

# ECONOMY HIGH-SPEED PWM CONTROLLER

### **FEATURES**

- Peak Current Mode, Average Current Mode, or Voltage Mode (with Feed-Forward) Control Methods
- Practical Operation Up to 1 MHz
- 50-ns Propagation Delay to Output
- ±1.5-A Peak Totem Pole Outputs
- 9-V to 30-V Nominal Operational Voltage Range
- Wide Bandwidth Error Amplifier
- Fully Latched Logic with Double Pulse Suppression
- Pulse-by-Pulse Current Limiting
- Programmable Maximum Duty Cycle Control
- Under-Voltage Lockout with Hysteresis
- Trimmed 5.1-V Reference with UVLO
- Same Functionality as UC3823 and UC3825

### **APPLICATIONS**

- Off-Line and DC/DC Power Supplies
- Converters Using Voltage Mode, Peak Current Mode, or Average Current Mode Control Methods
- Single-Ended or Two-Switch Topology Designs

### DESCRIPTION

The UC28023 and UC28025 are fixed-frequency PWM controllers optimized for high-frequency switched-mode power supply applications. The UC28023 is a single output PWM for single-ended topologies while the UC28025 offers dual alternating outputs for double-ended and full bridge topologies.

Targeted for cost effective solutions with minimal external components, UC2802x include an oscillator, a temperature compensated reference, a wide band width error amplifier, a high-speed current-sense comparator and high-current active-high totem-pole outputs to directly drive external MOSFETs.

Protection circuitry includes a current limit comparator with a 1-V threshold, a TTL compatible shutdown port, and a soft-start pin which will double as a maximum duty cycle clamp. The logic is fully latched to provide jitter free operation and prohibit multiple pulses at an output. An undervoltage lockout section with 800 mV of hysteresis assures low start-up current. During undervoltage lockout, the outputs are high impedance. Particular care was given to minimizing propagation delays through the comparators and logic circuitry while maximizing bandwidth and slew rate of the error amplifier.

Devices are available in the industrial temperature range of -40°C to 105°C. Package offerings are 16-pin SOICW (DW), or 16-pin PDIP (N) packages.

### ORDERING INFORMATION

	OUTPUT	EXTERNAL CURRENT	PACKAGED DEVICES		
$^{I}A = {^{I}J}$	TA = TJ CONFIGURATION LIMIT REFERENCE		PDIP-16 (N)	SOICW-16 (DW)	
4000 to 40500	Single	Yes	UC28023N	UC28023DW	
-40°C to 105°C	Dual Alternating	No	UC28025N	UC28025DW	

<sup>(1)</sup> The DW package are also available taped and reeled. Add an R suffix to the device type (i.e., UC28023DWR (2,000 devices per reel).





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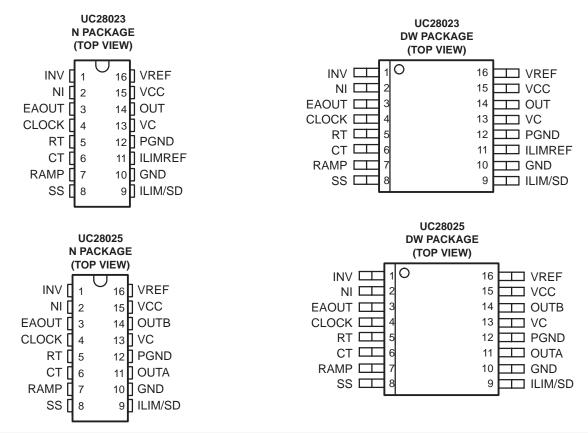
These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range unless otherwise noted<sup>(1)</sup>

