Floating Bootstrap or Ground-Reference **D PACKAGE High-Side Driver** (TOP VIEW) **Adaptive Dead-Time Control** ENABLE ☐ BOOT 50-ns Max Rise/Fall Times and 100-ns Max 13 IN  $\square$  □ NC Propagation Delay – 3.3-nF Load CROWBAR I 3 12 ☐ HIGHDR 11 NC  $\square$ **Ideal for High-Current Single or Multiphase** SYNC I 10 LOWDR **Power Supplies** □ NC DT  $\Box$ 9 2.4-A Typical Peak Output Current □ V<sub>CC</sub> PGND □ 8 4.5-V to 15-V Supply Voltage Range **PWP PACKAGE Internal Schottky Bootstrap Diode** (TOP VIEW) **SYNC Control for Synchronous or Nonsynchronous Operation** ENABLE \_\_\_ 10 ☐ BOOT IN  $\square$ 2Γ **1**13 □□ NC **CROWBAR for OVP, Protects Against** 3 12 CROWBAR \_\_\_ **Faulted High-Side Power FETs** NC  $\square$ 4 Thermal 11 TT BOOTLO Low Supply Current....3-mA Typical Pad 5  $I_{10}$ SYNC I □ LOWDR -40°C to 125°C Operating Virtual Junction

NC - No internal connection

DT  $\square$ 

PGND [

61

9

8

□ NC

### description

**Packages** 

**Temperature Range** 

Available in SOIC and TSSOP PowerPAD

The TPS2830 and TPS2831 are MOSFET drivers for synchronous-buck power stages. These devices are ideal for designing a high-performance power supply using switching controllers that do not have MOSFET drivers. The drivers are designed to deliver 2.4-A peak currents into large capacitive loads. The high-side driver can be configured as a ground-reference driver or as a floating bootstrap driver. An adaptive dead-time control circuit eliminates shoot-through currents through the main power FETs during switching transitions, providing higher efficiency for the buck regulator. The TPS2830/31 drivers have additional control functions: ENABLE, SYNC, and CROWBAR. Both drivers are off when ENABLE is low. The driver is configured as a nonsynchronous-buck driver, disabling the low side driver when SYNC is low. The CROWBAR function turns on the low-side power FET, overriding the IN signal, for over-voltage protection against faulted high-side power FETs.

The TPS2830 has a noninverting input. The TPS2831 has an inverting input. The TPS2830/31 drivers are available in 14-terminal SOIC and thermally-enhanced TSSOP PowerPAD™ packages, and operate over a virtual junction temperature range of -40°C to 125°C.

### **Related Synchronous MOSFET Drivers**

DEVICE NAME	ADDITIONAL FEATURES	INPUTS		
TPS2832	MANO ENIARI E OVAIO I OROMBAR	01400	Noninverted	
TPS2833	W/O ENABLE, SYNC, and CROWBAR	CMOS	Inverted	
TPS2834	ENABLE OVALO LODOVADAD		Noninverted	
TPS2835	ENABLE, SYNC, and CROWBAR	TTL	Inverted	
TPS2836	MANO ENIARI E OVAIO I OROMBAR		Noninverted	
TPS2837	W/O ENABLE, SYNC, and CROWBAR	TTL	Inverted	



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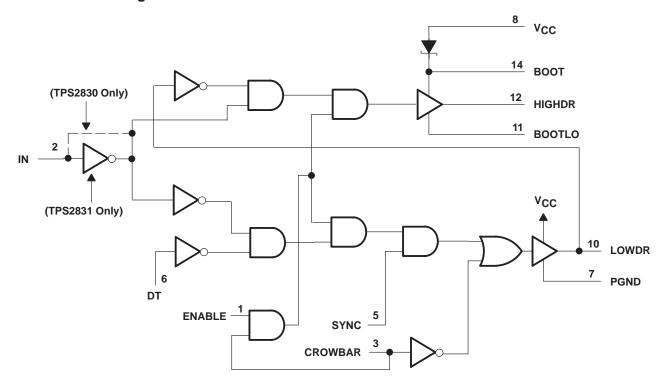


