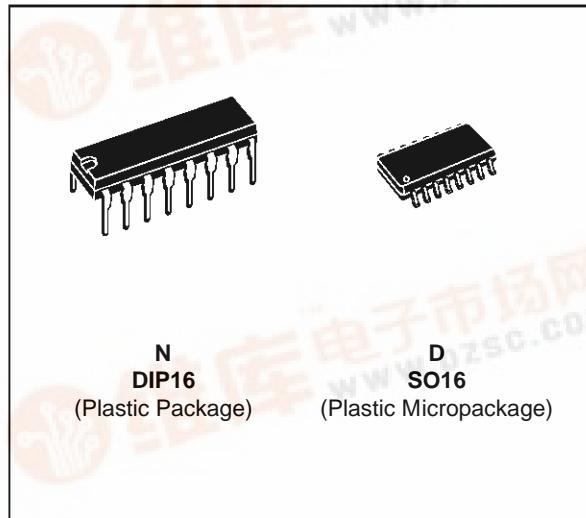




TD310

## TRIPLE IGBT/MOS DRIVER

- THREE POWER IGBT/MOS AND PULSE TRANSFORMER DRIVERS
- CURRENT SENSE COMPARATOR WITH 1ms INHIBITION TIME FUNCTION
- INSTANTANEOUS SIGNAL TRANSMISSION
- 0.6 Amp PER CHANNEL PEAK OUTPUT CURRENT CAPABILITY
- LOW OUTPUT IMPEDANCE TYP :  $7\Omega$  at 200mA
- CMOS/LSTTL COMPATIBLE INVERTING INPUT WITH HYSTERESIS
- 4V TO 16V SINGLE SUPPLY OPERATION
- CURRENT AMPLIFIER
- LOW BIAS CURRENT TYP : 1.5mA
- ADJUSTABLE UNDervoltage LOCKOUT LEVEL
- STAND-BY MODE
- DURING POWER UP NO RANDOM OUTPUT STATE
- ENHANCED LATCH-UP IMMUNITY
- CHANNEL PARALLELING CAPABILITY



## ORDER CODES

Part Number	Temperature Range	Package	
		N	D
TD310I	-40°C, +125°C	•	•

## DESCRIPTION

The TD310 is designed to drive one, two or three Power IGBT/MOS and has driving capability for pulse transformer. So it is perfectly suited to interface control IC with Power Switches in low side or half-bridge configuration.

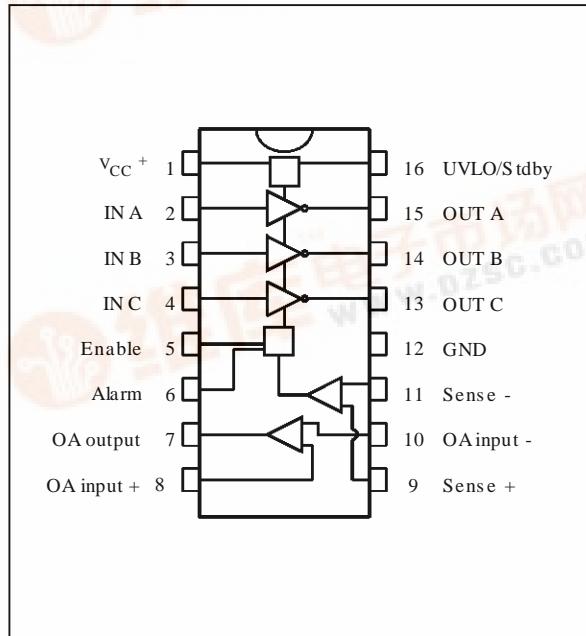
The typical application shown figure 1 implements the TD310 in a pulse controlled half-bridge drive. Positive and negative pulses are applied to the pulse transformer to charge and discharge the IGBT/MOS gate capacitance. More sophisticated secondary circuits provide low impedance gate drive and short-circuit protection as shown in application note n° AN461.

On Figure 2, TD310 is implemented as a low side driver in a typical 3 phase motor drive.

Figure 3 presents a general purpose low side gate drive.