	UC28023	UC28025	RATING	UNIT
Input voltage range,	VC, VCC	VC, VCC	30	V
Output current, IOUT(DC)	OUT	OUTA, OUTB	±0.5	А
Peak output current, pulsed 0.5 ms IOUT(pulsed)	OUT	OUTA, OUTB	±2.0	А
Capacitive load, CLOAD			200	pF
A set on transfer	INV, NI, RAMP	INV, NI, RAM	–0.3 V to 7 V	А
Analog inputs	SS, ILIM/SD	SS, ILIM/SD	±2.0	А
Output current, IREF	VREF	VREF	10	
Output current, ICLOCK	CLOCK	CLOCK	-5	
Soft-start sink current, ISINK_SS	SS	SS	5	mA
Output current, IOUT(EA)	EAOUT	EAOUT	20	
Oscillator charging current, IOSC_CHG	RT	RT	-5	
Power Dissipation at $T_A = 25^{\circ}C$ (all packages)	1	W		
Operating junction temperature range, TJ	-55 to 150			
Storage temperature, T <sub>stg</sub>	-65 to 150	°C		
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds, $\mathrm{T}_{\mathrm{SO}}$	300			

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. All voltages are with respect to GND. All currents are positive into and negative out of the specified terminal.





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### **ELECTRICAL CHARACTERISTICS**

 $T_A=-40^\circ C$  to  $105^\circ C$  ,  $T_J=T_{A,}~R_T=3.65~k\Omega,~C_T=1$  nF,  $V_{CC}=15~V$  (unless otherwise noted)

PARAMETER			TEST CONDITIONS		MIN	TYP	MAX	UNIT
REFERENC	CE							
VREF	Reference voltage		T <sub>J</sub> = 25°C,	I <sub>REF</sub> = 1 mA	5.05	5.10	5.15	V
	Line regulation voltage		$10 \text{ V} \le \text{V}_{CC} \le 30$	) V		2	15	
	Load regulation voltage		$1 \text{ mA} \leq I_{REF} \leq 1$	0 mA		5	15	mV
	Temperature stability(1)		$T_{(min)} < T_A < T_C$	(max)		0.2	0.4	mV/°0
	Total output voltage variation(1	)	Line, load, temp	erature	4.95		5.25	V
	Output noise voltage(1)		10 Hz < f < 10 k	Hz		50		μV
	Long term stability voltage <sup>(1)</sup>		TJ = 125°C,	1000 hours		5	25	mV
ISS	Short circuit current		V <sub>REF</sub> = 0 V		-20	-50	-100	mA
OSCILLAT	OR				•			
fosc	Initial accuracy(1)		Tj = 25°C		360	400	440	kHz
	Voltage stability(1)		$10 \text{ V} \le \text{V}_{CC} \le 30$	) V		0.2%	2.0%	
	Temperature stability(1)		$T_{(min)} < T_A < T_($			5%		
	Total voltage variation(1)		Line, temperatur		340		460	kHz
VCLOCK_H	High-level clock output voltage	9			3.9	4.5		
VCLOCK_L	Low-level clock output voltage					2.3	2.9	
VRAMP(p)	Ramp peak voltage <sup>(1)</sup>				2.6	2.8	3.0	V
VRAMP(v)	Ramp valley voltage <sup>(1)</sup>				0.70	1.00	1.25	1
VRAMP(v-p)	Ramp vally-to-peak voltage(1)				1.6	1.8	2.0	
	IPLIFIER				•			
VIN	Input offset voltage						15	mV
BIAS	Input bias current					0.6	3.0	
IN	Input offset current					0.1	1.0	μA
Avol	Open loop gain		$1 V \le V_{OUT} \le 4$	V	60	95		
CMRR	Common mode rejection ratio		$1.5 \text{ V} \le \text{V}_{CM} \le 5$	5.5 V	75	95		dB
PSRR	Power supply rejection ratio		$10 \text{ V} \le \text{V}_{CC} \le 30$	) V	85	110		
OUT(sink)	Output sink current		$V_{(EAOUT)} = 1$	V	1.0	2.5		
IOUT(src)	Output source current		V(EAOUT) = 4		-0.5	-1.3		mA
Voh	High-level output voltage		I(EAOUT) = -0.		4.0	4.7	5.0	
VOL	Low-level output voltage		$I_{(EAOUT)} = 1 \text{ m}$		0	0.5	1.0	V
	Unity gain bandwidth <sup>(1)</sup>				3.0	5.5		MHz
	Slew rate <sup>(1)</sup>				6	12		V/µs
PWM COM	PARATOR							
IBIAS	RAMP bias current		V <sub>RAMP</sub> = 0 V			-1	-5	μA
		UC28023			80%	90%		
	Maximum duty cycle	UC28025	(2)		40%	45%		
	Minimum duty cycle UC28023 UC28025 EAOUT zero DC threshold						0%	
							0%	
			V <sub>RAMP</sub> = 0 V		1.10	1.25	1.40	V
<sup>t</sup> DELAY	Delay to output time <sup>(1)</sup>					50	100	ns