### **AVAILABLE OPTIONS**

	PACKAGED DEVICES				
TJ	SOIC (D)	TSSOP (PWP)			
-40°C to 125°C	TPS2830D TPS2831D	TPS2830PWP TPS2831PWP			

The D and PWP packages are available taped and reeled. Add R suffix to device type (e.g., TPS2830DR)

## functional block diagram





### **Terminal Functions**

TERMII	NAL		DECORPTION
NAME	NO.	1/0	DESCRIPTION
BOOT	14	_	Bootstrap terminal. A ceramic capacitor is connected between BOOT and BOOTLO terminals to develop the floating bootstrap voltage for the high-side MOSFET. The capacitor value is typically between 0.1 $\mu$ F and 1 $\mu$ F. A 1-M $\Omega$ resistor should be connected across the bootstrap capacitor to provide a discharge path when the driver has been powered down.
BOOTLO	11	0	This terminal connects to the junction of the high-side and low-side MOSFETs.
CROWBAR	3	I	CROWBAR can to be driven by an external OVP circuit to protect against a short across the high-side MOSFET. If CROWBAR is driven low, the low-side driver will be turned on and the high-side driver will be turned off, independent of the status of all other control terminals.
DT	6	I	Dead-time control terminal. Connect DT to the junction of the high-side and low-side MOSFETs.
ENABLE	1	I	If ENABLE is low, both drivers are off.
HIGHDR	12	0	Output drive for the high-side power MOSFET
IN	2	I	Input signal to the MOSFET drivers (noninverting input for the TPS2830; inverting input for the TPS2831).
LOWDR	10	0	Output drive for the low-side power MOSFET
NC	4, 9, 13		No internal connection
PGND	7		Power ground. Connect to the FET power ground
SYNC	5	I	Synchronous Rectifier Enable terminal. If SYNC is low, the low-side driver is always off; If SYNC is high, the low-side driver provides gate drive to the low-side MOSFET.
VCC	8	I	Input supply. Recommended that a 1-μF capacitor be connected from V <sub>CC</sub> to PGND.

### detailed description

### low-side driver

The low-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink.

### high-side driver

The high-side driver is designed to drive low Rds(on) N-channel MOSFETs. The current rating of the driver is 2 A, source and sink. The high-side driver can be configured as a GND-reference driver or as a floating bootstrap driver. The internal bootstrap diode is a Schottky, for improved drive efficiency. The maximum voltage that can be applied from BOOT to ground is 30 V.

### dead-time (DT) control<sup>†</sup>

Dead-time control prevents shoot through current from flowing through the main power FETs during switching transitions by controlling the turn-on times of the MOSFET drivers. The high-side driver is not allowed to turn on until the gate drive voltage to the low-side FET is low, and the low-side driver is not allowed to turn on until the voltage at the junction of the power FETs (Vdrain) is low; the DT terminal connects to the junction of the power FETs.

### **ENABLE†**

The ENABLE terminal enables the drivers. When enable is low, the output drivers are low.

### IN†

The IN terminal is the input control signal for the drivers. The TPS2830 has a noninverting input; the TPS2831 has an inverting input.

†High-level input voltages on ENABLE, SYNC, CROWBAR, IN, and DT must be greater than or equal to 0.7V<sub>CC</sub>.



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### detailed description (continued)

### SYNC†

The SYNC terminal controls whether the drivers operate in synchronous or nonsynchronous mode. In synchronous mode, the low-side FET is operated as a synchronous rectifier. In nonsynchronous mode, the low-side FET is always off.

### CROWBAR†

The CROWBAR terminal overrides the normal operation of the driver. When the CROWBAR terminal is low, the low-side FET turns on to act as a clamp, protecting the output voltage of the dc/dc converter against over voltages due to a short across the high-side FET. V<sub>IN</sub> should be fused to protect the low-side FET.