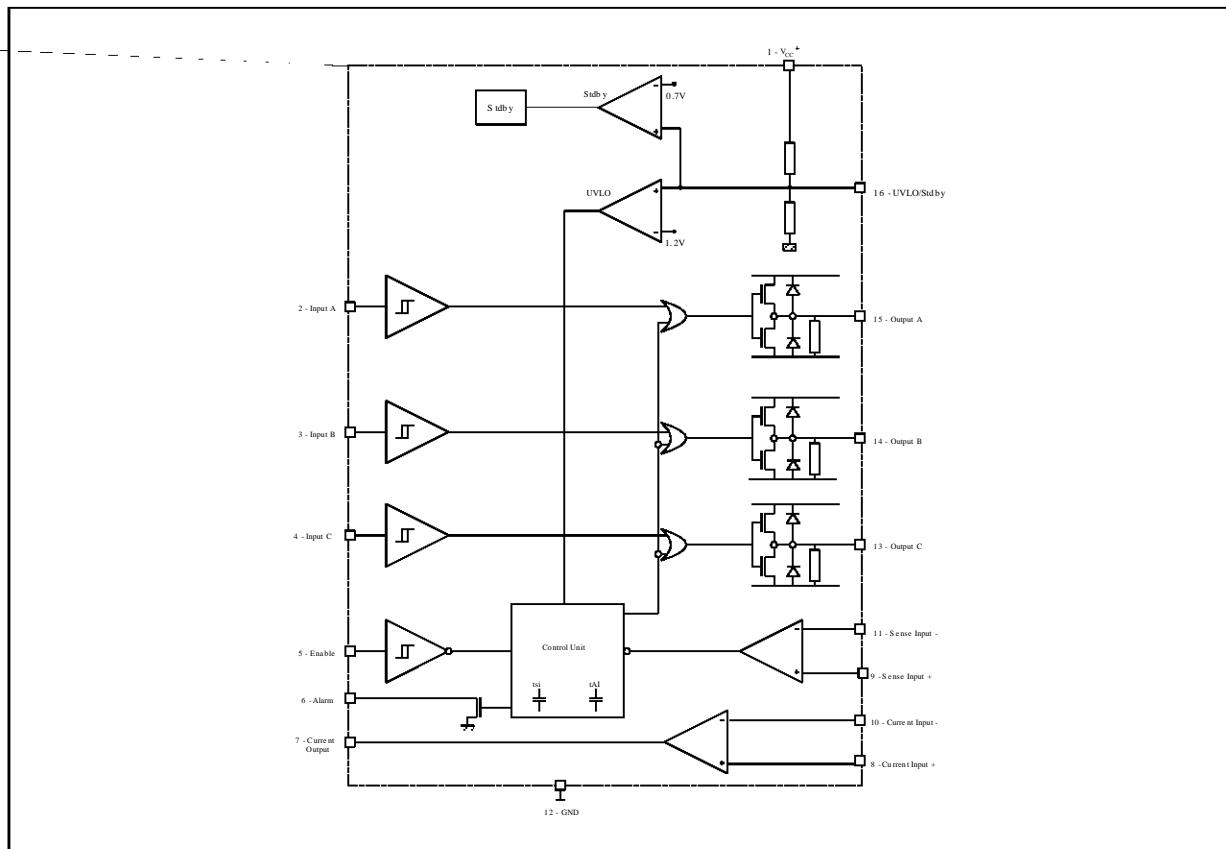
In both case, the current amplifier provides interfacing between a sense resistor and an A/D converter.

## PIN CONNECTIONS



# TD310

## BLOCK DIAGRAM



## ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
$V_{CC}$	Supply Voltage	18	V
$V_i$	Input Voltage	0 to $V_{CC}$	V
$V_{is}$	Sense Input Voltage	-0.3 to $V_{CC}$	V
$T_j$	Junction Temperature	-40 to 150	°C
$T_{amb}$	Ambient Temperature	-40 to 125	°C

## OPERATING CONDITIONS

Symbol	Parameter	Value	Unit
$V_{CC}$	Supply Voltage	4 to 16	V

## INSTRUCTIONS FOR USE

- 1 - The TD310 supply voltage must be decoupled with a  $1\mu F$  min. capacitor.
- 2 - If the application involving TD310 requires maximum output current capability, this current must be pulsed : pulse width  $1\mu sec$ , duty cycle 1% at  $T_{amb}$ .