(1) Ensured by design. Not production tested.
(2) Tested as 80% minimum for the oscillator which is the equivalent of 40% for UC28025.



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#### **ELECTRICAL CHARACTERISTICS**

 $T_A = -40^{\circ}C$  to  $105^{\circ}C$ ,  $T_J = T_{A_1}$ ,  $R_T = 3.65 \text{ k}\Omega$ ,  $C_T = 1 \text{ nF}$ ,  $V_{CC} = 15 \text{ V}$  (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SOFT-STA	RT		· · · · ·				
ICHG	Charge current		$V_{SS} = 0.5 V$	3	9	20	μΑ
IDISCHG	Discharge current		V <sub>SS</sub> = 1.0 V	1.0	7.5		mA
CURRENT	LIMIT/SHUTDOWN		· · · · ·				
LIMIT	Current limit bias current		$0 V < V_{(ILIM/SD)} < 4 V$			±10	μΑ
ILIMIT	Offset voltage	UC28023				15	mV
ILIMREF	Common mode range(1)	UC28023		1.00		1.25	
	Current limit threshold voltage	UC28025		0.9	1.0	1.1	V
	Shutdown threshold voltage			1.25	1.40	1.55	
<sup>t</sup> DELAY	Delay to output time <sup>(1)</sup>				50	80	ns
OUTPUT			· · · · ·				
V <sub>OL</sub>	Low-level output voltage		I <sub>OUT</sub> = 20 mA		0.25	0.40	
			I <sub>OUT</sub> = 200 mA		1.2	2.2	v
Maria	High-level output voltage		$I_{OUT} = -20 \text{ mA}$	13.0	13.5		
VOH			$I_{OUT} = -200 \text{ mA}$	12	13		
	Collector leakage		$V_{C} = 30 V$	100	500		μA
	Rise time / Fall time <sup>(1)</sup>		$C_{LOAD} = 1 \text{ nF}$	30	60		ns
UNDERVO	LTAGE LOCKOUT (UVLO)						
	Start threshold voltage			8.8	9.2	9.6	
	Hysteresis			0.4	0.8	1.2	V
SUPPLY C	URRENT						
	Start-up current		$V_{CC} = 8 V$		1.1	2.0	
ICC	Operating current		$V_{INV} = V_{RAMP} = V_{ILIM} = 0 V_{INV} = 1 V$		25	35	mA

(1) Ensured by design. Not production tested.

