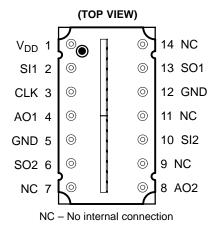


- 128 × 1 Sensor-Element Organization
- 200 Dots-Per-Inch (DPI) Sensor Pitch
- High Linearity and Uniformity
- Wide Dynamic Range . . . 2000:1 (66 dB)
- Output Referenced to Ground
- Low Image Lag . . . 0.5% Typ
- Operation to 5 MHz
- Single 5-V Supply
- Replacement for TSL202

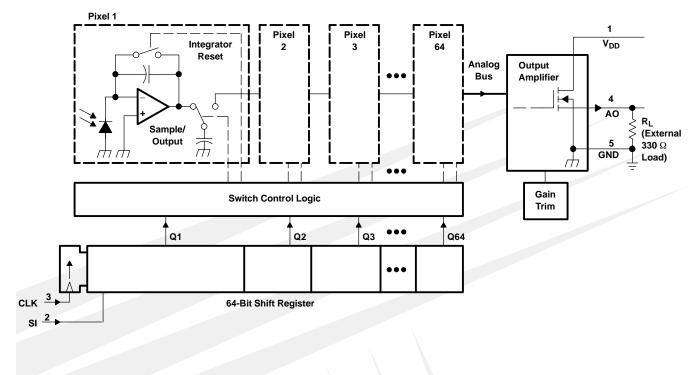


### **Description**

The TSL202R linear sensor array consists of two sections of 64 photodiodes and associated charge amplifier circuitry arranged to form a contiguous  $128 \times 1$  array. The pixels measure  $120 \, \mu m$  (H) by  $70 \, \mu m$  (W) with  $125 - \mu m$  center-to-center spacing and  $55 - \mu m$  spacing between pixels. Operation is simplified by internal control logic that requires only a serial-input (SI) signal and a clock.

The TSL202R is intended for use in a wide variety of applications including mark detection and code reading, optical character recognition (OCR) and contact imaging, edge detection and positioning as well as optical linear and rotary encoding.

## Functional Block Diagram (each section — pin numbers apply to section 1)



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#### **Terminal Functions**

TERMINAL		
NAME	NO.	DESCRIPTION
AO1	4	Analog output of section 1
AO2	8	Analog output of section 2
CLK	3	Clock. Clk controls charge transfer, pixel output, and reset.
GND	5,12	Ground (substrate). All voltages are referenced to GND.
NC	7, 9, 11, 14	No internal connection
SI1	2	Serial input (section 1). SI1 defines the start of the data-out sequence.
SI2	10	Serial input (section 2). SI2 defines the start of the data-out sequence.
SO1	13	Serial output (section 1). SO1 provides a signal to drive the SI2 input.
SO2	6	Serial output (section 2). SO2 provides a signal to drive the SI input of another device for cascading or as an end-of-data indication.
$V_{DD}$	1	Supply voltage. Supply voltage for both analog and digital circuitry.

#### **Detailed Description**

The sensor consists of 128 photodiodes arranged in a linear array. Light energy impinging on a photodiode generates photocurrent, which is integrated by the active integration circuitry associated with that pixel. During the integration period, a sampling capacitor connects to the output of the integrator through an analog switch. The amount of charge accumulated at each pixel is directly proportional to the light intensity and the integration time. The integration time is the interval between two consecutive output periods.

The output and reset of the integrators is controlled by a 128-bit shift register and reset logic. An output cycle is initiated by clocking in a logic 1 on SI for one positive going clock edge (see Figures1 and 2)<sup>†</sup>. As the SI pulse is clocked through the 128-bit shift register, the charge on the sampling capacitor of each pixel is sequentially connected to a charge-coupled output amplifier that generates a voltage output, AO. When the bit position goes low, the pixel integrator is reset. On the 129th clock rising edge, the SI pulse is clocked out of the shift register and the output assumes a high-impedance state. Note that this 129th clock pulse is required to terminate the output of the 128th pixel and return the internal logic to a known state. A subsequent SI pulse can be presented as early as the 130th clock pulse, thereby initiating another pixel output cycle.