### absolute maximum ratings over operating free-air temperature (unless otherwise noted)

Supply voltage range, V <sub>CC</sub> (see Note 1) –	-0.3 V to 16 V
Input voltage range: BOOT to PGND (high-side driver ON) –	-0.3 V to 30 V
BOOTLO to PGND –	-0.3 V to 16 V
BOOT to BOOTLO –	-0.3 V to 16 V
ENABLE, SYNC, and CROWBAR (see Note 2) –	-0.3 V to 16 V
IN (see Note 2) –	-0.3 V to 16 V
DT (see Note 2) –	-0.3 V to 30 V
Continuous total power dissipation See Dissipation	Rating Table
Operating virtual junction temperature range, T <sub>J</sub>	0°C to 125°C
Storage temperature range, T <sub>stq</sub> 6	5°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	260°C

<sup>‡</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTES: 1. Unless otherwise specified, all voltages are with respect to PGND.

### **DISSIPATION RATING TABLE**

PACKAGE	$\text{T}_{\text{A}} \leq 25^{\circ}\text{C}$	DERATING FACTOR	T <sub>A</sub> = 70°C	T <sub>A</sub> = 85°C
PWP with solder§	2668	26.68 mW/°C	1467	1067
PWP without solder§	1024	10.24 mW/°C	563	409
D	749	7.49 mW/°C	412	300

### JUNCTION-CASE THERMAL RESISTANCE TABLE

PWP	Junction-case thermal resistance	2.07 °C/W

<sup>§</sup> Test Board Conditions:

- 1. Thickness: 0.062"
- 2.  $3^{\prime\prime}\times3^{\prime\prime}$  (for packages <27 mm long)
- 3.  $4'' \times 4''$  (for packages >27 mm long)
- 4. 2 oz copper traces located on the top of the board (0.071 mm thick)
- 5. Copper areas located on the top and bottom of the PCB for soldering
- 6. Power and ground planes, 1 oz copper (0.036 mm thick)
- 7. Thermal vias, 0.33 mm diameter, 1.5 mm pitch
- 8. Thermal isolation of power plane

For more information, refer to TI technical brief, literature number SLMA002.

†High-level input voltages on ENABLE, SYNC, CROWBAR, IN, and DT must be greater than or equal to 0.7V<sub>CC</sub>.



<sup>2.</sup> High-level input voltages on the ENABLE, SYNC, CROWBAR, IN, and DT terminals must be greater than or equal to 0.7V<sub>CC</sub>.

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### recommended operating conditions

		MIN	NOM MAX	UNIT
Supply voltage,	Vcc	4.5	15	V
Input voltage	BOOT to PGND	4.5	28	V

# electrical characteristics over recommended operating virtual junction temperature range, $V_{CC}$ = 6.5 V, ENABLE = High, $C_L$ = 3.3 nF (unless otherwise noted)

### supply current

	PARAMETER		TEST CONDITIONS	5	MIN	TYP	MAX	UNIT
VCC	Supply voltage range				4.5		15	V
		VENABLE = LOW,	V <sub>CC</sub> =15 V				100	μΑ
		V <sub>ENABLE</sub> = HIGH,	V <sub>CC</sub> =15 V			0.1		
VCC	Quiescent current	VENABLE = HIGH, BOOTLO grounded, See Note 3	V <sub>CC</sub> =12 V, CHIGHDR = 50 pF,	$f_{SWX} = 200 \text{ kHz},$ $C_{LOWDR} = 50 \text{ pF},$		3		mA

NOTE 3: Ensured by design, not production tested.