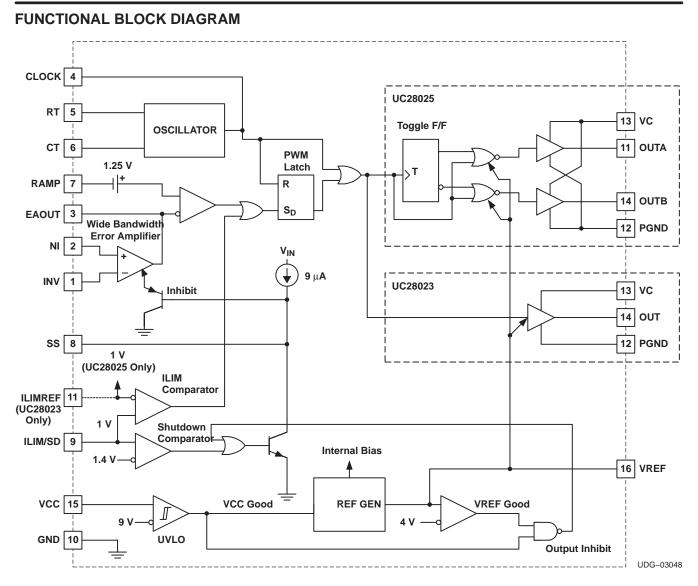
### THERMAL RESISTANCE

PACKAGE	θ <b>JA (°C/W)</b>	θJC (°C/W)	
N(2)	90(2)	45	
DW(2)	50-100(2)	27	

<sup>(2)</sup> Specified  $\theta_{JA}$  (junction-to-ambient) is for devices mountied to 5-square-inch FR4 PC board with one ounce copper where noted. When resistance range is given, lower values are for 5-square-inch aluminum PC board. Test PWB is 0.062 inches thick and typically uses 0.635 mm trace width for power packages and 1.3 mm trace widths for non-power packages with a 100x100 mil probe land area at the end of each trace.



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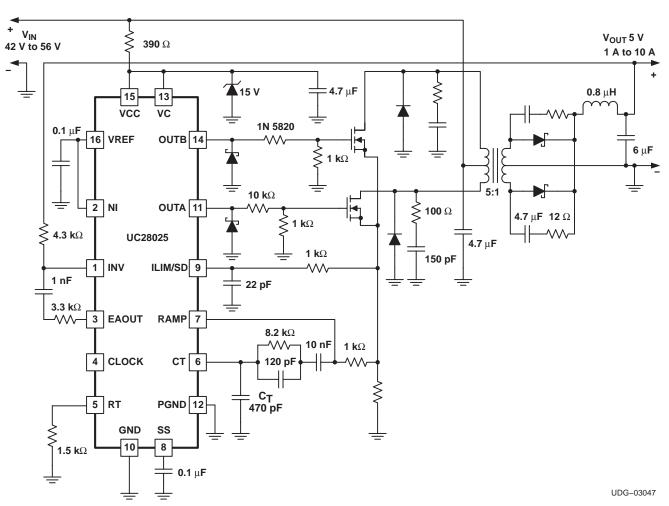
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### **TERMINAL FUNCTIONS**

	TERMINAL					
NAME	UC28023	UC28025	I/O	DESCRIPTION		
CLOCK	4	4	0	Output of the internal oscillator		
СТ	6	6	I	Timing capacitor connection pin for oscillator frequency programming. The timing capacitor sho be connected to the device ground using minimal trace length.		
EAOUT	3	3	0	Output of the error amplifier for compensation		
GND	10	10	-	Analog ground return pin.		
ILIM/SD	9	9	I	Input to the current limit comparator and the shutdown comparator.		
ILIMREF	11	-	I	Pin to set the current limit threshold externally.		
INV	1	1	I	Inverting input to the error amplifier		
NI	2	2	I	Non-inverting input to the error amplifier		
OUT	14	-	0	High current totem pole output of the on-chip drive stage.		
OUTA	-	11	0	High current totem pole output A of the on-chip drive stage.		
OUTB	-	14	0	High current totem pole output B of the on-chip drive stage.		
PGND	12	12	-	Ground return pin for the output driver stage		
RAMP	7	7	I	Non-inverting input to the PWM comparator with 1.25-V internal input offset. In voltage mode operation this serves as the input voltage feed-forward function by using the CT ramp. In peak current mode operation, this serves as the slope compensation input.		
RT	5	5	I	Timing resistor connection pin for oscillator frequency programming		
SS	8	8	I	Soft-start input pin which also doubles as the maximum duty cycle clamp.		
VC	13	13	-	Power supply pin for the output stage. This pin should be bypassed with a $0.1\mathchar`up F$ monolithic ceramic low ESL capacitor with minimal trace lengths.		
VCC	15	15	-	Power supply pin for the device. This pin should be bypassed with a $0.1\mathchar`uF$ monolithic ceramic low ESL capacitor with minimal trace lengths		
VREF	16	16	0	5.1–V reference. For stability, the reference should be bypassed with a 0.1- $\mu$ F monolithic ceramic low ESL capacitor and minimal trace length to the ground plane.		