The voltage developed at analog output (AO) is given by:

$$V_{out} = V_{drk} + (R_e) (E_e) (t_{int})$$

where:

 $V_{out}$  is the analog output voltage for white condition  $V_{drk}$  is the analog output voltage for dark condition

R<sub>e</sub> is the device responsivity for a given wavelength of light given in V/(μJ/cm<sup>2</sup>)

 $\begin{array}{ll} E_e & \text{is the incident irradiance in $\mu$W/cm$}^2 \\ t_{int} & \text{is integration time in seconds} \end{array}$ 

AO is driven by a source follower that requires an external pulldown resistor (330- $\Omega$  typical). The output is nominally 0 V for no light input, 2 V for normal white-level, and 3.4 V for saturation light level. When the device is not in the output phase, AO is in a high impedance state.

A 0.1  $\mu$ F bypass capacitor should be connected between  $V_{DD}$  and ground as close as possible to the device.

<sup>&</sup>lt;sup>†</sup> For proper operation, after meeting the minimum hold time condition, SI must go low before the next rising edge of the clock.



## Absolute Maximum Ratings<sup>†</sup>

Supply voltage range, V <sub>DD</sub>	0.3 V to 6 V
Input voltage range, V <sub>I</sub>	$\dots$ -0.3 V to V <sub>DD</sub> + 0.3V
Input clamp current, $I_{IK}$ ( $V_I < 0$ or $V_I > V_{DD}$ )	–20 mA to 20 mA
Output clamp current, $I_{OK}$ ( $V_O < 0$ or $V_O > V_{DD}$ )	–25 mA to 25 mA
Voltage range applied to any output in the high impedance or	
power-off state, V <sub>O</sub>	$-0.3 \text{ V to V}_{DD} + 0.3 \text{ V}$
Continuous output current, $I_O(V_O = 0 \text{ to } V_{DD})$	–25 mA to 25 mA
Continuous current through V <sub>DD</sub> or GND	–40 mA to 40 mA
Analog output current range, I <sub>O</sub>	–25 mA to 25 mA
Operating free-air temperature range, T <sub>A</sub>	–25°C to 85°C
Storage temperature range, T <sub>sta</sub>	–25°C to 85°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds .	260°C
ESD tolerance, human body model	2000 V

<sup>†</sup> Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### **Recommended Operating Conditions (see Figure 1 and Figure 2)**

	MIN	NOM	MAX	UNIT
Supply voltage, V <sub>DD</sub>	4.5	5	5.5	V
Input voltage, V <sub>I</sub>	0		$V_{DD}$	V
High-level input voltage, V <sub>IH</sub>	2		$V_{DD}$	V
Low-level input voltage, V <sub>IL</sub>	0		0.8	V
Wavelength of light source, $\lambda$	400		1000	nm
Clock frequency, f <sub>clock</sub>	5		5000	kHz
Sensor integration time, serial, t <sub>int</sub>	0.026		100	ms
Sensor integration time, parallel, t <sub>int</sub>	0.013		100	ms
Operating free-air temperature, T <sub>A</sub>	0		70	°C
Load resistance, R <sub>L</sub>	300		4700	Ω
Load capacitance, C <sub>L</sub>			420	pF

