

# 2-Phase Stepper Motor Unipolar Driver ICs

## ■Absolute Maximum Ratings

| Parameter                | Symbol    | Ratings                | Units |
|--------------------------|-----------|------------------------|-------|
| Motor supply voltage     | $V_{CC}$  | 46                     | V     |
| FET Drain-Source voltage | $V_{DSS}$ | 100                    | V     |
| Control supply voltage   | $V_S$     | 46                     | V     |
| TTL input voltage        | $V_{IN}$  | 7                      | V     |
| Reference voltage        | $V_{REF}$ | 2                      | V     |
| Output current           | $I_O$     | 1                      | A     |
| Power dissipation        | $P_D$     | 2.5 (Without Heatsink) | W     |
| Channel temperature      | $T_{ch}$  | +150                   | °C    |
| Storage temperature      | $T_{stg}$ | -40 to +150            | °C    |

## ■Electrical Characteristics

| Parameter                       | Symbol                 | Ratings                     |     |      | Units   |
|---------------------------------|------------------------|-----------------------------|-----|------|---------|
|                                 |                        | min                         | typ | max  |         |
| Control supply current          | $I_S$                  |                             | 5   | 7.5  | mA      |
|                                 | Condition              | $V_S=44V$                   |     |      |         |
| Control supply voltage          | $V_S$                  | 10                          | 24  | 44   | V       |
| FET Drain-Source voltage        | $V_{DSS}$              | 100                         |     |      | V       |
|                                 | Condition              | $V_S=44V, I_{DSS}=250\mu A$ |     |      |         |
| FET ON voltage                  | $V_{DS}$               |                             |     | 0.85 | V       |
|                                 | Condition              | $I_b=1A, V_S=14V$           |     |      |         |
| FET drain leakage current       | $I_{DSS}$              |                             |     | 4    | mA      |
|                                 | Condition              | $V_{DSS}=100V, V_S=44V$     |     |      |         |
| FET diode forward voltage       | $V_{SD}$               |                             |     | 1.2  | V       |
|                                 | Condition              | $I_b=1A$                    |     |      |         |
| TTL input current               | $I_{IH}$               |                             |     | 40   | $\mu A$ |
|                                 | Condition              | $V_{IH}=2.4V, V_S=44V$      |     |      |         |
|                                 | $I_{IL}$               |                             |     | -0.8 | mA      |
| Condition                       | $V_{IL}=0.4V, V_S=44V$ |                             |     |      |         |
| TTL input voltage (Active High) | $V_{IH}$               | 2                           |     |      | V       |
|                                 | Condition              | $I_b=1A$                    |     |      |         |
|                                 | $V_{IL}$               |                             |     | 0.8  |         |
| Condition                       | $V_{DSS}=100V$         |                             |     |      |         |
| TTL input voltage (Active Low)  | $V_{IH}$               | 2                           |     |      | V       |
|                                 | Condition              | $V_{DSS}=100V$              |     |      |         |
|                                 | $V_{IL}$               |                             |     | 0.8  |         |
| Condition                       | $I_b=1A$               |                             |     |      |         |
| Switching time                  | $T_r$                  |                             | 0.5 |      | $\mu s$ |
|                                 | Condition              | $V_S=24V, I_b=0.8A$         |     |      |         |
|                                 | $T_{stg}$              |                             | 0.7 |      |         |
|                                 | Condition              | $V_S=24V, I_b=0.8A$         |     |      |         |
|                                 | $T_f$                  |                             | 0.1 |      |         |
| Condition                       | $V_S=24V, I_b=0.8A$    |                             |     |      |         |

