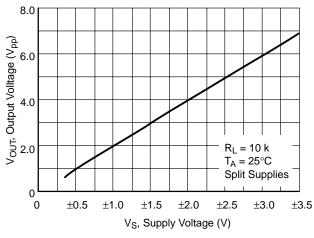


Figure 24. Output Voltage Swing vs. Supply Voltage

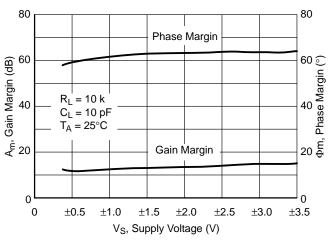


Figure 25. Gain and Phase Margin vs. Supply Voltage

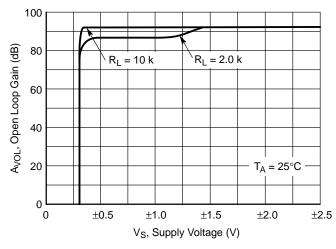


Figure 26. Open Loop Voltage Gain vs. Supply Voltage

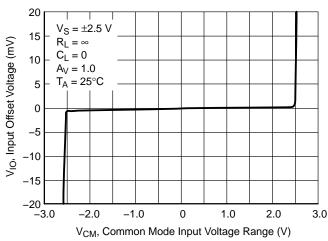


Figure 27. Input Offset Voltage vs. Common Mode Input Voltage Range  $V_S = \pm 2.5 \text{ V}$ 

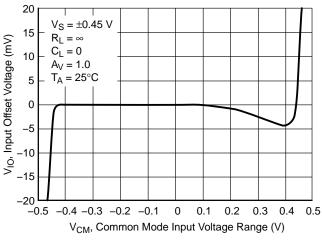


Figure 28. Input Offset Voltage vs. Common Mode Input Voltage Range,  $V_S = \pm 0.45 \text{ V}$ 

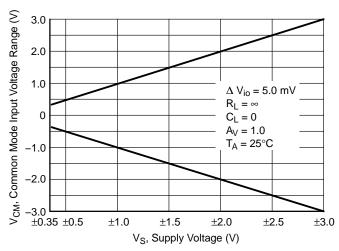


Figure 29. Common-Mode Input Voltage Range vs. Power Supply Voltage

#### APPLICATION INFORMATION AND OPERATING DESCRIPTION

#### **GENERAL INFORMATION**

The NCS2001 is an industry first rail–to–rail input, rail–to–rail output amplifier that features guaranteed sub–one voltage operation. This unique feature set is achieved with the use of a modified analog CMOS process that allows the implementation of depletion MOSFET devices. The amplifier has a 1.0 MHz gain bandwidth product,  $2.2~V/\mu s$  slew rate and is operational over a power supply range less than 0.9~V to as high as 7.0~V.

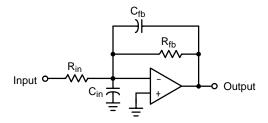
### Inputs

The input topology chosen for this device series is unconventional when compared to most low voltage operational amplifiers. It consists of an N–Channel depletion mode differential transistor pair that drives a folded cascade stage and current mirror. This configuration extends the input common mode voltage range to encompass the  $V_{EE}$  and  $V_{CC}$  power supply rails, even when powered from a combined total of less than 0.9 V. Figures 27 and 28 show the input common mode voltage range versus power supply voltage.

The differential input stage is laser trimmed in order to minimize offset voltage. The N–Channel depletion mode MOSFET input stage exhibits an extremely low input bias current of less than 10 pA. The input bias current versus temperature is shown in Figure 4. Either one or both inputs can be biased as low as  $V_{\rm EE}$  minus 300 mV to as high as 7.0 V without causing damage to the device. If the input common mode voltage range is exceeded, the output will not display a phase reversal. If the maximum input positive or negative voltage ratings are to be exceeded, a series resistor must be used to limit the input current to less than 2.0 mA.

The ultra low input bias current of the NCS2001 allows the use of extremely high value source and feedback resistor without reducing the amplifier's gain accuracy. These high value resistors, in conjunction with the device input and printed circuit board parasitic capacitances  $C_{\rm in}$ , will add an additional pole to the single pole amplifier in Figure 30. If low enough in frequency, this additional pole can reduce the phase margin and significantly increase the output settling time. The effects of  $C_{\rm in}$ , can be canceled by placing a zero into the feedback loop. This is accomplished with the addition of capacitor  $C_{\rm fb}$ . An approximate value for  $C_{\rm fb}$  can be calculated by:

$$C_{fb} = \frac{R_{in} \times C_{in}}{R_{fb}}$$



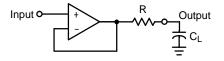
Cin = Input and printed circuit board capacitance

Figure 30. Input Capacitance Pole Cancellation

#### Output

The output stage consists of complimentary P and N–Channel devices connected to provide rail–to–rail output drive. With a 2.0~k load, the output can swing within 50~mV of either rail. It is also capable of supplying over 75~mA when powered from 5.0~V and 1.0~mA when powered from 0.9~V.

