查询NTP35N15供应商

NTP35N15

Preferred Device

Power MOSFET 37 Amps, 150 Volts N-Channel TO-220

Features

- Source-to-Drain Diode Recovery Time Comparable to a Discrete Fast Recovery Diode
- Avalanche Energy Specified
- I_{DSS} and R_{DS(on)} Specified at Elevated Temperature

Typical Applications

- PWM Motor Controls
- Power Supplies
- Converters

MAXIMUM RATINGS (T_C = 25° C unless otherwise noted)

Rating	Symbol	Value	Unit	
Drain-to-Source Voltage	V _{DSS}	150	Vdc	
Drain-to-Source Voltage ($R_{GS} = 1.0 M\Omega$)	V _{DGR}	150	Vdc	
Gate–to–Source Voltage – Continuous – Non–Repetitive (t _p ≤ 10 ms)	V _{GS} V _{GSM}	±20 ±40	Vdc	
Drain Current – Continuous @ T _A 25°C – Continuous @ T _A 100°C – Pulsed (Note 1.)	I _D I _D I _{DM}	37 23 111	Adc	
Total Power Dissipation @ T _A = 25°C Derate above 25°C	P _D	178 1.43	W W/°C	
Operating and Storage Temperature Range	T _J , T _{stg}	-55 to +150	°C	
Single Drain-to-Source Avalanche Energy – Starting T _J = 25°C (V_{DD} = 100 Vdc, V_{GS} = 10 Vdc, I _L (pk) = 21.6 A, L = 3.0 mH, R _G = 25 Ω)	E _{AS}	700	mJ	
Thermal Resistance – Junction-to-Case – Junction-to-Ambient	R _{θJC} R _{θJA}	0.7 62.5	°C/W	
Maximum Lead Temperature for Soldering Purposes, 1/8" from case for 10 seconds	TL	260	°C	

1. Pulse Test: Pulse Width = 10 μ s, Duty Cycle = 2%.

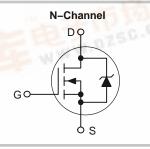


ON Semiconductor®

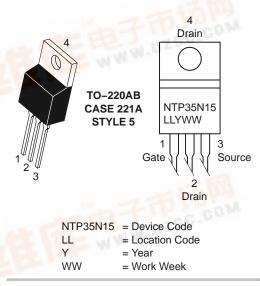
捷多邦,专业PCB打样工厂,24小时加急出货

http://onsemi.com

37 AMPERES 150 VOLTS 50 mΩ @ V_{GS} = 10 V



MARKING DIAGRAM & PIN ASSIGNMENT



ORDERING INFORMATION

Device	Package	Shipping
NTP35N15	TO-220AB	50 Units/Rail

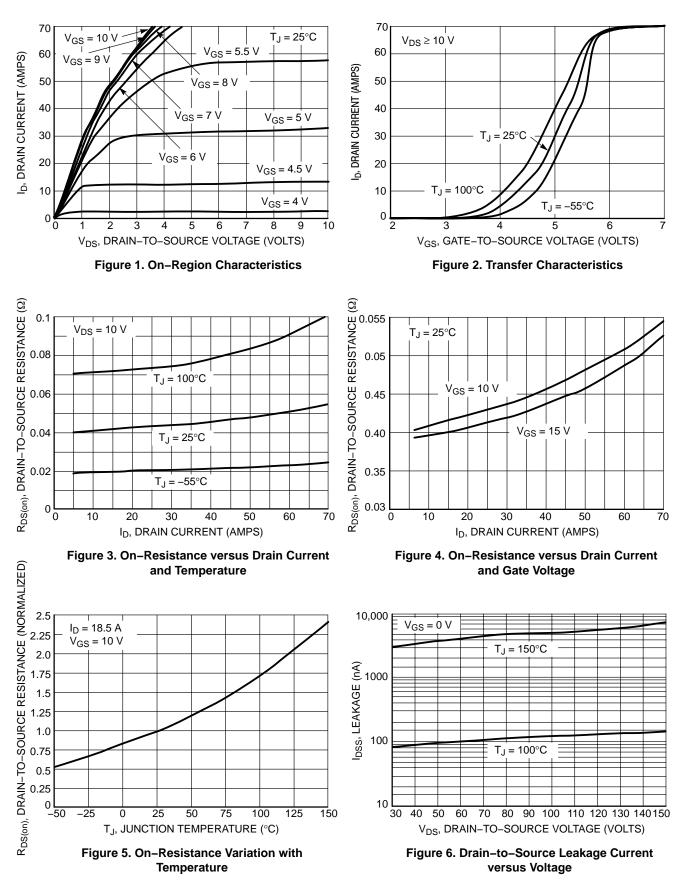
Preferred devices are recommended choices for future use and best overall value.