Internal Block Diagram

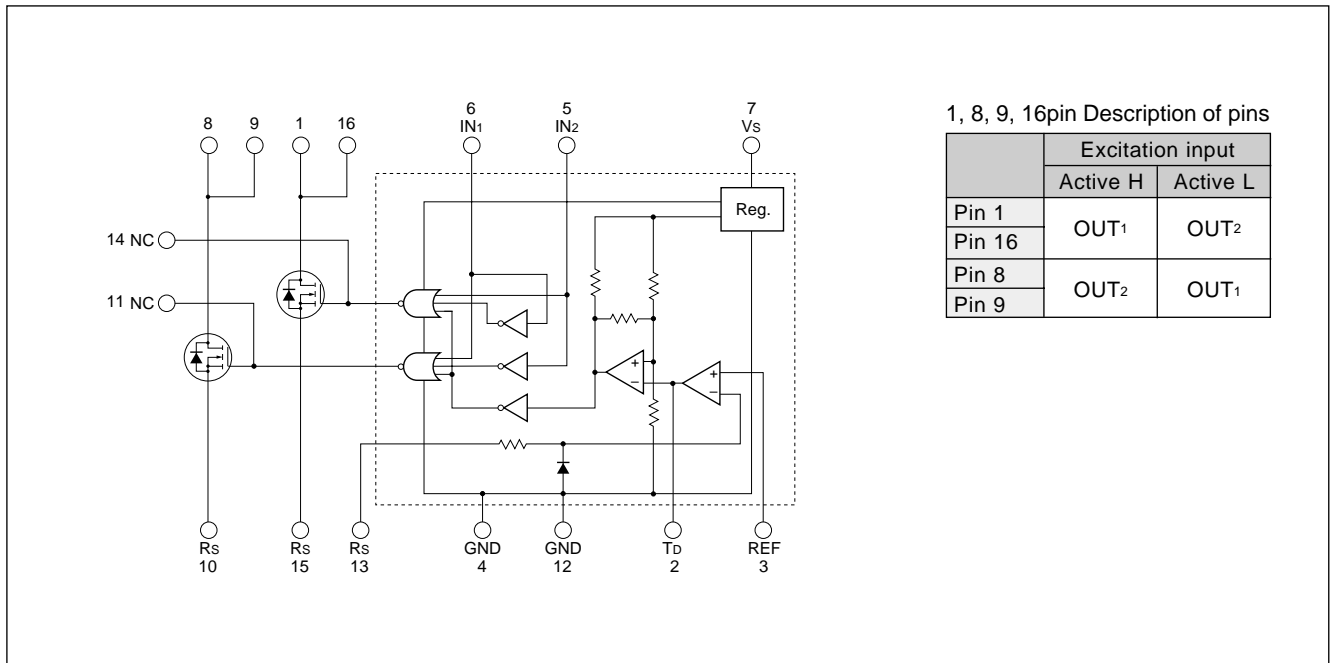