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APPLICATION INFORMATION

Figure 1. Typical Application: 1.5 MHz, 48-V to 5-V DC/DC Push-Pull Converter Using UC28025



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#### **APPLICATION INFORMATION**

#### PCB LAYOUT CONSIDERATIONS

High speed circuits demand careful attention to layout and component placement. To assure proper performance of the UC2802x follow these rules:

- 1. Use a ground plane.
- 2. Damp or clamp parasitic inductive kick energy from the gate of driven MOSFETs. Do not allow the output pins to ring below ground. A series gate resistor or a shunt 1-A Schottky diode at the output pin serves this purpose.
- Bypass VCC, VC, and VREF. Use 0.1-μF monolithic ceramic capacitors with low equivalent series inductance. Allow less than 1-cm of total lead length for each capacitor between the bypassed pin and the ground plane.
- 4. Treat the timing capacitor, C<sub>T</sub>, as a bypass capacitor.

#### ERROR AMPLIFIER

Figure 2 shows a simplified schematic of the UC2802x error amplifier and Figures 3 and 4 show its characteristics.

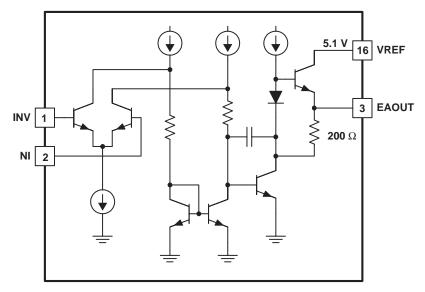
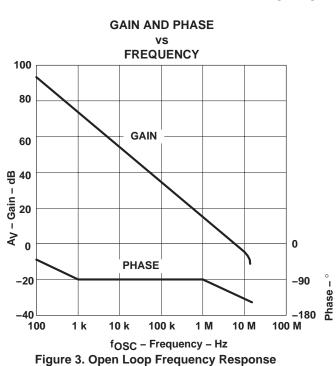
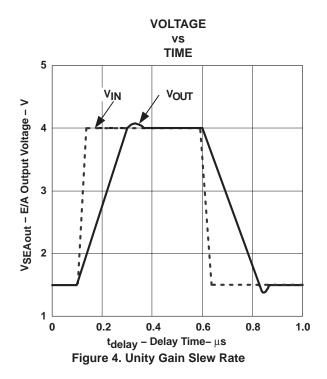


Figure 2. Simplified Error Amplifier Schematic

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**APPLICATION INFORMATION** 



#### **CONTROL METHODS**

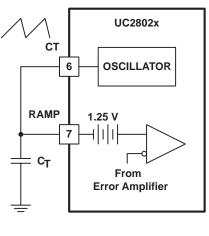
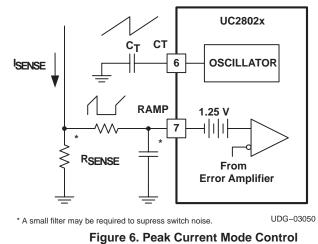