## Electrical Characteristics at $f_{clock}$ = 1 MHz, $V_{DD}$ = 5 V, $T_A$ = 25°C, $\lambda_p$ = 640 nm, $t_{int}$ = 5 ms, $R_L = 330 \Omega$ , $E_e = 16.5 \mu W/cm^2$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
V <sub>out</sub>	Analog output voltage (white, average over 128 pixels)	See Note 1	1.6	2	2.4	V	
$V_{drk}$	Analog output voltage (dark, average over 128 pixels)		0	50	150	mV	
PRNU	Pixel response nonuniformity	See Notes 2 & 3		±4%	±10%		
	Nonlinearity of analog output voltage	See Note 3		±0.4%		FS	
	Output noise voltage	See Note 4		1		mVrms	
R <sub>e</sub>	Responsivity		18	23	30	V/ (μJ/cm <sup>2</sup> )	
SE	Saturation exposure	See Note 5		142		nJ/cm <sup>2</sup>	
V <sub>sat</sub>	Analog output saturation voltage		2.5	3.4		V	
DSNU	Dark signal nonuniformity	All pixels See Note 6		25	120	mV	
IL	Image lag	See Note 7		0.5%			
I <sub>DD</sub>	Supply current, output idle			5	8	mA	
I <sub>IH</sub>	High-level input current	$V_I = V_{DD}$			10	μΑ	
I <sub>IL</sub>	Low-level input current	V <sub>I</sub> = 0			10	μΑ	
.,	High-level output voltage, SO1 and SO2	I <sub>O</sub> = 50 μA	4.5	4.95		V	
V <sub>OH</sub>		I <sub>O</sub> = 4 mA		4.6			
.,	Low-level output voltage, SO1 and SO2	I <sub>O</sub> = 50 μA		0.01	0.1	V	
$V_{OL}$		I <sub>O</sub> = 4 mA		0.4			
C <sub>i(SI)</sub>	Input capacitance, SI			5		pF	
C <sub>i(CLK)</sub>	Input capacitance, CLK			10		pF	

- NOTES: 1. The array is uniformly illuminated with a diffused LED source having a peak wavelength of 640 nm.
  - 2. PRNU is the maximum difference between the voltage from any single pixel and the average output voltage from all pixels of the device under test when the array is uniformly illuminated at the white irradiance level. PRNU includes DSNU.
  - 3. Nonlinearity is defined as the maximum deviation from a best-fit straight line over the dark-to-white irradiance levels, as a percent of analog output voltage (white).
  - 4. RMS noise is the standard deviation of a single-pixel output under constant illumination as observed over a 5-second period.
  - 5. Minimum saturation exposure is calculated using the minimum V<sub>sat</sub>, the maximum V<sub>drk</sub>, and the maximum R<sub>e</sub>.
  - 6. DSNU is the difference between the maximum and minimum output voltage in the absence of illumination.
  - 7. Image lag is a residual signal left in a pixel from a previous exposure. It is defined as a percent of white-level signal remaining after a pixel is exposed to a white condition followed by a dark condition:

$$IL = \frac{V_{out (IL)} - V_{drk}}{V_{out (white)} - V_{drk}} \times 100$$

### Timing Requirements (see Figure 1 and Figure 2)

		MIN	NOM	MAX	UNIT
t <sub>su(SI)</sub>	Setup time, serial input (see Note 8)	20			ns
t <sub>h(SI)</sub>	Hold time, serial input (see Note 8 and Note 9)	0			ns
t <sub>w</sub>	Pulse duration, clock high or low	50			ns
t <sub>r</sub> , t <sub>f</sub>	Input transition (rise and fall) time	0		500	ns

NOTES: 8. Input pulses have the following characteristics:  $t_f = 6$  ns,  $t_f = 6$  ns.

9. SI must go low before the rising edge of the next clock pulse.

# Dynamic Characteristics over recommended ranges of supply voltage and operating free-air temperature (see Figure 2)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ts	Analog output settling time to $\pm 1\%$	$R_L = 330 \ \Omega,  C_L = 10 \ pF$		185		ns
t <sub>pd(SO)</sub>	Propagation delay time, SO1, SO2			50		ns

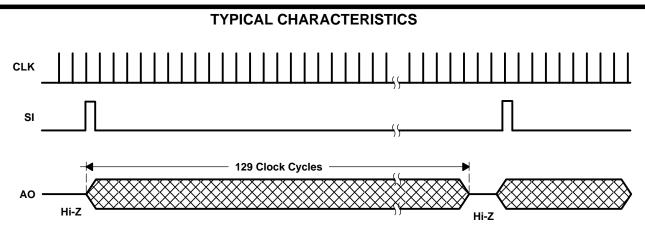


Figure 1. Timing Waveforms

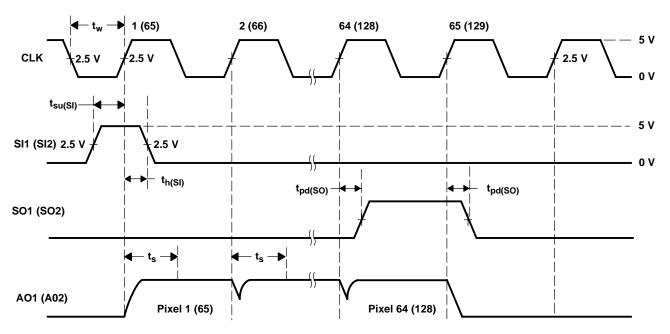


Figure 2. Operational Waveforms (each section)