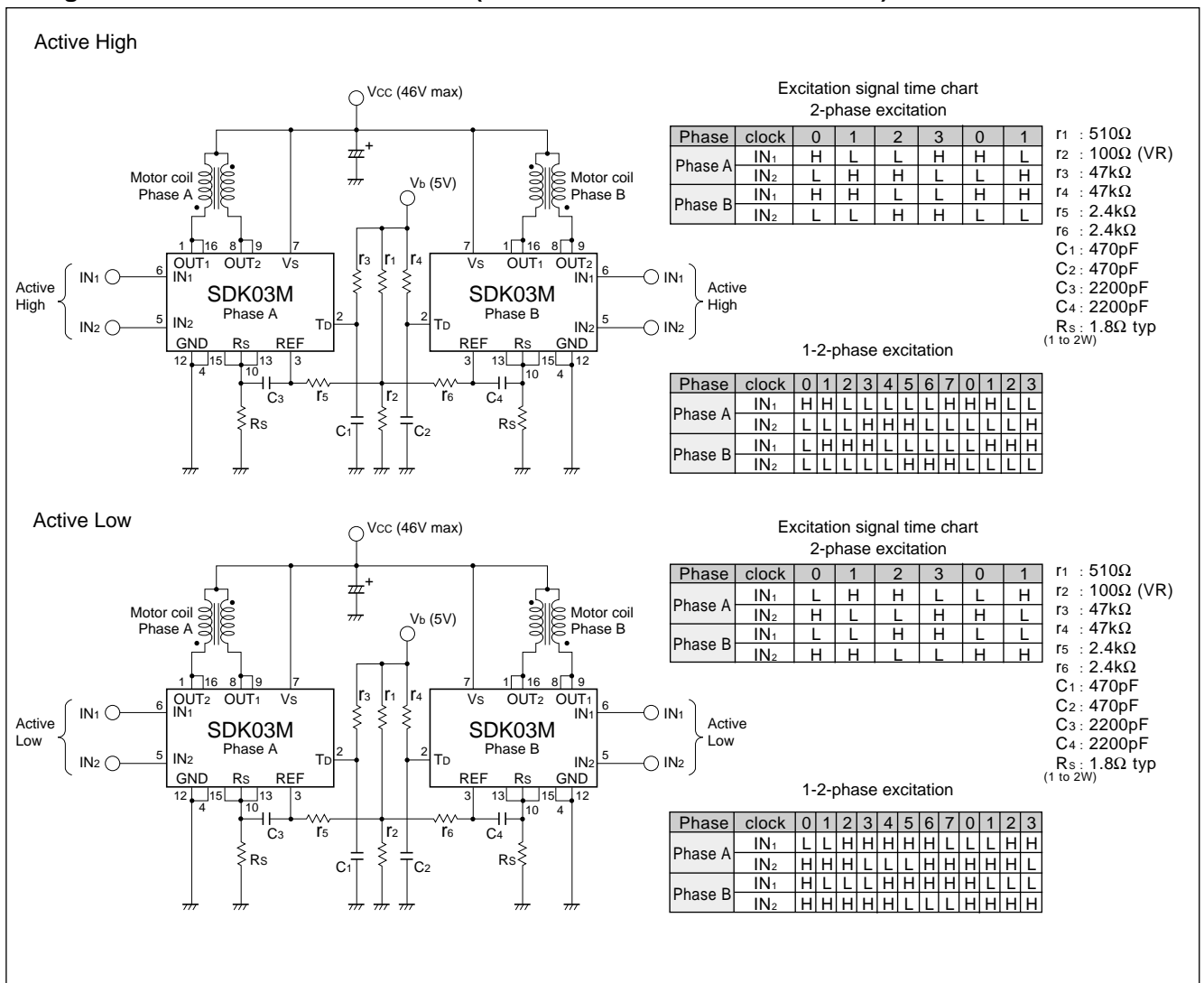
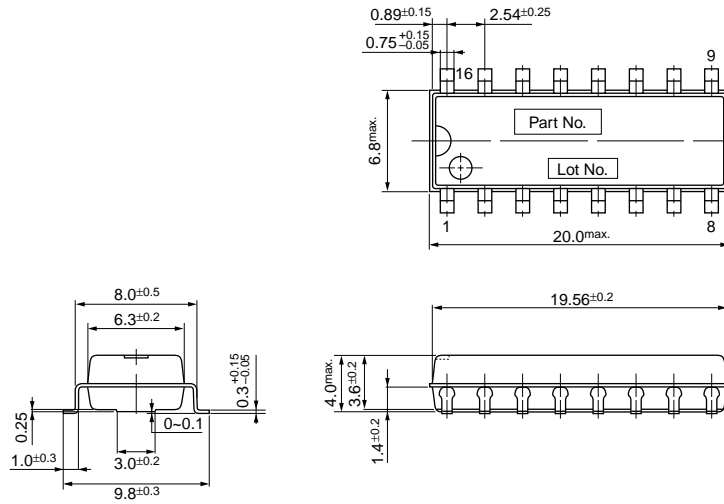