### output drivers

	PARAMETE	२	TEST CONDIT	TIONS	MIN	TYP	MAX	UNIT
		Duty cycle < 2%,	V <sub>BOOT</sub> - V <sub>BOOTLO</sub> = 4.5 V	, VHIGHDR = 4 V	0.7	1.1		
	High-side sink (see Note 4)	t <sub>pw</sub> < 100 μs	VBOOT - VBOOTLO = 6.5 V	, VHIGHDR = 5 V	1.1	1.5		Α
	(500 14010 4)	(see Note 3)	VBOOT - VBOOTLO = 12 V,	VHIGHDR = 10.5 V	2	2.4		
	High-side	Duty cycle < 2%,	VBOOT - VBOOTLO = 4.5 V	, VHIGHDR = 0.5V	1.2	1.4		
	source	t <sub>pw</sub> < 100 μs	VBOOT - VBOOTLO = 6.5 V	, VHIGHDR = 1.5 V	1.3	1.6		Α
Peak output-	(see Note 4)	(see Note 3)	VBOOT - VBOOTLO = 12 V,	VHIGHDR = 1.5 V	2.3	2.7		
current		Duty cycle < 2%,	$V_{CC} = 4.5 \text{ V},$	V <sub>LOWDR</sub> = 4 V	1.3	1.8		
	Low-side sink (see Note 4)	t <sub>pw</sub> < 100 μs	$V_{CC} = 6.5 \text{ V},$	V <sub>LOWDR</sub> = 5 V	2	2.5		Α
	(000 11010 1)	(see Note 3)	$V_{CC} = 12 V$ ,	$V_{LOWDR} = 10.5 V$	3	3.5		
	Low-side	Duty cycle < 2%,	$V_{CC} = 4.5 \text{ V},$	$V_{LOWDR} = 0.5V$	1.4	1.7		
	source	t <sub>pw</sub> < 100 μs (see Note 3)	V <sub>CC</sub> = 6.5 V,	$V_{LOWDR} = 1.5 V$	2	2.4		Α
	(see Note 4)		V <sub>CC</sub> = 12 V,	V <sub>LOWDR</sub> = 1.5 V	2.5	3		
			VBOOT - VBOOTLO = 4.5 V	, VHIGHDR = 0.5 V			5	
	High-side sink (s	ee Note 4)	$V_{BOOT} - V_{BOOTLO} = 6.5 V$			5	Ω	
			VBOOT - VBOOTLO = 12 V, VHIGHDR = 0.5 V				5	
			V <sub>BOOT</sub> – V <sub>BOOTLO</sub> = 4.5 V, V <sub>HIGHDR</sub> = 4 V				75	
	High-side source	(see Note 4)	VBOOT - VBOOTLO = 6.5 V, VHIGHDR = 6 V				75	Ω
Output			VBOOT - VBOOTLO = 12 V, VHIGHDR =11.5 V				75	
resistance			V <sub>DRV</sub> = 4.5 V,	$V_{LOWDR} = 0.5 V$			9	
	Low-side sink (se	ee Note 4)	V <sub>DRV</sub> = 6.5 V	$V_{LOWDR} = 0.5 V$			7.5	Ω
			$V_{DRV} = 12 V$ ,	$V_{LOWDR} = 0.5 V$			6	
			V <sub>DRV</sub> = 4.5 V,	V <sub>LOWDR</sub> = 4 V			75	
	Low-side source	(see Note 4)	V <sub>DRV</sub> = 6.5 V,	V <sub>LOWDR</sub> = 6 V			75	Ω
			V <sub>DRV</sub> = 12 V,	V <sub>LOWDR</sub> = 11.5 V			75	

### NOTES: 3. Ensured by design, not production tested.

<sup>4.</sup> The pullup/pulldown circuits of the drivers are bipolar and MOSFET transistors in parallel. The peak output current rating is the combined current from the bipolar and MOSFET transistors. The output resistance is the Rds(on) of the MOSFET transistor when the voltage on the driver output is less than the saturation voltage of the bipolar transistor.



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electrical characteristics over recommended operating virtual junction temperature range,  $V_{CC}$  = 6.5 V, ENABLE = High,  $C_L$  = 3.3 nF (unless otherwise noted) (continued)

### dead-time control

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input voltage	LOWIDD	Over the Man was reading Nets 2)	0.7V <sub>CC</sub>			V
VIL	Low-level input voltage	LOWDR	Over the V <sub>CC</sub> range (see Note 3)			1	V
$V_{IH}$	High-level input voltage	DT	Over the Valarange	0.7V <sub>CC</sub>			V
VIL	Low-level input voltage	יטן	Over the V <sub>CC</sub> range			1	V

NOTE 3: Ensured by design, not production tested.