**ELECTRICAL CHARACTERISTICS**V<sub>CC</sub> = 15V, T<sub>amb</sub> = 25°C (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I <sub>CC</sub>	Supply Current with Inputs in High State		1.5	2	mA
<b>LOGIC INPUT (all inputs)</b>					
V <sub>IH</sub>	High Input Voltage	2			V
V <sub>IL</sub>	Low Input Voltage			0.8	V
I <sub>IH</sub>	High Input Current		10		pA
I <sub>IL</sub>	Low Input Current		10		pA
t <sub>dH,t<sub>eH</sub></sub> t <sub>dL,t<sub>eL</sub></sub>	Propagation Delay (10% input to 10% output) Output Rise Output Fall T <sub>min</sub> ≤ T <sub>amb</sub> ≤ T <sub>max</sub> .		200 60	400 400	ns
t <sub>ii</sub>	Input Inhibiting Time		100		ns
t <sub>dd</sub>	Differential Delay Time Between Channels		20		ns
<b>OUTPUT DRIVERS</b>					
V <sub>sod</sub>	Sourcing Drop Voltage (A/B/C outputs) I <sub>source</sub> = 200mA			3	V
V <sub>sid</sub>	Sinking Drop Voltage (A/B/C outputs) I <sub>sink</sub> = 200mA			5	V
V <sub>dem</sub>	Demagnetising Drop Voltage (A/B/C outputs) I <sub>demag.</sub> = 100mA			2	V
R <sub>opd</sub>	Output Pull Down Resistor		47		kΩ
<b>ALARM OUTPUT</b>					
I <sub>s</sub>	Low Level Sinking Current V <sub>O</sub> = 0.8V	5	35		mA
I <sub>sh</sub>	High Level Sinking Current			1	μA
t <sub>A</sub>	Alarm Output : Delay Time to Alarm Fall if Sense Input Triggered			500	ns
<b>SENSE INPUT</b>					
V <sub>ios</sub>	Input Offset Voltage			20	mV
t <sub>AI</sub>	Inhibition Time if Sense Input Triggered		1		ms
t <sub>s</sub>	Delay Time to Output Fall if Sense Input Triggered All outputs inhibited			600	ns
t <sub>si</sub>	Inhibition Time of Sense Input		300		ns
V <sub>shys</sub>	Sense Hysteresis		40		mV
<b>OPERATIONAL AMPLIFIER</b>					
V <sub>icm</sub>	Common Mode Input Voltage Range	0 to V <sub>CC</sub> <sup>+</sup> - 1.5			V
V <sub>io</sub>	Input Offset Voltage			10	mV
GBP	Gain Bandwidth Product		1		MHz
A <sub>vd</sub>	Open Loop Gain	60			dB
SR	Slew Rate at Unity Gain R <sub>L</sub> = 100kΩ, C <sub>L</sub> = 100pF, V <sub>i</sub> = 3 to 7V		0.6		V/μs
<b>STAND-BY</b>					
V <sub>stdby</sub>	Standby Mode Threshold Voltage	0.3		1.1	V
I <sub>stdby</sub>	Standby Mode Supply Current		30		μA
<b>UNDER VOLTAGE LOCKOUT</b>					
I <sub>adj</sub>	Under Voltage Level Adjust Current		1		μA/V
V <sub>st1</sub>	Internal Stop Threshold (without external adjustment)	10.7		13.3	V
V <sub>hys</sub>	Threshold Hysteresis		0.8		V

## TD310

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### UVLO/stdby pin functionning modes

Due to the wide supply voltage range of the TD310, the UVLO function (Under Voltage Lock Out) is externally adjustable by a resistor bridge.

The bridge rate can be calculated in relation with the expected UVLO protection level as follows :

$$V_{UVLO} \times \frac{R1}{R1+R2} = 1.2V \text{ (where R1 is the lower resistor of the bridge)}$$

The internal resistor sets the default UVLO value to 12V (\*) and might influence the external bridge rate if the values of the external resistors are too high.

The standby threshold value depends of the UVLO value as follows :

$$V_{stdby} = 0.7/1.2 V_{UVLO}$$

Both UVLO and stdby functions can be inhibited by connecting the UVLO/stdby pin to  $V_{CC}^+$  via a pull up resistor (ex 150kΩ).