Diagram of Standard External Circuit (Recommended Circuit Constants)



External Dimensions

(Unit: mm)



## Application Notes

### ■Determining the Output Current

Fig. 1 shows the waveform of the output current (motor coil current). The method of determining the peak value of the output current ( $I_o$ ) based on this waveform is shown below.

(Parameters for determining the output current  $I_o$ )

$V_b$ : Reference supply voltage

$r_1, r_2$ : Voltage-divider resistors for the reference supply voltage

$R_s$ : Current sense resistor

(1) Normal rotation mode

$I_o$  is determined as follows when current flows at the maximum level during motor rotation. (See Fig.2.)

$$I_o \cong \frac{r_2}{r_1+r_2} \cdot \frac{V_b}{R_s} \dots\dots\dots (1)$$

(2) Power down mode

The circuit in Fig.3 ( $r_x$  and  $T_r$ ) is added in order to decrease the coil current.  $I_o$  is then determined as follows.

$$I_{OPD} \cong \frac{1}{1 + \frac{r_1(r_2+r_x)}{r_2 \cdot r_x}} \cdot \frac{V_b}{R_s} \dots\dots\dots (2)$$

Equation (2) can be modified to obtain equation to determine  $r_x$ .

$$r_x = \frac{1}{\frac{1}{r_1} \left( \frac{V_b}{R_s \cdot I_{OPD}} - 1 \right) - \frac{1}{r_2}}$$

Fig. 4 and 5 show the graphs of equations (1) and (2) respectively.

Fig. 1 Waveform of coil current (Phase A excitation ON)