### digital control terminals (IN, CROWBAR, ENABLE, SYNC)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIH	High-level input voltage	Over the Ver years	0.7V <sub>CC</sub>			V
VIL	Low-level input voltage	Over the V <sub>CC</sub> range			1	V

# switching characteristics over recommended operating virtual junction temperature range, ENABLE = High, $C_L = 3.3$ nF (unless otherwise noted)

F	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT		
		VBOOT = 4.5 V,	V <sub>BOOTLO</sub> = 0 V			60			
l	HIGHDR output (see Note 3)	$V_{BOOT} = 6.5 V$	V <sub>BOOTLO</sub> = 0 V			50	ns		
Dies time		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			50			
Rise time		V <sub>CC</sub> = 4.5 V				40			
	LOWDR output (see Note 3)	V <sub>CC</sub> = 6.5 V				30	ns		
		V <sub>CC</sub> = 12 V				30			
		$V_{BOOT} = 4.5 \text{ V},$	V <sub>BOOTLO</sub> = 0 V			60			
	HIGHDR output (see Note 3)	$V_{BOOT} = 6.5 \text{ V},$	V <sub>BOOTLO</sub> = 0 V			50	ns		
Fall time		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			50			
raii iime	LOWDR output (see Note 3)	V <sub>CC</sub> = 4.5 V				40			
		V <sub>CC</sub> = 6.5 V				30	) ns		
		V <sub>CC</sub> = 12 V				30			
	HIGHDR going low (excluding dead time) (see Note 3)	$V_{BOOT} = 4.5 V$	$V_{BOOTLO} = 0 V$			130			
		$V_{BOOT} = 6.5 \text{ V},$	V <sub>BOOTLO</sub> = 0 V			100	ns		
Dranagation delay time		V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			75			
Propagation delay time	LOW/DD are in a bring	$V_{BOOT} = 4.5 V,$	V <sub>BOOTLO</sub> = 0 V			80			
	LOWDR going high (excluding dead time) (see Note 3)	$V_{BOOT} = 6.5 \text{ V},$	V <sub>BOOTLO</sub> = 0 V			70	ns		
	(exercise)	V <sub>BOOT</sub> = 12 V,	V <sub>BOOTLO</sub> = 0 V			60			
	1.014/55	V <sub>CC</sub> = 4.5 V				80			
Propagation delay time	LOWDR going low (excluding dead time) (see Note 3)	$V_{CC} = 6.5 \text{ V}$				70	ns		
	(excluding acad time) (ecc rete e)	V <sub>CC</sub> = 12 V				60			
	DT (= LOWED = = d	V <sub>CC</sub> = 4.5 V		40		170			
Driver nonoverlap time	DT to LOWDR and LOWDR to HIGHDR (see Note 3)	V <sub>CC</sub> = 6.5 V		25		135	ns		
	LOWER TO FINGUE (See Note 3)	V <sub>CC</sub> = 12 V		15		85			

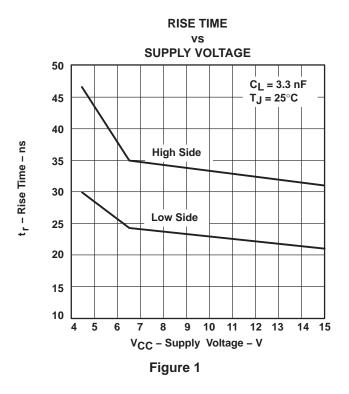
NOTE 3: Ensured by design, not production tested.