The following table summarizes the functions of the TD310 :

	Pin	16	9/11	5	2/3/4	15/14 /13	6	7/8/10	
Configuration	UVLO/stdby	Sense+/Sense-	Enable	In A/B/C	Out A/B/C	Alarm	Op. Amp.	Consumption	
Normal	1	H	+ > -	X	X	L	L	OK	H (1.5mA)
			+ < -	H	IN	IN	H		
Stdby	2	L	+ > -	X	X	L	L	HZ	L (30µA)
			+ < -				H		
UVLO	3	M	X	X	X	L	L	OK	H

#### Configuration 1 : UVLO/stdby = H

The TD310 is in a normal consumption state (1.5mA), the operational amplifier is normally functionning and the buffer outputs are determined by the sense comparator inputs, the enable inputs and the buffer inputs.

#### Configuration 2 : UVLO/stdby = L

The TD310 is in a low consumption state (standby mode 30µA), the buffer outputs are set to low state and the operational amplifier is in high impedance state.

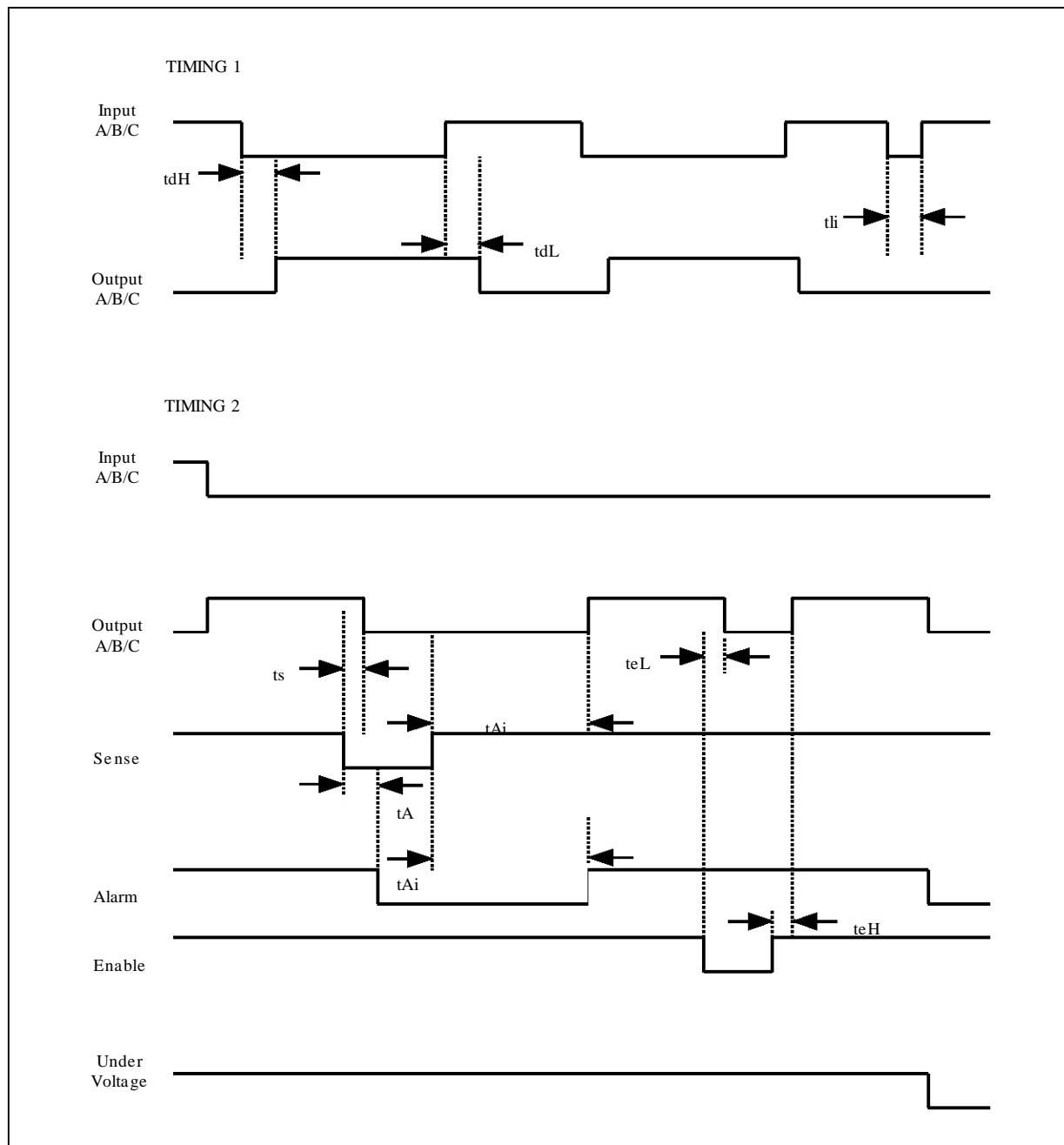
#### Configuration 3 : UVLO/stdby = M

The  $V_{CC}$  supply voltage is between  $V_{UVLO}$  and  $V_{stdby}$  (\*\*). The TD310 remains in a normal consumption state and the operational amplifier is normally functionning but the buffer outputs and the alarm pin are set to low state.

(\*) If the UVLO level remains unadjusted, it is recommended to bypass the UVLO/stdby pin with a 1nF capacitor.

(\*\*) If the supply voltage falls below  $V_{stdby}$ , the TD310 is set in standby mode (configuration 2).