Figure 5. Voltage Mode Control



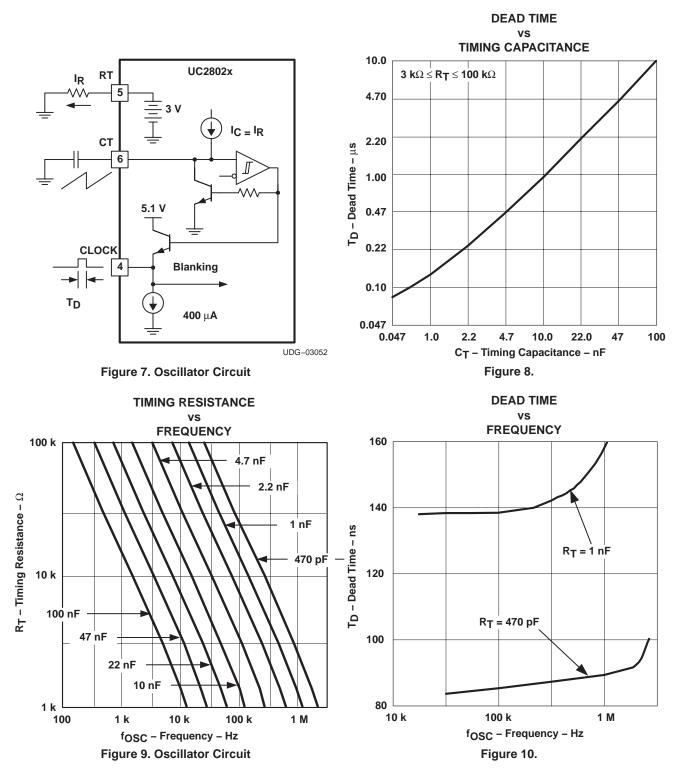
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#### **APPLICATION INFORMATION**

### OSCILLATOR





#### **APPLICATION INFORMATION**

#### **SYNCHRONIZATION**

Figure 11 shows a generalized synchronization. Figure 12 shows a synchronozed operation of two units in close proximity.

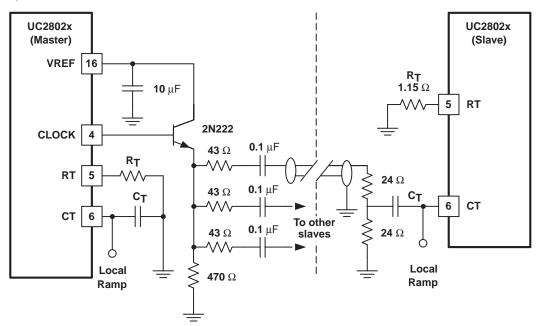


Figure 11. Generalized Synchronization

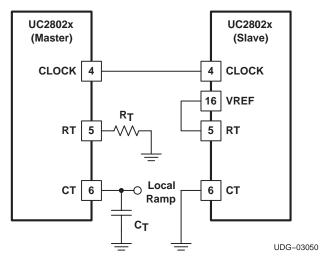


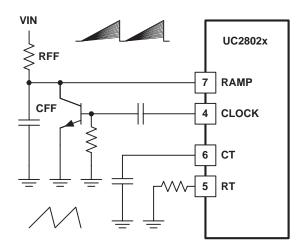
Figure 12. Synchronization of Two Units In Close Proximity



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#### **APPLICATION INFORMATION**

### FEEDFORWARD CIRCUIT



UDG-03050



#### CONSTANT VOLT-SECOND CLAMP CIRCUIT

The circuit for the UC28023 shown in Figure 14 describes achievement a constant volt-second product clamp over varying input voltages. The ramp generator components,  $R_T$  and  $C_R$  are chosen so that the ramp at Pin 9 (ILIM/SD) crosses the 1-V threshold at the same time the desired maximum volt-second product is reached. The delay through the functional inverter block must be such that the ramp capacitor can be completely discharged during the minimum deadtime.