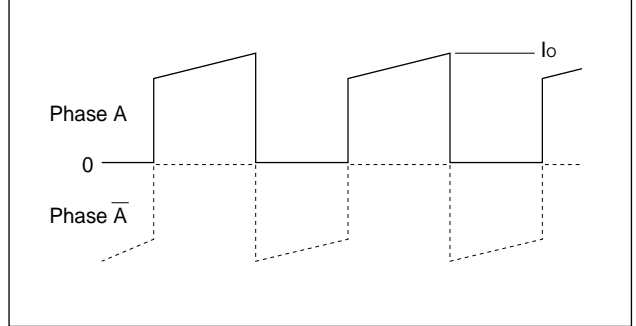


Fig. 2 Normal mode

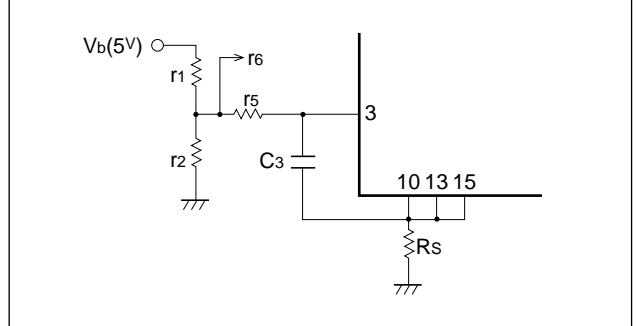


Fig. 3 Power down mode

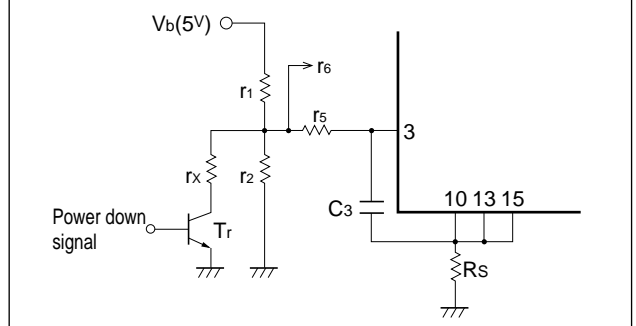


Fig. 4 Output current  $I_o$  vs. Current sense resistor  $R_s$

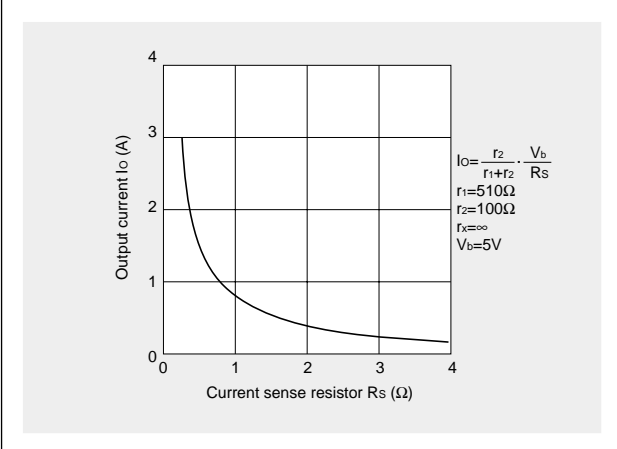
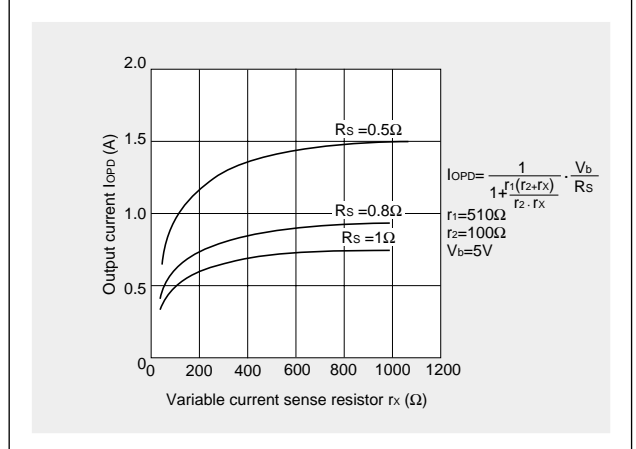


Fig. 5 Output current  $I_{OPD}$  vs. Variable current sense resistor  $r_x$



**(NOTE)**

Ringing noise is produced in the current sense resistor  $R_s$  when the MOSFET is switched ON and OFF by chopping. This noise is also generated in feedback signals from  $R_s$  which may therefore cause the comparator to malfunction. To prevent chopping malfunctions,  $r_5(r_6)$  and  $C_3(C_4)$  are added to act as a noise filter.

However, when the values of these constants are increased, the response from  $R_s$  to the comparator becomes slow. Hence the value of the output current  $I_o$  is somewhat higher than the calculated value.

**Determining the chopper frequency**

Determining T<sub>OFF</sub>

SDK03M is self-excited choppers. The chopping OFF time T<sub>OFF</sub> is fixed by r<sub>3</sub>/C<sub>1</sub> and r<sub>4</sub>/C<sub>2</sub> connected to terminal T<sub>d</sub>.

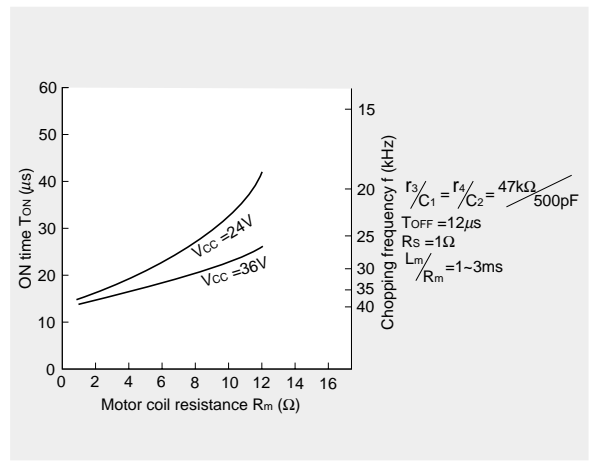
T<sub>OFF</sub> can be calculated using the following formula:

$$T_{OFF} \approx -r_3 \cdot C_1 \ln \left(1 - \frac{2}{V_b}\right) = -r_4 \cdot C_2 \ln \left(1 - \frac{2}{V_b}\right)$$

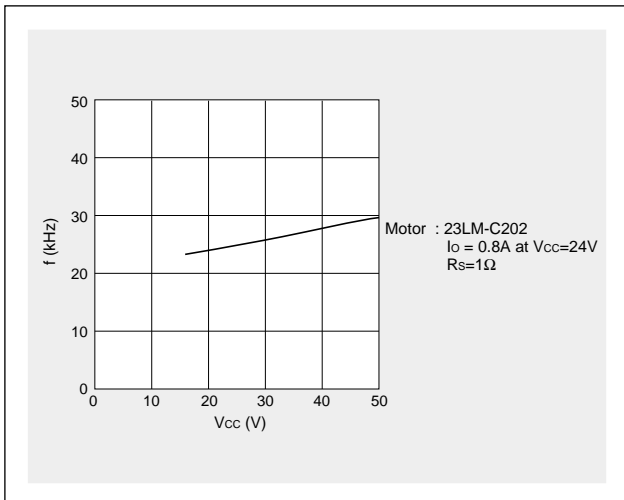
The circuit constants and the T<sub>OFF</sub> value shown below are recommended.