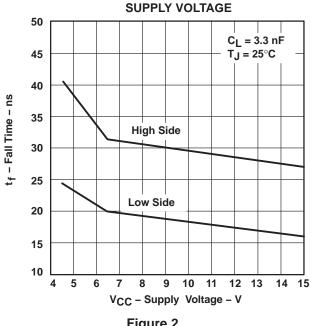


**FALL TIME** 

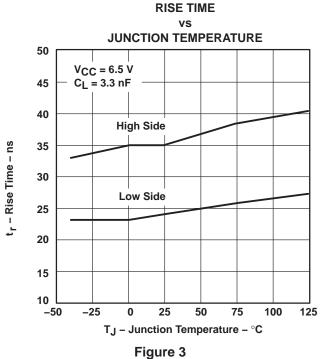
vs

### **TYPICAL CHARACTERISTICS**









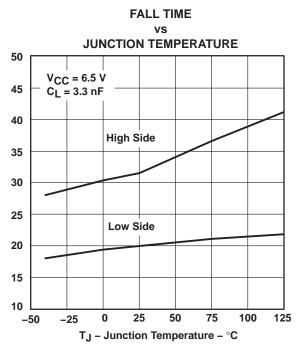
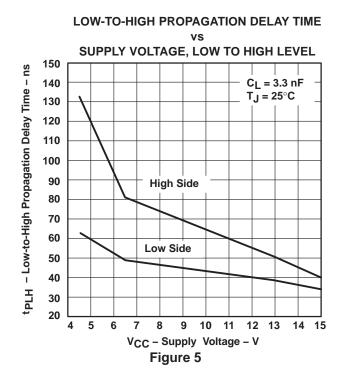
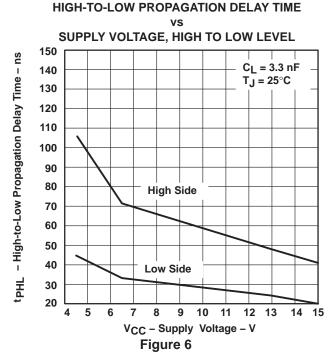


Figure 4

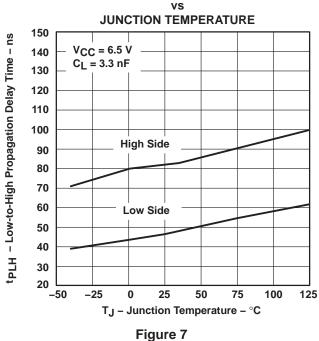
tf - Fall Time - ns

### TYPICAL CHARACTERISTICS





# **LOW-TO-HIGH PROPAGATION DELAY TIME**



### **HIGH-TO-LOW PROPAGATION DELAY TIME**

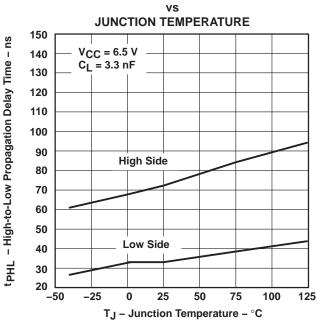
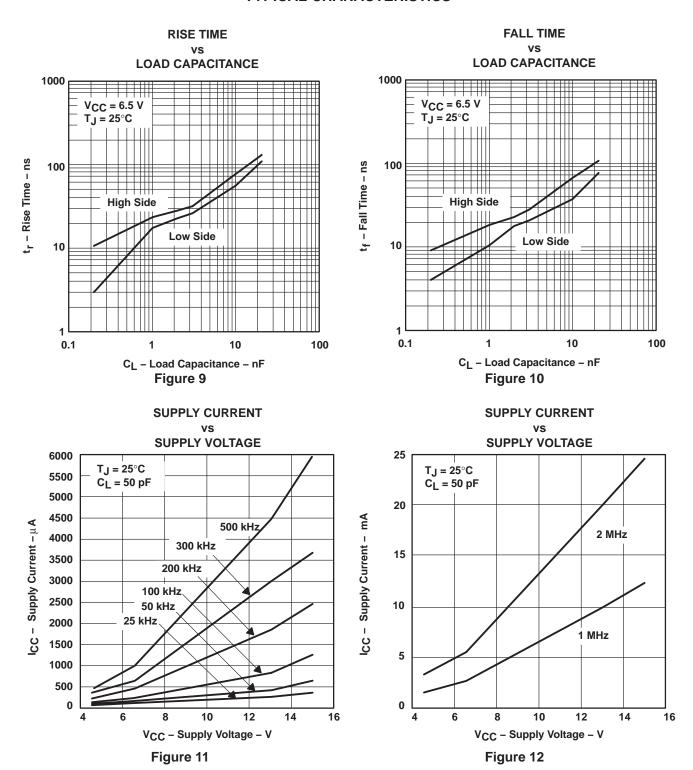


