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### **Upstream CATV Amplifiers**

### **General Description**

The MAX3514/MAX3516/MAX3517 programmable-gain amplifiers are designed for use in CATV upstream applications. The MAX3514/MAX3517 drive up to +61dBmV (QPSK) into a 75 $\Omega$  load when driven with a +34dBmV nominal input signal. The MAX3516 drives up to +64dBmV (QPSK). Both input and output ports are differential, requiring that an external balun be used at the output port. The variable gain feature provides greater than 56dB of dynamic range, which is controlled by an SPI™ 3-wire interface. Gain control is available in 0.5dB steps. The devices operate over a frequency range of 5MHz to 65MHz.

The MAX3514 is a pin-for-pin compatible upgrade for the MAX3510. Like the MAX3510, the MAX3514 is internally matched for use with a 2:1 (voltage ratio) balun. The MAX3517 utilizes an external output resistor for greater load-matching flexibility, and offers the same performance as the MAX3514. The MAX3516 is a higher power version of the MAX3514 with 3dB more gain and output power capability, and is offered in a smaller thermally enhanced TSSOP-EP package.

These devices operate from a single +5VDC supply and draw 120mA during transmit (100% duty cycle, +61dBmV out). The MAX3516 can be operated at up to +9VDC supply for improved harmonic distortion performance. The bias current is automatically adjusted based on the output level to increase efficiency. Additionally, the devices are shut off between bursts to minimize noise and save power while still maintaining a match at the output port. Shutdown mode disables all circuitry and reduces current consumption to 10µA (typ).

The MAX3514/MAX3517 are available in a 20-pin QSOP package and the MAX3516 is available in a 20-pin TSSOP-EP package. All devices operate in the extended industrial temperature range (-40°C to +85°C).

### **Applications**

DOCSIS™/EuroDOCSIS™ and DVB Cable Modems

OpenCable™ Set-Top Box Telephony over Cable

**CATV Status Monitor** 

**VEAXIVI** 

#### Typical Operating Circuit appears at end of data sheet.

SPI is a trademark of Motorola Corp. DOCSIS/EuroDOCSIS/OpenCable are trademarks of CableLabs.

### Features

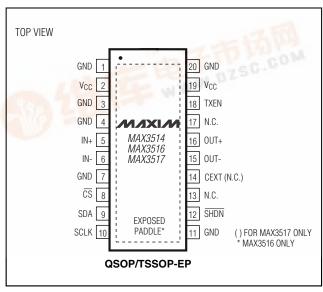
- ♦ Accurate Gain Control
- Gain Programmable in 0.5dB Steps
- ♦ 56dB of Gain Control Range
- ◆ -55dBc Harmonic Distortion at 65MHz Input
- Low Burst On/Off Transient
- ♦ High Efficiency: 35mA at +34dBmV Out 8mA Transmit Disable Mode

### **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE
MAX3514EEP	-40°C to +85°C	20 QSOP
MAX3516EUP	-40°C to +85°C	20 TSSOP-EP*
MAX3517EEP	-40°C to +85°C	20 QSOP

<sup>\*</sup>Exposed paddle

### Pin Configuration



Maxim Integrated Products 1

#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> , OUT+, OUT	0.3V to +10.0V	Operating Te
Input Voltage Levels (all inputs)	0.3V to $(V_{CC} + 0.3V)$	Junction Tem
Continuous Input Voltage (IN+, IN-)	2Vp-p	Storage Tem
Continuous Current (OUT+, OUT-).	120mA	Lead Tempe
Continuous Power Dissipation (TA =	= +70°C)	
20-Pin QSOP (derate 12.3mW/°C	above $T_A = +70^{\circ}C$ )988mW	
20-Pin TSSOP-EP		
(derate $27 \text{mW/}^{\circ}\text{C}$ above $T_{\Delta} = +7$	0°C) 2200mW	