ELECTRICAL CHARACTERISTICS ($T_C = 25^{\circ}C$ unless otherwise noted)

Characteristic		Symbol	Min	Тур	Мах	Unit
OFF CHARACTERISTICS						
Drain-to-Source Breakdown Voltage ($V_{GS} = 0 Vdc, I_D = 250 \mu Adc$) Temperature Coefficient (Positive)		V _{(BR)DSS}	150 -	_ 240		Vdc mV/°C
Zero Gate Voltage Collector Current $(V_{GS} = 0 \text{ Vdc}, V_{DS} = 150 \text{ Vdc}, T_J = 25^{\circ}\text{C})$ $(V_{GS} = 0 \text{ Vdc}, V_{DS} = 150 \text{ Vdc}, T_J = 125^{\circ}\text{C})$		I _{DSS}			5.0 50	μAdc
Gate-Body Leakage Current (VG	$_{\rm S} = \pm 20 \rm V dc, V_{\rm DS} = 0)$	I _{GSS}	-	-	±100	nAdc
ON CHARACTERISTICS						
Gate Threshold Voltage $V_{DS} = V_{GS}$, I _D = 250 μ Adc) Temperature Coefficient (Negative)		V _{GS(th)}	2.0	2.9 -8.56	4.0	Vdc mV/°C
Static Drain-to-Source On-State Resistance (V_{GS} = 10 Vdc, I_D = 18.5 Adc) (V_{GS} = 10 Vdc, I_D = 18.5 Adc, T_J = 125°C)		R _{DS(on)}		0.042 -	0.050 0.120	Ω
Drain-to-Source On-Voltage $(V_{GS} = 10 \text{ Vdc}, I_D = 18.5 \text{ Adc})$		V _{DS(on)}	-	1.55	1.78	Vdc
Forward Transconductance (V_{DS} = 10 Vdc, I_D = 18.5 Adc)		9 FS	-	26	-	mhos
OYNAMIC CHARACTERISTICS						
Input Capacitance		C _{iss}	-	2275	3200	pF
Output Capacitance	$(V_{DS} = 25 \text{ Vdc}, V_{GS} = 0 \text{ Vdc}, f = 1.0 \text{ MHz})$	C _{oss}	-	450	650	
Reverse Transfer Capacitance		C _{rss}	-	90	175	
SWITCHING CHARACTERISTICS	(Notes 2. & 3.)					
Turn–On Delay Time		t _{d(on)}	-	20	35	ns
Rise Time	$(V_{DD} = 120 \text{ Vdc}, I_D = 37 \text{ Adc}, V_{GS} = 10 \text{ Vdc},$	t _r	-	125	225	
Turn-Off Delay Time	$R_G = 9.1 \Omega$)	t _{d(off)}	-	90	175	
Fall Time		t _f	-	120	210	
Gate Charge	(V _{DS} = 120 Vdc, I _D = 37 Adc, V _{GS} = 10 Vdc)	Q _{tot}	-	70	100	nC
		Q _{gs}	-	14	-	-
		Q _{gd}	-	32	-	
BODY-DRAIN DIODE RATINGS (I	Note 2.)					
Forward On–Voltage	$(I_{S} = 37 \text{ Adc}, V_{GS} = 0 \text{ Vdc})$ $(I_{S} = 37 \text{ Adc}, V_{GS} = 0 \text{ Vdc}, T_{J} = 125^{\circ}\text{C})$	V _{SD}	-	1.00 0.88	1.5 -	Vdc
Reverse Recovery Time		t _{rr}	-	170	_	ns
	(I _S = 37 Adc, V _{GS} = 0 Vdc, dI _S /dt = 100 A/μs)	ta	-	112	-	
		t _b	-	58	-	
Reverse Recovery Stored Charge		Q _{RR}	-	1.14	-	μC

Pulse Test: Pulse Width ≤ 300 μs, Duty Cycle ≤ 2%.
Switching characteristics are independent of operating junction temperature.



POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals (Δt) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain–gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ($I_{G(AV)}$) can be made from a rudimentary analysis of the drive circuit so that

 $t = Q/I_{G(AV)}$

During the rise and fall time interval when switching a resistive load, V_{GS} remains virtually constant at a level known as the plateau voltage, V_{SGP} . Therefore, rise and fall times may be approximated by the following:

 $t_r = Q_2 x R_G / (V_{GG} - V_{GSP})$ $t_f = Q_2 x R_G / V_{GSP}$

where

 V_{GG} = the gate drive voltage, which varies from zero to V_{GG}

 R_{G} = the gate drive resistance

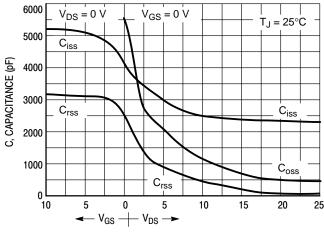
and Q_2 and V_{GSP} are read from the gate charge curve.

During the turn–on and turn–off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in an RC network. The equations are:

 $t_{d(on)} = R_G C_{iss} In [V_{GG}/(V_{GG} - V_{GSP})]$ $t_{d(off)} = R_G C_{iss} In (V_{GG}/V_{GSP})$ The capacitance (C_{iss}) is read from the capacitance curve at a voltage corresponding to the off–state condition when calculating $t_{d(on)}$ and is read at a voltage corresponding to the on–state when calculating $t_{d(off)}$.

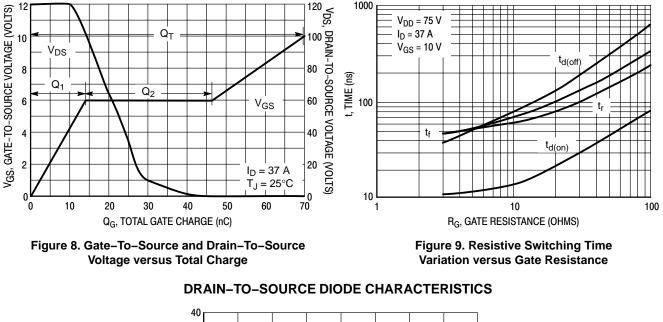
At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by Ldi/dt, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.



GATE-TO-SOURCE OR DRAIN-TO-SOURCE VOLTAGE (VOLTS)

Figure 7. Capacitance Variation



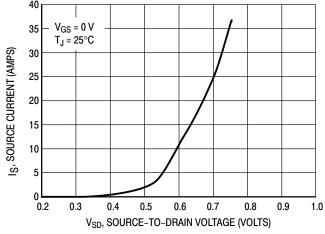


Figure 10. Diode Forward Voltage versus Current

SAFE OPERATING AREA

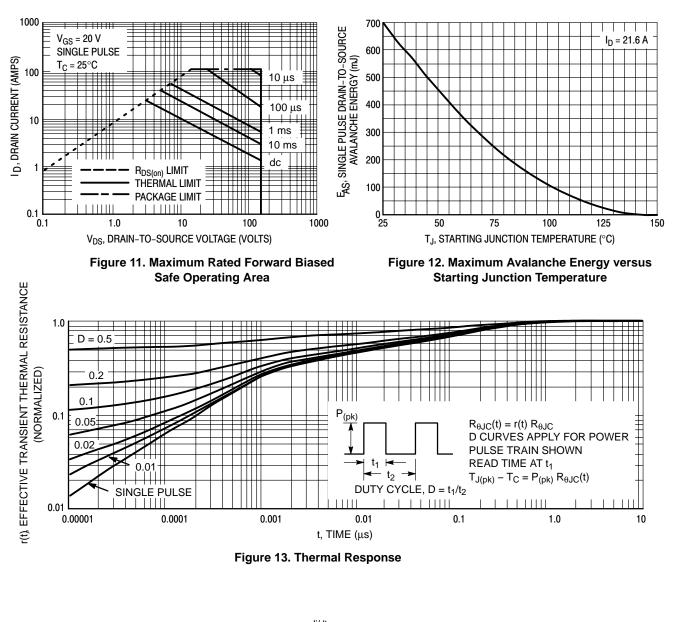
The Forward Biased Safe Operating Area curves define the maximum simultaneous drain–to–source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature (T_C) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance–General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current (I_{DM}) nor rated voltage (V_{DSS}) is exceeded and the transition time (t_p,t_f) do not exceed 10 µs. In addition the total power averaged over a complete switching cycle must not exceed (T_{J(MAX)} – T_C)/(R_{θJC}).