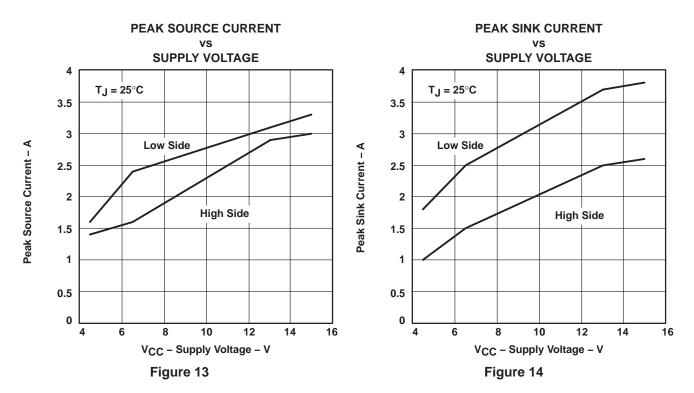
Figure 8

### **TYPICAL CHARACTERISTICS**





### **TYPICAL CHARACTERISTICS**



### INPUT THRESHOLD VOLTAGE

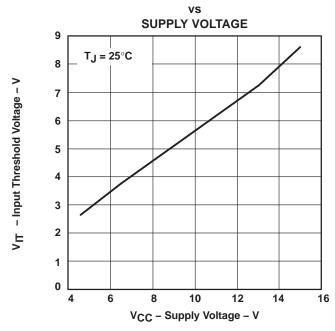


Figure 15

### **APPLICATION INFORMATION**

Figure 16 shows the circuit schematic of a 100-kHz synchronous-buck converter implemented with a TL5001A pulse-width-modulation (PWM) controller and a TPS2831 driver. The converter operates over an input range from 4.5 V to 12 V and has a 3.3-V output. The circuit can supply 3 A continuous load. The converter achieves an efficiency of 94% for  $V_{\text{IN}} = 5 \text{ V}$ ,  $I_{\text{load}} = 1 \text{ A}$ , and 93% for  $V_{\text{in}} = 5 \text{ V}$ ,  $I_{\text{load}} = 3 \text{ A}$ .

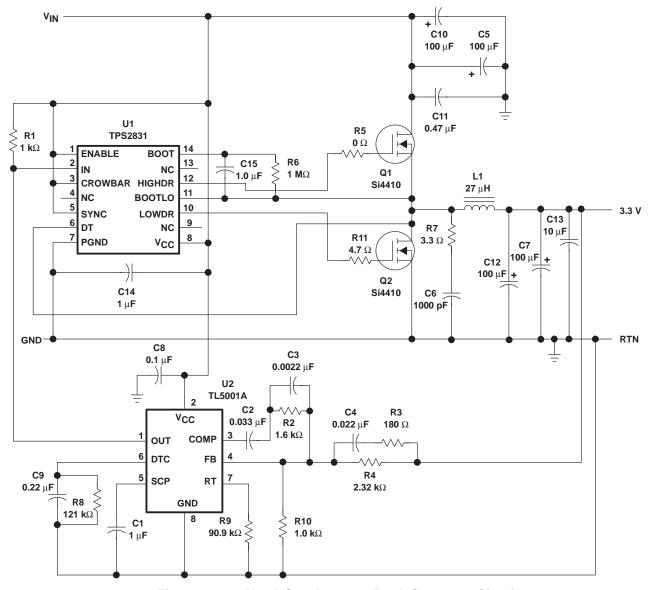


Figure 16. 3.3-V 3-A Synchronous-Buck Converter Circuit

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### APPLICATION INFORMATION

Great care should be taken when laying out the PC board. The power-processing section is the most critical and will generate large amounts of EMI if not properly configured. The junction of Q1, Q2, and L1 should be very tight. The connection from Q1 drain to the positive sides of C5, C10, and C11 and the connection from Q2 source to the negative sides of C5, C10, and C11 should be as short as possible. The negative terminals of C7 and C12 should also be connected to Q2 source.

Next, the traces from the MOSFET driver to the power switches should be considered. The BOOTLO signal from the junction of Q1 and Q2 carries the large gate drive current pulses and should be as heavy as the gate drive traces. The bypass capacitor (C14) should be tied directly across  $V_{CC}$  and PGND.

The next most sensitive node is the FB node on the controller (terminal 4 on the TL5001A) This node is very sensitive to noise pickup and should be isolated from the high-current power stage and be as short as possible. The ground around the controller and low-level circuitry should be tied to the power ground as the output. If these three areas are properly laid out, the rest of the circuit should not have any other EMI problems and the power supply will be relatively free of noise.