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### DC ELECTRICAL CHARACTERISTICS—MAX3514/MAX3516/MAX3517

(Typical operating circuit;  $V_{CC} = +4.75V$  to +5.25V,  $V_{GND} = 0$ ,  $TXEN = \overline{SHDN} = high$ ,  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ . Typical parameters are at  $V_{CC} = +5V$ ,  $T_A = +25^{\circ}C$ , unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Supply Voltage	Vcc		4.75		5.25	V
Supply Current Transmit Mode	Icc	D7 = 1, gain code = 125 (A <sub>V</sub> = 27dB)		120	150	mA
(MAX3514/MAX3517)	100	D7 = 0, gain code = 100 (A <sub>V</sub> = 0dB)		35		IIIA
Supply Current Transmit Mode	loo	D7 = 1, gain code = 125 (A <sub>V</sub> = 31dB)		160	195	mA
(MAX3516)	Icc	D7 = 0, gain code = 94 (Ay = 0.5dB)		30		IIIA
Supply Current Transmit Disable Mode	Icc	TXEN = low		8	12	mA
Supply Current Low-Power Standby	Icc	SHDN = low		10		μА
LOGIC INPUTS						
Input High Voltage	VINH		2.0			V
Input Low Voltage	V <sub>INL</sub>				0.8	V
Input High Current	IBIASH	$V_{INH} = +3.6V$			100	μA
Input Low Current	IBIASL	VINL = 0	-100			μΑ

### **AC ELECTRICAL CHARACTERISTICS—MAX3514**

(MAX3514 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		D7 = 1, gain code = 125, T <sub>A</sub> = 0°C to +85°C	26.7	27.7	28.7	
		D7 = 1, gain code = 110, T <sub>A</sub> = 0°C to +85°C	19.2	20.2	21.2	
		D7 = 1, gain code = 87, T <sub>A</sub> = 0°C to +85°C	7.7	8.7	9.7	
Voltage Gain, f <sub>IN</sub> = 5MHz	۸۰۰	D7 = 0, gain code = 115, T <sub>A</sub> = 0°C to +85°C	6.7	7.7	8.7	dB
(Note 2)	Ay	D7 = 0, gain code = 100, T <sub>A</sub> = 0°C to +85°C	-0.8	0.2	1.2	иь
		D7 = 0, gain code = 80, T <sub>A</sub> = 0°C to +85°C	-10.8	-9.8	-8.8	
		D7 = 0, gain code = 60, T <sub>A</sub> = 0°C to +85°C	-20.8	-19.8	-18.8	
		D7 = 0, gain code = 48, T <sub>A</sub> = 0°C to +85°C	-27.0	-26.0	-25.0	
Voltage Gain, f <sub>IN</sub> = 65MHz	Av	D7 = 1, gain code = 127, T <sub>A</sub> = -40°C to +85°C; Notes 3, 4	26.3			dB
Coin Dolloff		V <sub>OUT</sub> = 61dBmV, f <sub>IN</sub> = 5MHz to 42MHz (Notes 3, 4)		-0.3	-0.5	٩D
Gain Rolloff		$V_{OUT}$ = 61dBmV, $f_{IN}$ = 5MHz to 65MHz (Notes 3, 4)		-1.0	-1.5	dB
		$f_{IN}$ = 5MHz to 65MHz, $A_V$ = -26dB to +27dB		0.5		
Gain Step Size		$f_{IN}$ = 5MHz to 65MHz, Av = -26dB to +27dB, any 2-bit transition of D0, D1	0.7	1	1.3	dB
		f <sub>IN</sub> = 5MHz to 65MHz, D7 = 0, gain code = 115; to D7 = 1, gain code = 87	0.7	1.0	1.3	
Transmit-Disable Mode Noise		TXEN = low, BW = 160kHz, f <sub>IN</sub> = 5MHz to 65MHz; Note 3			-71	dBmV
Isolation in Transmit-Disable Mode		TXEN = low, f <sub>IN</sub> = 5MHz to 65MHz (Note 3)	60			dB
Transmit Mode Noise		BW = 160kHz, $f_{IN}$ = 5MHz to 65MHz, A <sub>V</sub> = -26dB to +27dB; Note 3			-59	dBc
Transmit Enable Transient Duration		TXEN input rise/fall time < 0.1µs, T <sub>A</sub> = +25°C (Note 3)			2	μs
Transmit Disable Transient Duration		TXEN input rise/fall time < 0.1µs, T <sub>A</sub> = +25°C (Note 3)			2	μs
Transmit Disable/Transmit Enable		D7 = 1, gain code = 125 (A <sub>V</sub> = 27dB), T <sub>A</sub> = +25°C		30	100	ma\ /
Transient Step Size		D7 = 0, gain code = 100 (A <sub>V</sub> = 0.2dB), T <sub>A</sub> = +25°C		1		mVp-p
Input Impedance	Z <sub>IN</sub>	f <sub>IN</sub> = 5MHz to 65MHz, single ended; Note 3	1	1.5		kΩ
Output Return Loss		$f_{IN}$ = 5MHz to 42MHz in 75 $\Omega$ system, D7 = 1 gain code = 125 (Av = 27dB) (Note 4)		10		dB