A Power MOSFET designated E–FET can be safely used in switching circuits with unclamped inductive loads. For reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non–linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E–FETs can withstand the stress of drain–to–source avalanche at currents up to rated pulsed current (I_{DM}), the energy rating is specified at rated continuous current (I_D), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 12). Maximum energy at currents below rated continuous I_D can safely be assumed to equal the values indicated.

SAFE OPERATING AREA



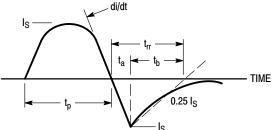
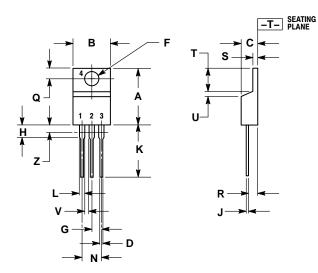


Figure 14. Diode Reverse Recovery Waveform

PACKAGE DIMENSIONS

TO-220 THREE-LEAD TO-220AB CASE 221A-09 **ISSUE AA**



- NOTES: 1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982. 2. CONTROLLING DIMENSION: INCH. 3. DIMENSION Z DEFINES A ZONE WHERE ALL BODY AND LEAD IRREGULARITIES ARE ALLOWED.

	INCHES		MILLIMETERS	
DIM	MIN	MAX	MIN	MAX
Α	0.570	0.620	14.48	15.75
В	0.380	0.405	9.66	10.28
С	0.160	0.190	4.07	4.82
D	0.025	0.035	0.64	0.88
F	0.142	0.147	3.61	3.73
G	0.095	0.105	2.42	2.66
Н	0.110	0.155	2.80	3.93
L	0.018	0.025	0.46	0.64
κ	0.500	0.562	12.70	14.27
L	0.045	0.060	1.15	1.52
Ν	0.190	0.210	4.83	5.33
Ø	0.100	0.120	2.54	3.04
R	0.080	0.110	2.04	2.79
S	0.045	0.055	1.15	1.39
Т	0.235	0.255	5.97	6.47
U	0.000	0.050	0.00	1.27
۷	0.045		1.15	
Ζ		0.080		2.04

STYLE 5: PIN 1. GATE 2. DRAIN

ON Semiconductor and are registered trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other application in which the BSCILLC product create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, and distributors harmless against all claims, costs, damages, and expenses, and reasonable atorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use personal and specific copyright laws and is not for resale in any manner.

PUBLICATION ORDERING INFORMATION

LITERATURE FULFILLMENT:

Literature Distribution Center for ON Semiconductor P.O. Box 5163, Denver, Colorado 80217 USA Phone: 303–675–2175 or 800–344–3860 Toll Free USA/Canada Fax: 303–675–2176 s00–344–3867 Toll Free USA/Canada Email: orderlit@onsemi.com

N. American Technical Support: 800–282–9855 Toll Free USA/Canada

Japan: ON Semiconductor, Japan Customer Focus Center 2–9–1 Kamimeguro, Meguro–ku, Tokyo, Japan 153–0051 Phone: 81–3–5773–3850 ON Semiconductor Website: http://onsemi.com

Order Literature: http://www.onsemi.com/litorder

For additional information, please contact your local Sales Representative.