When connected as a unity gain follower, the NCS2001 can directly drive capacitive loads in excess of 820 pF at room temperature without oscillating but with significantly reduced phase margin. The unity gain follower configuration exhibits the highest bandwidth and is most prone to oscillations when driving a high value capacitive load. The capacitive load in combination with the amplifier's output impedance, creates a phase lag that can result in an under-damped pulse response or a continuous oscillation. Figure 32 shows the effect of driving a large capacitive load in a voltage follower type of setup. When driving capacitive loads exceeding 820 pF, it is recommended to place a low value isolation resistor between the output of the op amp and the load, as shown in Figure 31. The series resistor isolates the capacitive load from the output and enhances the phase margin. Refer to Figure 33. Larger values of R will result in a cleaner output waveform but excessively large values will degrade the large signal rise and fall time and reduce the output amplitude. Depending upon the capacitor characteristics, the isolation resistor value will typically be between 50 to 500  $\Omega$ . The output drive capability for resistive and capacitive loads is shown in Figures 2, 3, and 23.



Isolation resistor R = 50 to 500

Figure 31. Capacitance Load Isolation

Note that the lowest phase margin is observed at cold temperature and low supply voltage.

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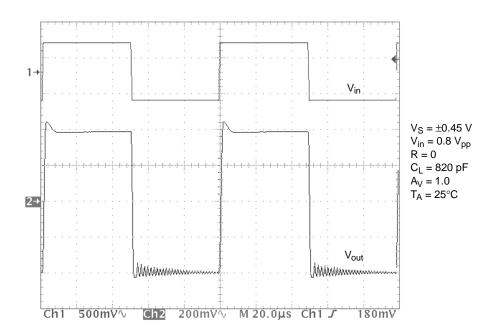


Figure 32. Small Signal Transient Response with Large Capacitive Load

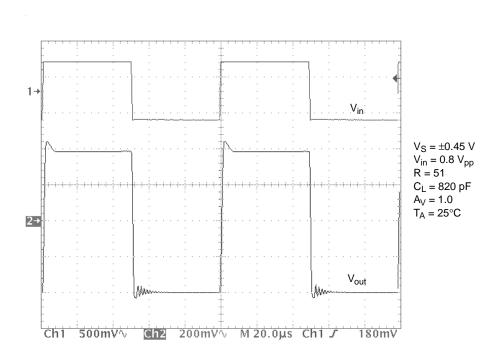


Figure 33. Small Signal Transient Response with Large Capacitive Load and Isolation Resistor

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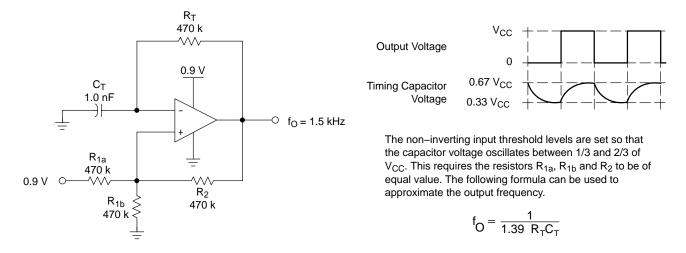


Figure 34. 0.9 V Square Wave Oscillator

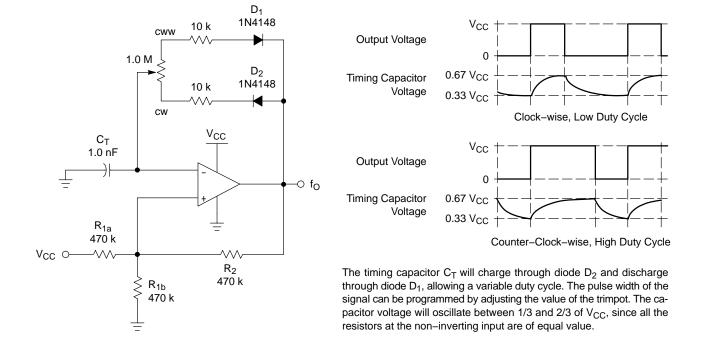


Figure 35. Variable Duty Cycle Pulse Generator

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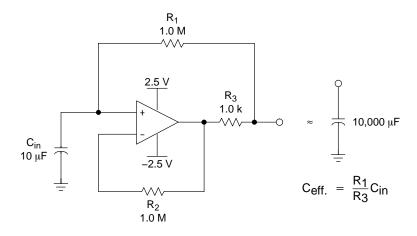