T<sub>OFF</sub> = 12μs at r<sub>3</sub>=47kΩ, C<sub>1</sub>=500pF, V<sub>b</sub>=5V

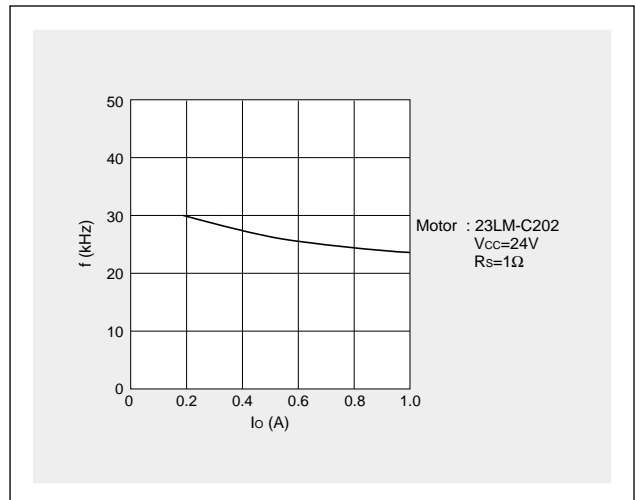
**Fig. 6 Chopper frequency vs. Motor coil resistance**



**Chopper frequency vs. Supply voltage**



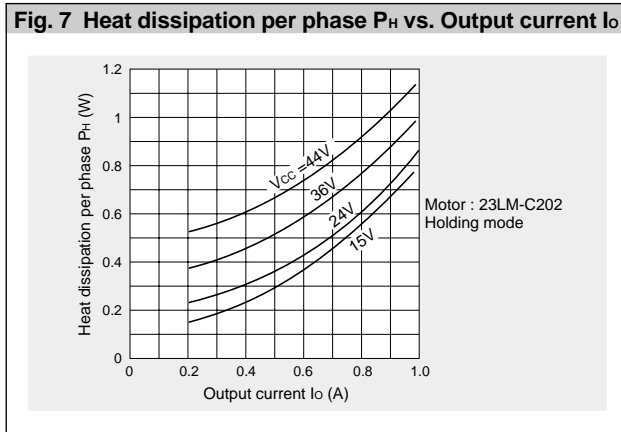
**Chopper frequency vs. Output current**



**Thermal Design**

An outline of the method for computing heat dissipation is shown below.

- (1) Obtain the value of  $P_H$  that corresponds to the motor coil current  $I_o$  from Fig. 7 "Heat dissipation per phase  $P_H$  vs. Output current  $I_o$ ."