## **TYPICAL CHARACTERISTICS**

## 

Figure 3

#### LOAD CAPACITANCE AND RESISTANCE 600 470 pF $V_{DD} = 5 V$ $V_{out} = 1 V$ 500 t<sub>s</sub> — Settling Time to 1% — ns 220 pF 400 100 pF 300 10 pF 200 100 0 200 400 600 800 1000 1200 0 $\mbox{R}_{\mbox{\scriptsize L}}$ – Load Resistance – $\Omega$

ANALOG OUTPUT SETTLING TIME

#### **APPLICATION INFORMATION**

#### **Power Supply Considerations**

For optimum device performance, power-supply lines should be decoupled by a 0.01- $\mu$ F to 0.1- $\mu$ F capacitor with short leads mounted close to the device package (see Figure 5 and Figure 6).

#### **Connection Diagrams**

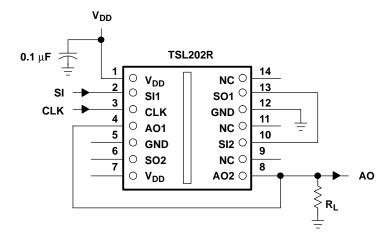
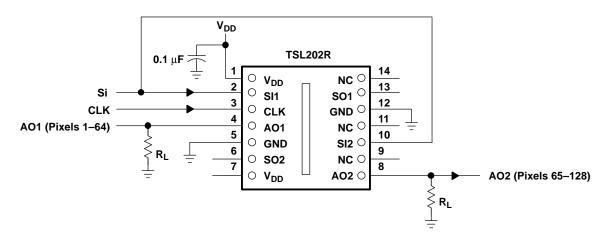


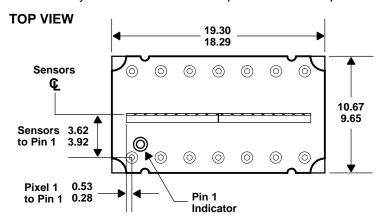
Figure 5. Serial Connection

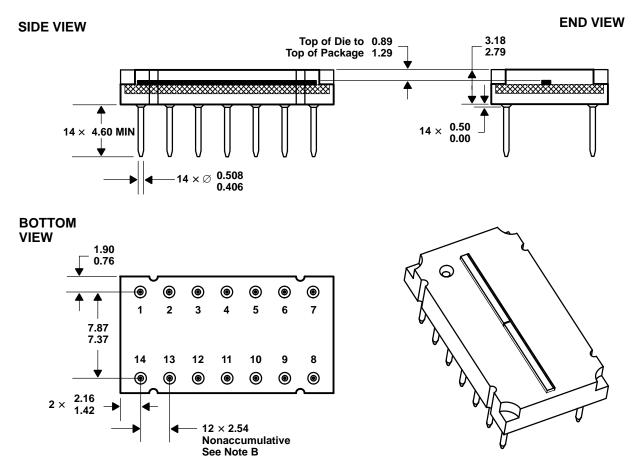


**Figure 6. Parallel Connection** 

#### **MECHANICAL INFORMATION**

This assembly consists of 2 sensor chips mounted on a printed-circuit board in a clear molded plastic package.





NOTES: A. All linear dimensions are in millimeters.

- B. The true-position spacing is 2.54 mm between lead centerlines. Each pin centerline is located within 0.25 mm of its true longitudinal positions.
- C. Index of refraction of clear plastic is 1.52.
- D. This drawing is subject to change without notice.

Figure 7. Packaging Configuration



TAOS032B - AUGUST 2002

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