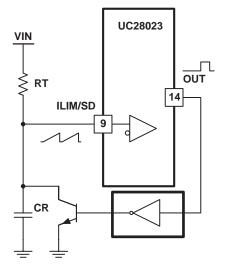


Figure 14. Achieving Constant Volt-Second Product Clamp with the UC28023



UDG-03050

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The circuit for the UC28025 shown in Figure 15 describes achievement a constant volt-second product clamp over varying input voltages. The ramp generator components,  $R_T$  and  $C_R$  are chosen so that the ramp at Pin 9 (ILIM/SD) crosses the 1-V threshold at the same time the desired maximum volt-second product is reached. The delay through the functional inverter block must be such that the ramp capacitor can be completely discharged during the minimum deadtime.

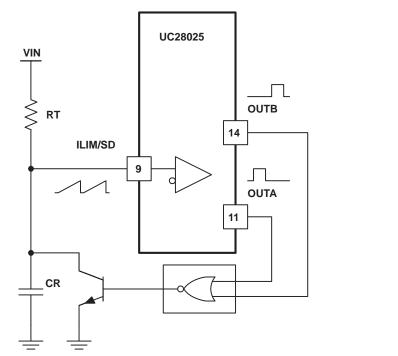


Figure 15. Achieving Constant Volt-Second Product Clamp with the UC28025



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### **APPLICATION INFORMATION**

## OUTPUTS

UC28023 has one output and UC28025 has dual alternating outputs.

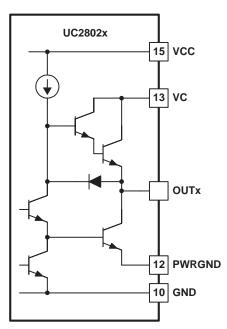
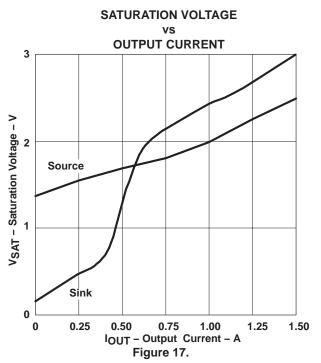
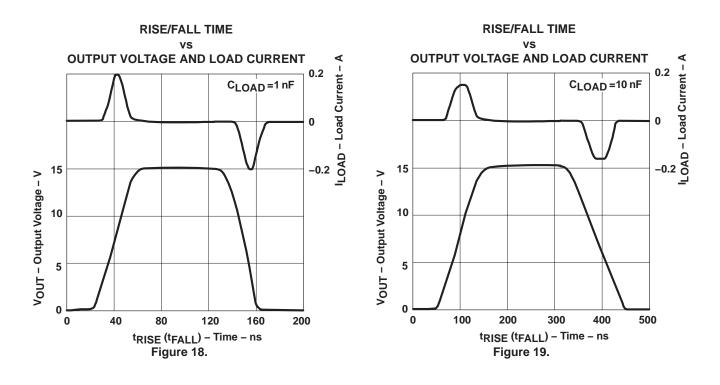


Figure 16. Simplified Schematic







#### **APPLICATION INFORMATION**

#### **Open Loop Laboratory Test Fixture**

The following test fixture is useful for exercising many of the UC28025's functions and measuring their specifications. As with any wideband circuit, careful ground and by-pass procedures should be followed. The use of a ground plane is highly recommended.