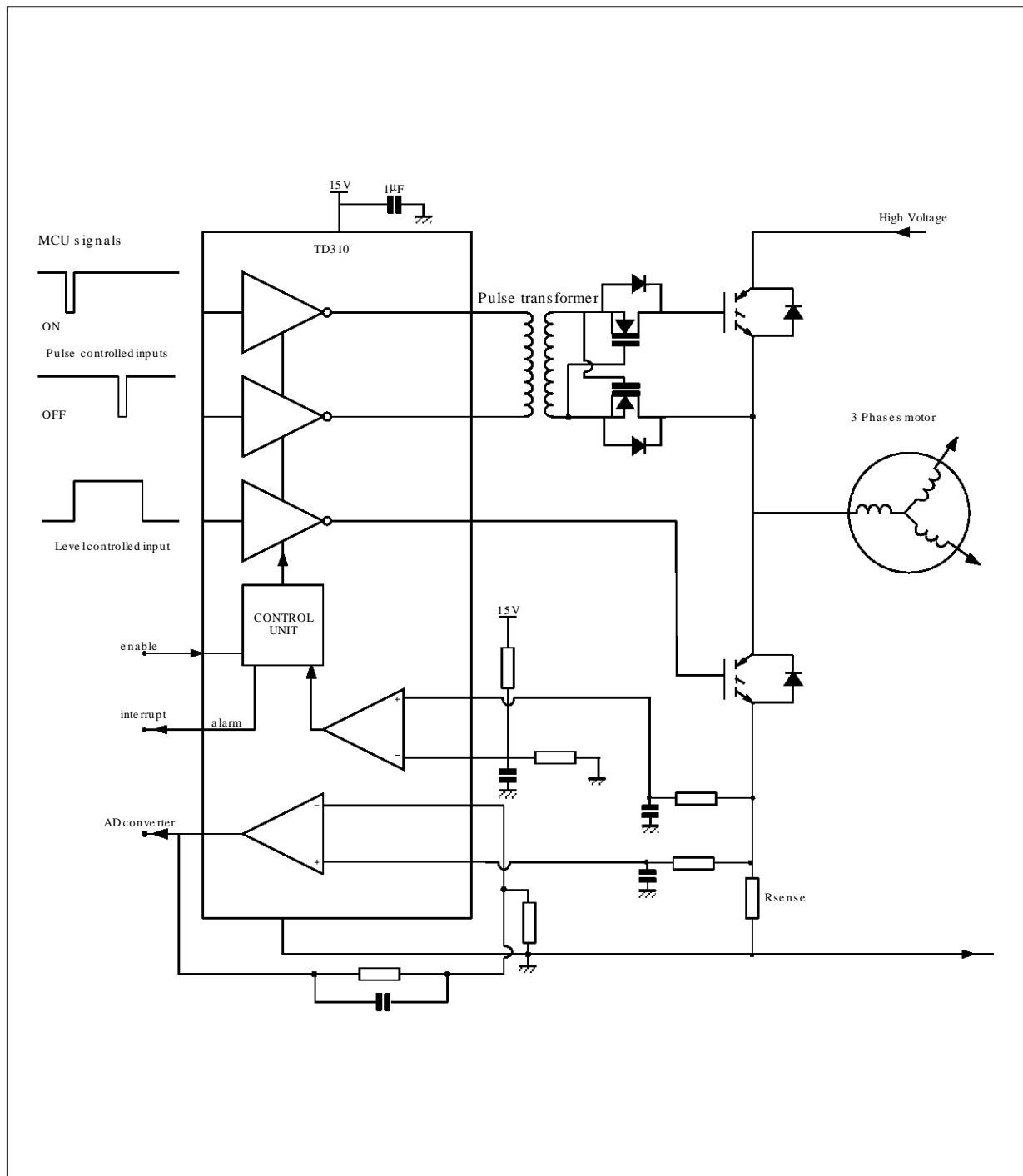
## TIMING DIAGRAM

