

# High Common-Mode Rejection Differential Line Receiver

SSM2141

#### **FEATURES**

**High Common-Mode Rejection** 

DC: 100 dB typ 60 Hz: 100 dB typ 20 kHz: 70 dB typ 40 kHz: 62 dB typ Low Distortion: 0.001% typ

Fast Slew Rate: 9.5 V/µs typ Wide Bandwidth: 3 MHz typ

Low Cost

Complements SSM2142 Differential Line Driver

APPLICATIONS Line Receivers Summing Amplifiers Buffer Amplifiers-Drives 600  $\Omega$  Load

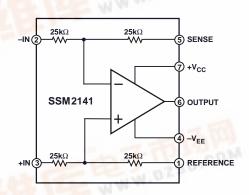
## GENERAL DESCRIPTION

The SSM2141 is an integrated differential amplifier intended to receive balanced line inputs in audio applications requiring a high level of noise immunity and optimum common-mode rejection. The SSM2141 typically achieves 100 dB of common-mode rejection (CMR), whereas implementing an op amp with four off-the-shelf precision resistors will typically achieve only 40 dB of CMR—inadequate for high-performance audio.

The SSM2141 achieves low distortion performance by maintaining a large slew rate of 9.5 V/µs and high open-loop gain. Distortion is less than 0.002% over the full audio bandwidth. The SSM2141 complements the SSM2142 balanced line driver. Together, these devices comprise a fully integrated solution for equivalent transformer balancing of audio signals without the problems of distortion, EMI fields, and high cost.

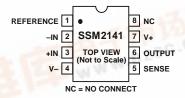
Additional applications for the SSM2141 include summing signals, differential preamplifiers, and 600  $\Omega$  low distortion buffer amplifiers. For similar performance with G = 1/2, see SSM2143.

### **FUNCTIONAL BLOCK DIAGRAM**



### PIN CONNECTIONS

8-Pin Plastic Mini-DIP (P Suffix) Narrow Body SO (S Suffix)



## SSM2141-SPECIFICATIONS

## **ELECTRICAL CHARACTERISTICS** (@ $V_S = \pm 18 \text{ V}$ , $T_A = +25 ^{\circ}\text{C}$ , unless otherwise noted)

			SSM2141			
Parameter	Symbol	Conditions	Min	Тур	Max	Units
OFFSET VOLTAGE	Vos	$V_{CM} = 0 V$	-1000	25	1000	μV
GAIN ERROR		No Load, $V_{IN} = \pm 10 \text{ V}$ , $R_S = 0 \Omega$		0.001	0.01	%
INPUT VOLTAGE RANGE	IVR	(Note 1)	±10			V
COMMON-MODE REJECTION	CMR	$V_{CM} = \pm 10 \text{ V}$	80	100		dB
POWER SUPPLY REJECTION RATIO	PSRR	$V_S = \pm 6 \text{ V to } \pm 18 \text{ V}$		0.7	15	μV/V
OUTPUT SWING	Vo	$R_L = 2 \text{ k}\Omega$	±13	±14.7		V
SHORT-CIRCUIT CURRENT LIMIT	$I_{SC}$	Output Shorted to Ground	+45/-15			mA
SMALL-SIGNAL BANDWIDTH (-3 dB)	BW	$R_L = 2 k\Omega$		3		MHz
SLEW RATE	SR	$R_L = 2 k\Omega$	6	9.5		V/µs
TOTAL HARMONIC DISTORTION	THD	$R_L = 100 \text{ k}\Omega$ $R_L = 600 \Omega$		0.001 0.01		%
CAPACITIVE LOAD DRIVE CAPABILITY	C <sub>L</sub>	No Oscillation		300		pF
SUPPLY CURRENT	I <sub>SY</sub>	No Load		2.5	3.5	mA

NOTES

Specifications subject to change without notice

## ELECTRICAL CHARACTERISTICS (@ $V_S = \pm 18 \text{ V}, -40 ^{\circ}\text{C} \le T_A \le +85 ^{\circ}\text{C}$ )

Parameter	Symbol	Conditions	Min	Тур	Max	Units
OFFSET VOLTAGE	V <sub>OS</sub>	$V_{CM} = 0 V$	-2500	200	2500	μV
GAIN ERROR		No Load, $V_{IN} = \pm 10 \text{ V}$ , $R_S = 0 \Omega$		0.002	0.02	%
INPUT VOLTAGE RANGE	IVR	(Note 1)	±10			V
COMMON-MODE REJECTION	CMR	$V_{CM} = \pm 10 \text{ V}$	75	90		dB
POWER SUPPLY REJECTION RATIO	PSRR	$V_S = \pm 6 \text{ V to } \pm 18 \text{ V}$		1.0	20	μV/V
OUTPUT SWING	Vo	$R_L = 2 k\Omega$	±13	±14.7		V
SLEW RATE	SR	$R_L = 2 \text{ k}\Omega$		9.5		V/µs
SUPPLY CURRENT	$I_{SY}$	No Load		2.6	4.0	mA