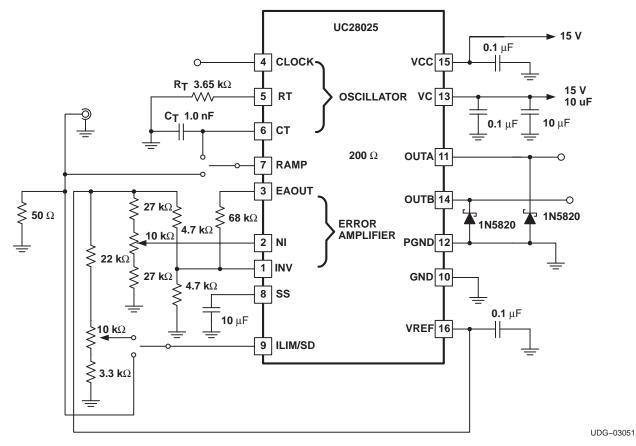


Figure 20. Laboratory Test Fixture

### References

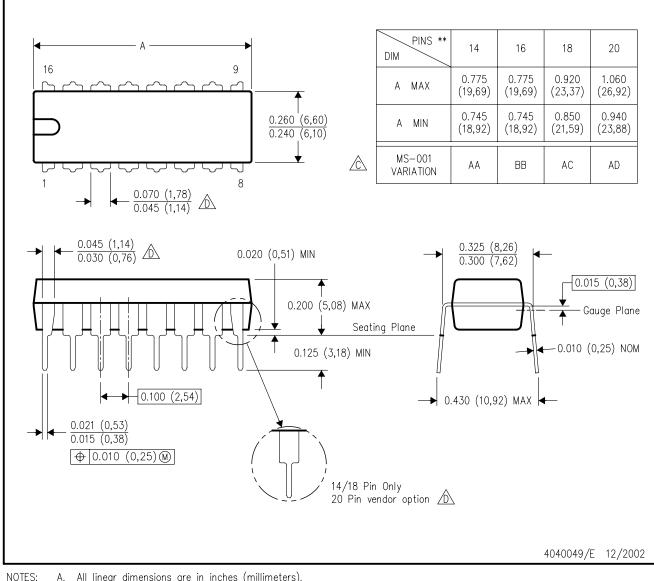
- 1. 1.5-MHz Current Mode IC Controlled 50–Watt Power Supply, Texas Instruments Application Note Literature No. SLUA053.
- 2. The UC3823A,B and UC3825A,B Enhanced Generation of PWM Controllers, Texas Instruments Application Note Literature No. SLUA125.



# N (R-PDIP-T\*\*)

PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN



A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

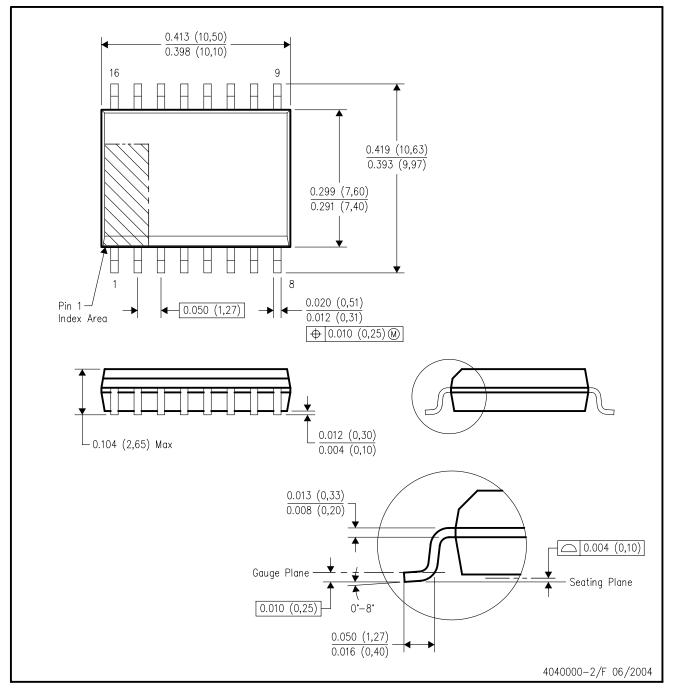
🖄 Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).

The 20 pin end lead shoulder width is a vendor option, either half or full width.



DW (R-PDSO-G16)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).

D. Falls within JEDEC MS-013 variation AA.



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