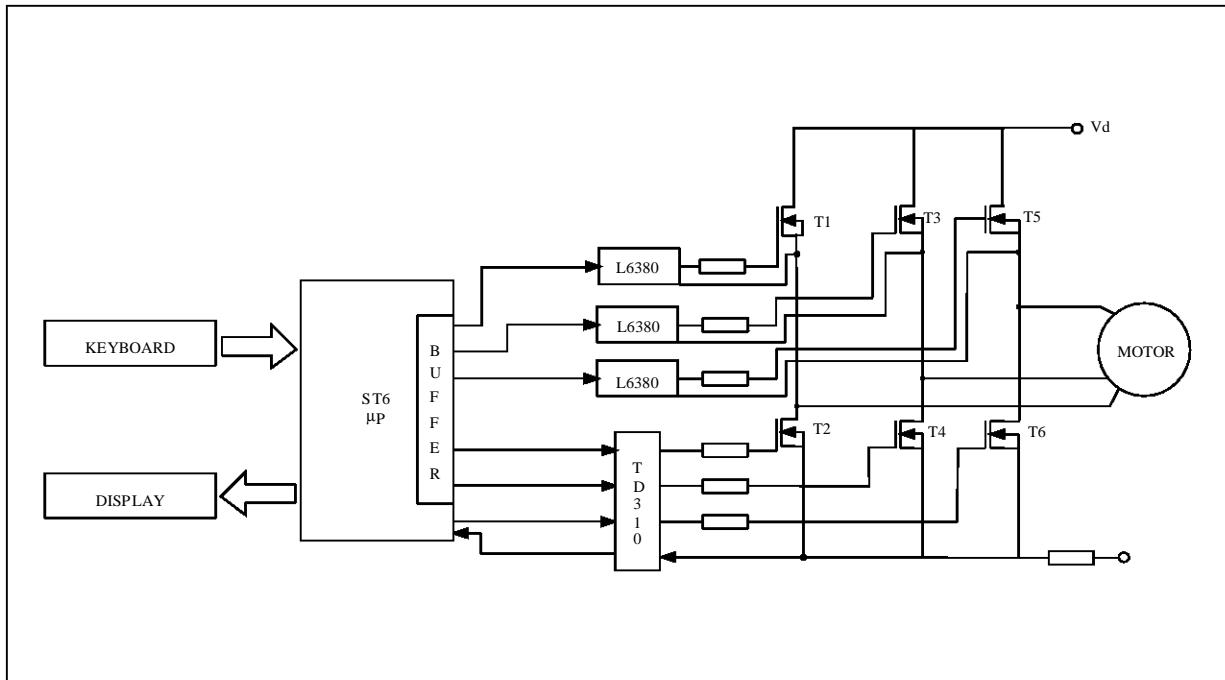
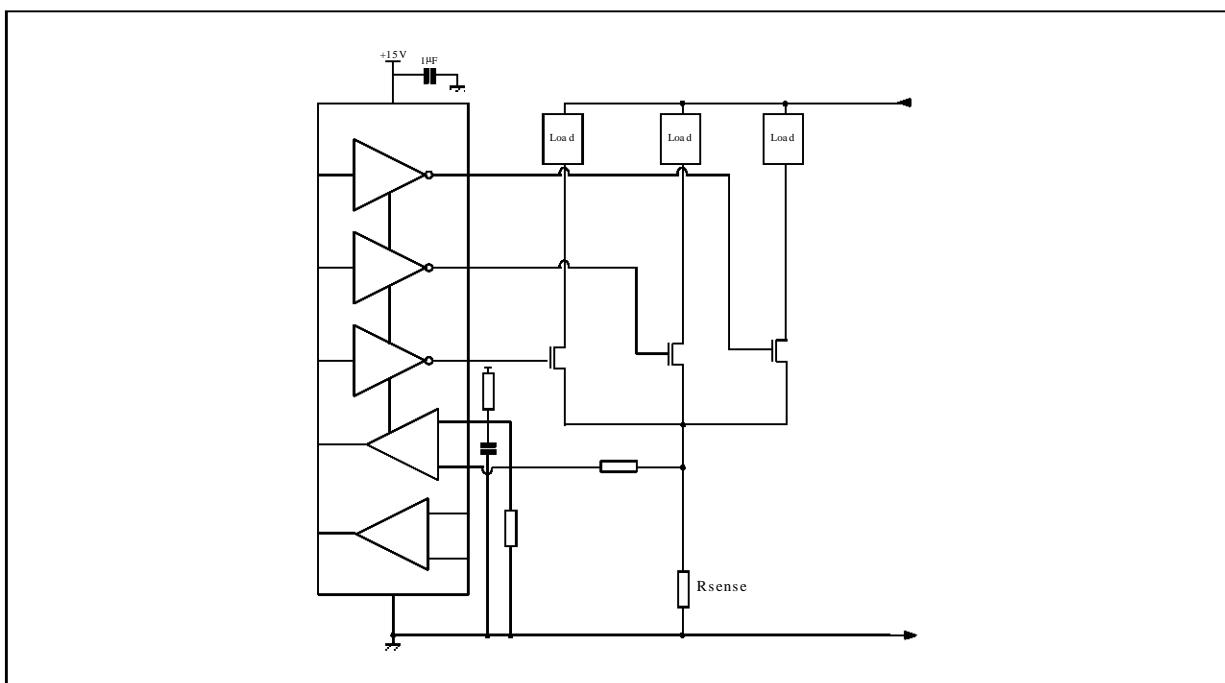