NOTES

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<sup>&</sup>lt;sup>1</sup>Input Voltage Range Guaranteed by CMR test.

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## SSM2141

## ABSOLUTE MAXIMUM RATINGS<sup>1</sup>

Supply Voltage	±18 V
Input Voltage <sup>1</sup> S	
Output Short-Circuit Duration	
Storage Temperature Range	
P Package65	°C to +150°C
Lead Temperature (Soldering, 60 sec)	+300°C
Junction Temperature	+150°C
Operating Temperature Range4	$0^{\circ}$ C to +85 $^{\circ}$ C

Package Type	$\theta_{\mathrm{JA}}^{2}$	$\theta_{JC}$	Units	
8-Pin Plastic DIP (P)	103	43	°C/W	

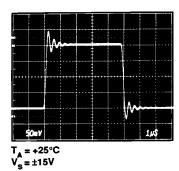
#### NOTES

 $^1\mbox{For}$  supply voltages less than  $\pm 18$  V, the absolute maximum input voltage is equal to the supply voltage.

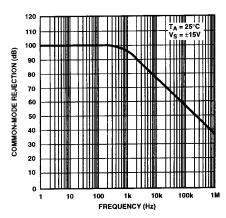
### **ORDERING GUIDE**

Model	Operating Temperature	Package	Package	
	Range	Description	Option	
SSM2141P SSM2141S	, , ,		N-8 SO-8	

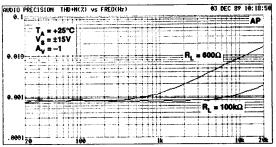
## **Typical Performance Characteristics**



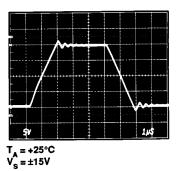
Small Signal Transient Response



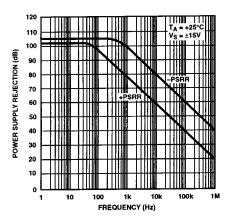
Common-Mode Rejection vs. Frequency



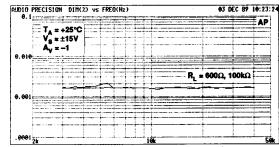
79 100 IR III



Large Signal Transient Response

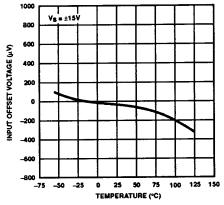


Power Supply Rejection vs. Frequency

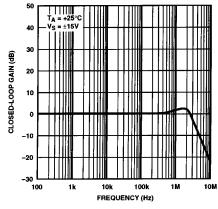


 $<sup>^2\</sup>theta_{JA}$  is specified for worst case mounting conditions, i.e.,  $\theta_{JA}$  is specified for device in socket for P-DIP package.

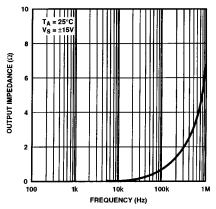
## SSM2141-Typical Performance Characteristics



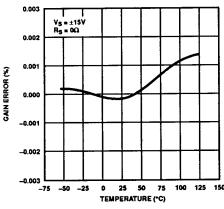
Input Offset Voltage vs. Temperature



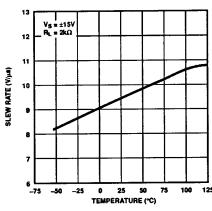
Closed-Loop Gain vs. Frequency



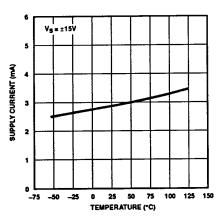
Closed-Loop Output Impedance vs. Frequency