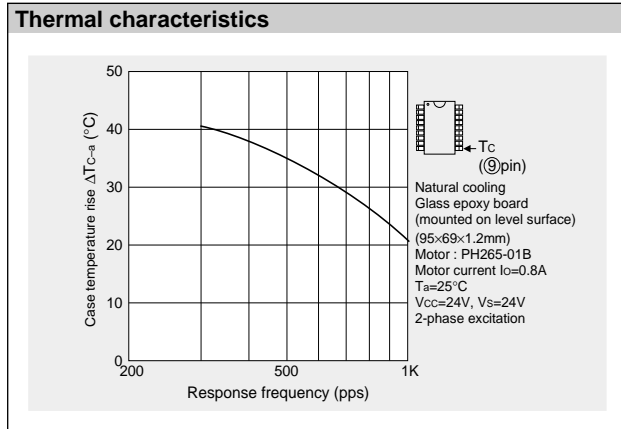
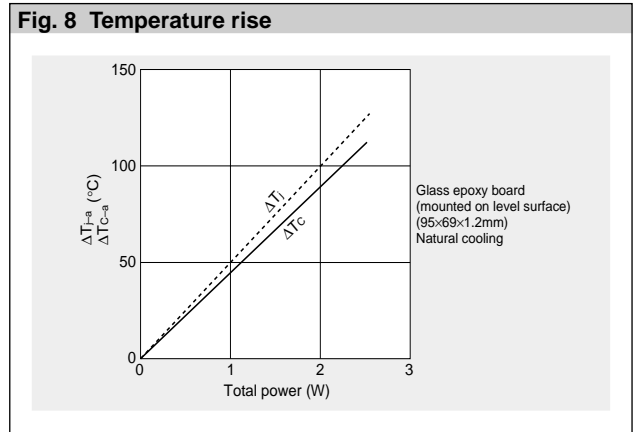


- (2) The power dissipation  $P_{diss}$  is obtained using the following formula.

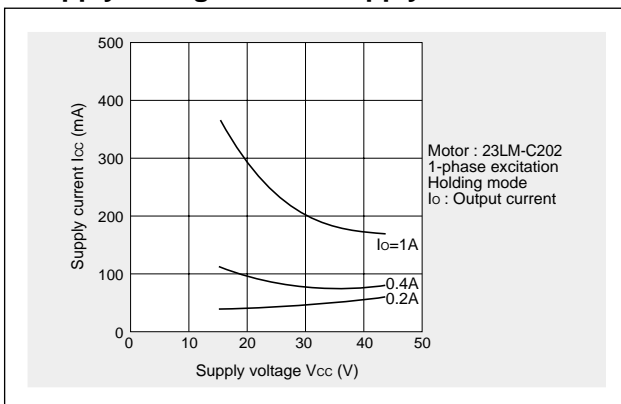
2-phase excitation:  $P_{diss} \cong P_H + 0.0075 \times V_s$  (W)

1-2 phase excitation:  $P_{diss} \cong \frac{3}{4} P_H + 0.0075 \times V_s$  (W)

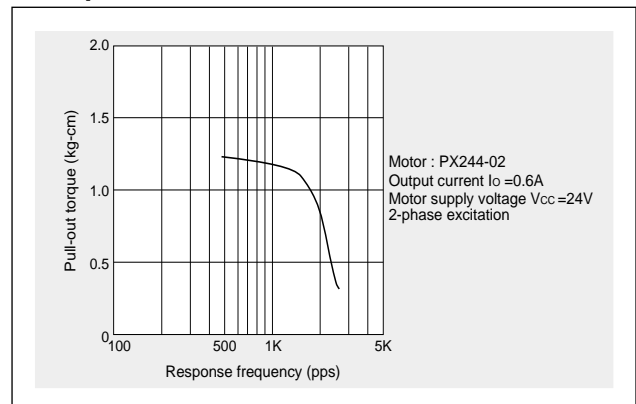
- (3) Obtain the temperature rise that corresponds to the calculated value of  $P_{diss}$  from Fig. 8 "Temperature rise."



**Supply Voltage Vcc vs. Supply Current Icc**



**Torque Characteristics**



**Note**

The excitation input signals of the SDK03M can be used as either Active High or Active Low. Note, However, that the corresponding output (OUT) changes depending on the input (IN).

**Active High**

| Input                  | Corresponding output        |
|------------------------|-----------------------------|
| IN <sub>1</sub> (pin6) | OUT <sub>1</sub> (pin1, 16) |
| IN <sub>2</sub> (pin5) | OUT <sub>2</sub> (pin8, 9)  |

**Active Low**

| Input                  | Corresponding output        |
|------------------------|-----------------------------|
| IN <sub>1</sub> (pin6) | OUT <sub>1</sub> (pin8, 9)  |
| IN <sub>2</sub> (pin5) | OUT <sub>2</sub> (pin1, 16) |