## TD310

### TYPICAL APPLICATIONS

Figure 1 : THREE PHASE MOTOR DRIVE

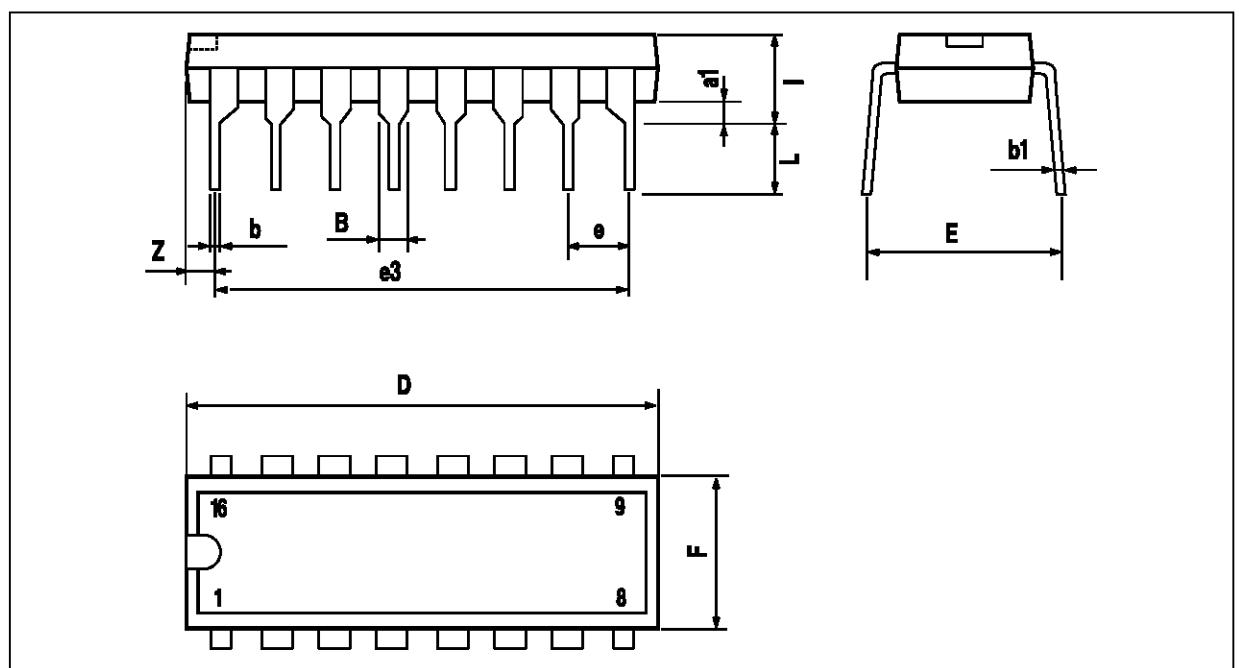


**Figure 2 : THREE PHASE MOTOR LOW SIDE DRIVE****Figure 3 : LOW SIDE DRIVE**

## TD310

### PACKAGE MECHANICAL DATA

16 PINS - PLASTIC DIP

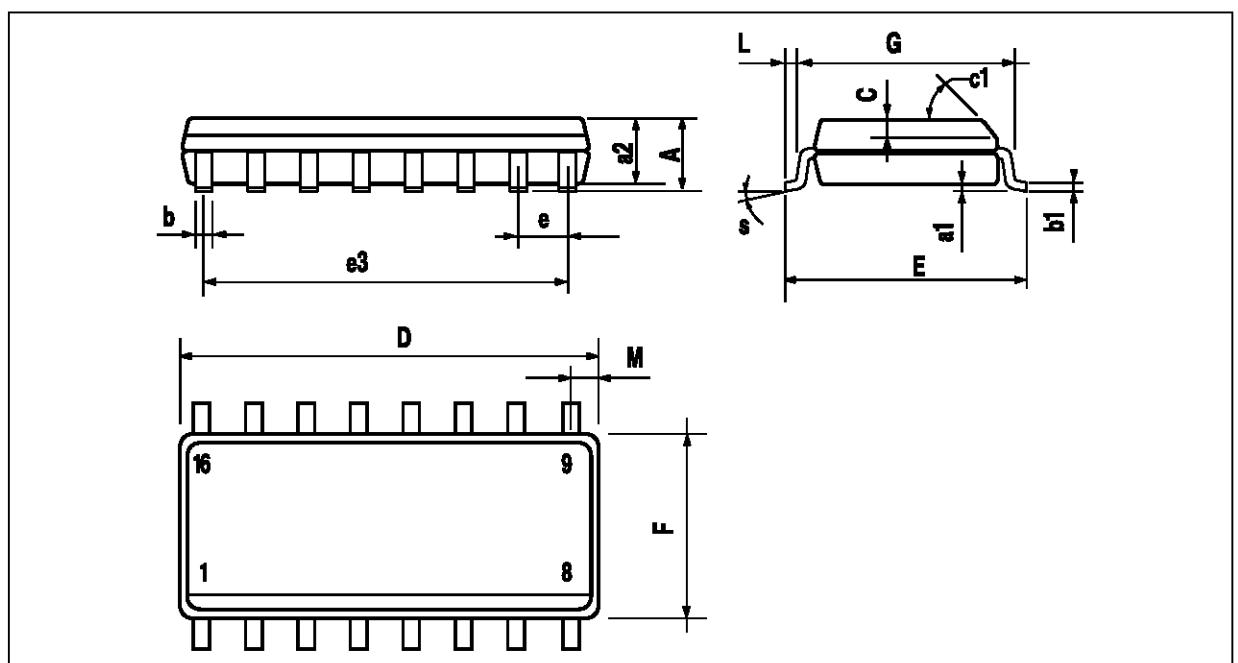


PM-DIP16.EPS

DIP16.TBL

Dimensions	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
a1	0.51			0.020		
B	0.77		1.65	0.030		0.065
b		0.5			0.020	
b1		0.25			0.010	
D			20			0.787
E		8.5			0.335	
e		2.54			0.100	
e3		17.78			0.700	
F			7.1			0.280
i			5.1			0.201
L		3.3			0.130	
Z			1.27			0.050

**PACKAGE MECHANICAL DATA**  
16 PINS - PLASTIC MICROPACKAGE (SO)



PM-SO16.EPS

Dimensions	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.069
a1	0.1		0.2	0.004		0.008
a2			1.6			0.063
b	0.35		0.46	0.014		0.018
b1	0.19		0.25	0.007		0.010
C		0.5			0.020	
c1		45° (typ.)				
D	9.8		10	0.386		0.394
E	5.8		6.2	0.228		0.244
e		1.27			0.050	
e3		8.89			0.350	
F	3.8		4.0	0.150		0.157
G	4.6		5.3	0.181		0.209
L	0.5		1.27	0.020		0.050
M			0.62			0.024

SO16.TBL

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