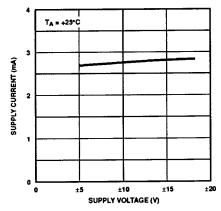
Gain Error vs. Temperature



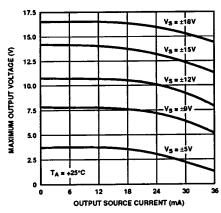
Slew Rate vs. Temperature



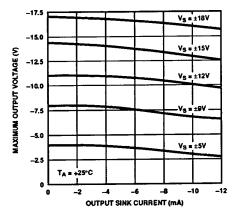
Supply Current vs. Temperature



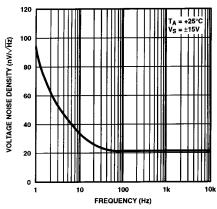
Supply Current vs. Supply Voltage



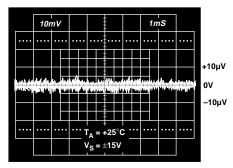
Maximum Output Voltage vs. Output Current (Source)



Maximum Output Voltage vs. Output Current (Sink)

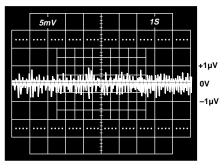


Voltage Noise Density vs. Frequency



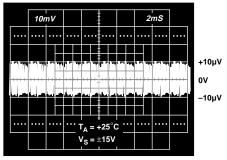
NOTE: EXTERNAL AMPLIFIER GAIN = 1000; THEREFORE, VERTICAL SCALE =  $10\mu V/DIV$ .

Voltage Noise from 0 kHz to 1 kHz



0.1 TO 10Hz PEAK-TO-PEAK NOISE

Low Frequency Voltage Noise

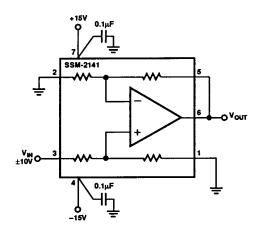


NOTE: EXTERNAL AMPLIFIER GAIN = 1000; THEREFORE, VERTICAL SCALE = 10µV/DIV.

Voltage Noise from 0 kHz to 10 kHz

### APPLICATIONS INFORMATION

The SSM2141 represents a versatile analog building block. In order to capitalize on fast settling time, high slew rate, and high CMR, proper decoupling and grounding techniques must be employed. For decoupling, place 0.1  $\mu F$  capacitor located within close proximity from each supply pin to ground.



Slew Rate Test Circuit

## SSM2141

### MAINTAINING COMMON-MODE REJECTION

In order to achieve the full common-mode rejection capability of the SSM2141, the source impedance must be carefully controlled. Slight imbalances of the source resistance will result in a degradation of DC CMR—even a 5  $\Omega$  imbalance will degrade CMR by 20 dB. Also, the matching of the reactive source impedance must be matched in order to preserve the CMRR over frequency.

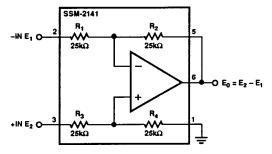


Figure 1. Precision Difference Amplifier. Rejects

Common-Mode Signal =  $\frac{[E_1 + E_2]}{2}$  by 100 dB

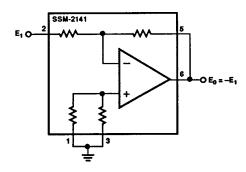


Figure 2. Precision Unity Gain Inverting Amplifier

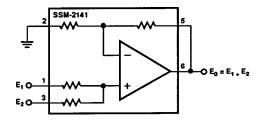


Figure 3. Precision Summing Amplifier

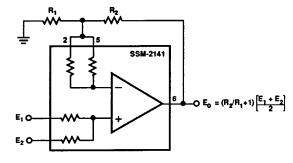


Figure 4. Precision Summing Amplifier with Gain

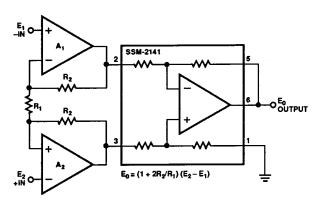
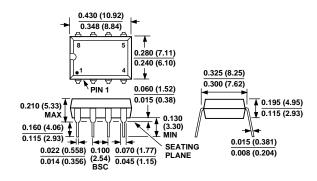


Figure 5. Suitable Instrumentation Amplifier Requirements can be Addressed by Using an Input Stage Consisting of  $A_1$ ,  $A_2$ ,  $R_1$  and  $R_2$ 

#### **OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

### 8-Pin Epoxy Mini-DIP P-Suffix (N-8)



### 8-Pin Narrow Body SO S-Suffix (SO-8)