Figure 36. Positive Capacitance Multiplier

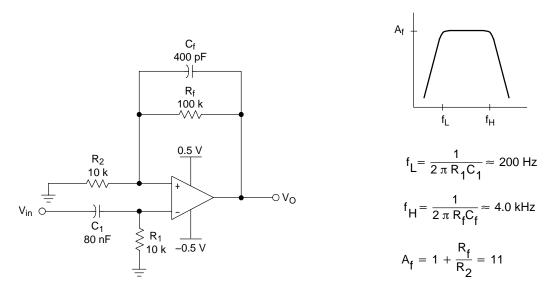


Figure 37. 1.0 V Voiceband Filter

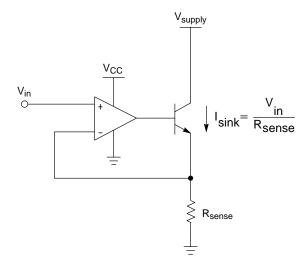
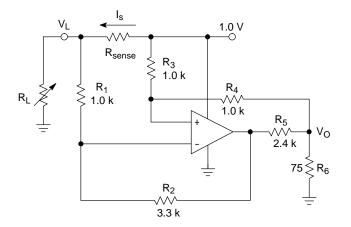


Figure 38. High Compliance Current Sink



Is	v <sub>o</sub>
435 mA	34.7 mV
212 mA	36.9 mV

For best performance, use low tolerance resistors.

Figure 39. High Side Current Sense

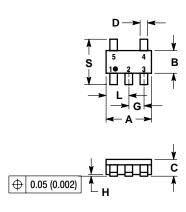
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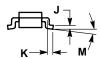
Device	Package	Shipping <sup>†</sup>
NCS2001SN1T1	SOT23-5	3000 / Tape & 7" Reel
NCS2001SN1T1G	SOT23-5 (Pb-Free)	3000 / Tape & 7" Reel
NCS2001SN2T1	SOT23-5	3000 / Tape & 7" Reel
NCS2001SQ1T1	SC70-5	3000 / Tape & 7" Reel
NCS2001SQ1T1G	SC70-5 (Pb-Free)	3000 / Tape & 7" Reel
NCS2001SQ2T1	SC70-5	3000 / Tape & 7" Reel
NCS2001SQ1T2	SC70-5	3000 / Tape & 7" Reel
NCS2001SQ2T2	SC70-5	3000 / Tape & 7" Reel

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

### **PACKAGE DIMENSIONS**

SOT23-5 N SUFFIX PLASTIC PACKAGE CASE 483-02 ISSUE C

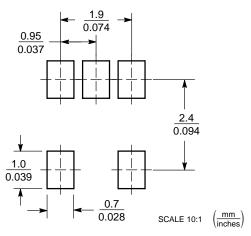




- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
  2. CONTROLLING DIMENSION: MILLIMETER.
  3. MAXIMUM LEAD THICKNESS INCLUDES LEAD FINISH THICKNESS. MINIMUM LEAD THICKNESS IS THE MINIMUM THICKNESS OF BASE MATERIAL.
  4. A AND B DIMENSIONS DO NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS.

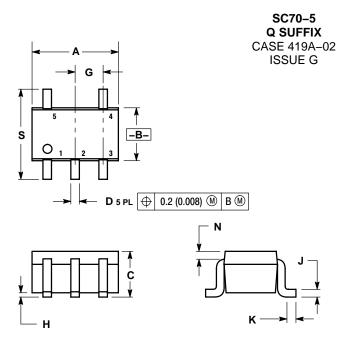
	MILLIMETERS		INC	HES
DIM	MIN	MAX	MIN	MAX
Α	2.90	3.10	0.1142	0.1220
В	1.30	1.70	0.0512	0.0669
С	0.90	1.10	0.0354	0.0433
D	0.25	0.50	0.0098	0.0197
G	0.85	1.05	0.0335	0.0413
Н	0.013	0.100	0.0005	0.0040
J	0.10	0.26	0.0040	0.0102
K	0.20	0.60	0.0079	0.0236
L	1.25	1.55	0.0493	0.0610
M	0	10	0	10
S	2.50	3.00	0.0985	0.1181

### **SOLDERING FOOTPRINT\***



\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

#### PACKAGE DIMENSIONS

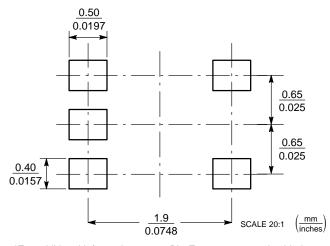


#### NOTES:

- 1. DIMENSIONING AND TOLERANCING
- PER ANSI Y14.5M, 1982.
  2. CONTROLLING DIMENSION: INCH.
- 419A-01 OBSOLETE. NEW STANDARD 419A-02.
- 4. DIMENSIONS A AND B DO NOT INCLUDE MOLD FLASH, PROTRUSIONS, OR GATE BURRS.

	INC	HES	MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.071	0.087	1.80	2.20
В	0.045	0.053	1.15	1.35
С	0.031	0.043	0.80	1.10
D	0.004	0.012	0.10	0.30
G	0.026	BSC	0.65	BSC
Н		0.004		0.10
J	0.004	0.010	0.10	0.25
K	0.004	0.012	0.10	0.30
N	0.008	REF	0.20 REF	
S	0.079	0.087	2 00	2 20

#### **SOLDERING FOOTPRINT\***



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