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### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
TPS2830D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2830DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2830DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2830DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2830PWP	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2830PWPG4	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2830PWPR	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2830PWPRG4	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2831D	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2831DG4	ACTIVE	SOIC	D	14	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2831DR	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2831DRG4	ACTIVE	SOIC	D	14	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2831PWP	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2831PWPG4	ACTIVE	HTSSOP	PWP	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2831PWPR	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TPS2831PWPRG4	ACTIVE	HTSSOP	PWP	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

<sup>&</sup>lt;sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <a href="http://www.ti.com/productcontent">http://www.ti.com/productcontent</a> for the latest availability information and additional product content details.



### PACKAGE OPTION ADDENDUM

16-Mar-2007

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2830DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TPS2830PWPR	HTSSOP	PWP	14	2000	330.0	12.4	6.67	5.4	1.6	8.0	12.0	Q1
TPS2831DR	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TPS2831PWPR	HTSSOP	PWP	14	2000	330.0	12.4	6.67	5.4	1.6	8.0	12.0	Q1





\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2830DR	SOIC	D	14	2500	346.0	346.0	33.0
TPS2830PWPR	HTSSOP	PWP	14	2000	346.0	346.0	29.0
TPS2831DR	SOIC	D	14	2500	346.0	346.0	33.0
TPS2831PWPR	HTSSOP	PWP	14	2000	346.0	346.0	29.0

# D (R-PDSO-G14)

# PLASTIC SMALL-OUTLINE PACKAGE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
- Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
- E. Reference JEDEC MS-012 variation AB.



# PWP (R-PDSO-G\*\*)

## PowerPAD™ PLASTIC SMALL-OUTLINE PACKAGE

20 PIN SHOWN



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusions. Mold flash and protrusion shall not exceed 0.15 per side.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <a href="https://www.ti.com">www.ti.com</a>.
- E. Falls within JEDEC MO-153

PowerPAD is a trademark of Texas Instruments.



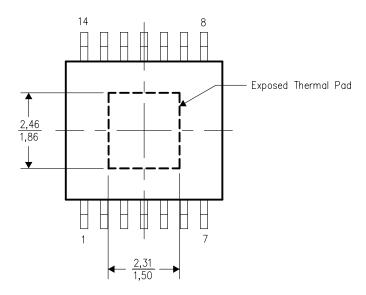
# THERMAL PAD MECHANICAL DATA PWP (R-PDSO-G14)

### THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

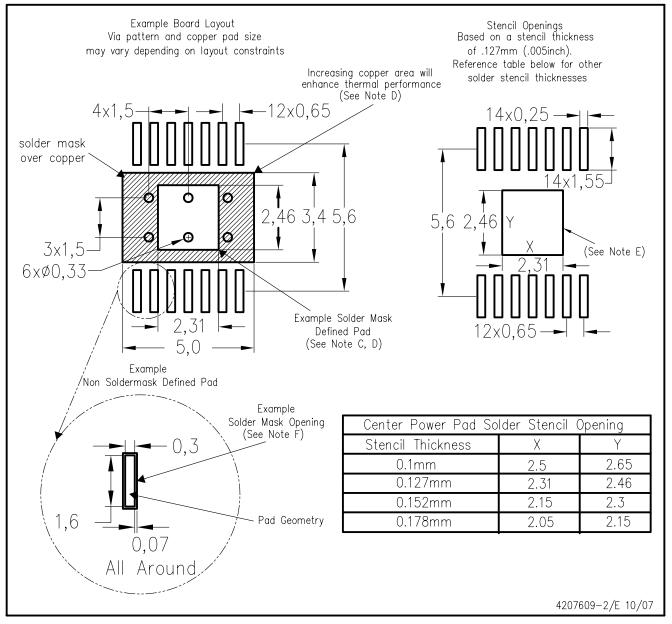


Top View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

# PWP (R-PDSO-G14) PowerPAD™



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="https://www.ti.com">www.ti.com</a>. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
- F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PowerPAD is a trademark of Texas Instruments.



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