### AC ELECTRICAL CHARACTERISTICS—MAX3514 (continued)

(MAX3514 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Return Loss in Transmit- Disable Mode		$f_{\text{IN}}$ = 5MHz to 42MHz, in 75 $\Omega$ system, TXEN = low; Note 4		10		dB
Two Tone Third Order Distortion	IMO	Input tones at 42MHz and 42.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone; Note 3		-53	-47	alD o
Two-Tone Third-Order Distortion	IM3	Input tones at 65MHz and 65.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone		-49		dBc
and Harmonia Distortion	HD0	f <sub>IN</sub> = 33MHz, V <sub>OUT</sub> = +61dBmV; Note 3		-55	-53	dDo
2nd-Harmonic Distortion	HD2	f <sub>IN</sub> = 65MHz, V <sub>OUT</sub> = +61dBmV; Note 3		-55	-52	dBc
3rd-Harmonic Distortion	HD3	f <sub>IN</sub> = 22MHz, V <sub>OUT</sub> = +61dBmV	-55	-55	-50.5	dBc
Sid-Harmonic Distortion	מחט	$f_{IN} = 65MHz$ , $V_{OUT} = +61dBmV$		-55	-50.5	UDC
AMA AM	0.04/0.04	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 42MHz		0.1		۵D
AM to AM	AM/AM	$A_V = 27$ dB, $V_{IN} = +34$ dBmV to $+38$ dBmV, $f_{IN} = 65$ MHz		0.1		dB
AM to PM		$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 42MHz		1		dograda
	AM/PM	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 65MHz		1		degrees

### **AC ELECTRICAL CHARACTERISTICS—MAX3516**

(MAX3516 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		D7 = 1, gain code = 125, T <sub>A</sub> = 0°C to +85°C	30	31	32	
		D7 = 1, gain code = 119, T <sub>A</sub> = 0°C to +85°C	27	28	29	
		D7 = 1, gain code = 104, T <sub>A</sub> = 0°C to +85°C	19.5	20.5	21.5	
		D7 = 1, gain code = 81, T <sub>A</sub> = 0°C to +85°C	8	9	10	
Voltage Gain, f <sub>IN</sub> = 5MHz (Note 2)	A <sub>V</sub>	D7 = 0, gain code = 109, T <sub>A</sub> = 0°C to +85°C	7	8	9	dB
		D7 = 0, gain code = 94, T <sub>A</sub> = 0°C to +85°C	-0.5	0.5	1.5	
		D7 = 0, gain code = 74, T <sub>A</sub> = 0°C to +85°C	-10.5	-9.5	-8.5	
		D7 = 0, gain code = 54, T <sub>A</sub> = 0°C to +85°C	-20.5	-19.5	-18.5	
		D7 = 0, gain code = 42, T <sub>A</sub> = 0°C to +85°C	-26.5	-25.5	-24.5	
Voltage Gain, f <sub>IN</sub> = 65MHz	Av	D7 = 1, gain code = 127, T <sub>A</sub> = -40°C to +85°C (Notes 3, 4)	28.1			dB
Gain Rolloff		V <sub>OUT</sub> = 64dBmV, f <sub>IN</sub> = 5MHz to 42MHz (Notes 3, 4)		-0.3	-0.6	dB
Galli holloli		V <sub>OUT</sub> = 64dBmV, f <sub>IN</sub> = 5MHz to 65MHz (Notes 3, 4)		-1.1	-1.7	uв
		$f_{IN}$ = 5MHz to 65MHz, Av = -26dB to +30dB		0.5		
Gain Step Size		$f_{IN}$ = 5MHz to 65MHz, Ay = -26dB to +30dB, any 2-bit transition of D0, D1	0.7	1.0	1.3	dB
		f <sub>IN</sub> = 5MHz to 42MHz, Ay = -26dB to +30dB, D7 = 0, gain code = 109; to D7 = 1, gain code = 81	0.7	1.0	1.3	
Transmit-Disable Mode Noise		TXEN = low, BW = 160kHz, f <sub>IN</sub> = 5MHz to 65MHz			-71	dBmV
Isolation in Transmit-Disable Mode		TXEN = low, f <sub>IN</sub> = 5MHz to 65MHz (Note 3)	60			dB
Transmit Mode Noise		$BW = 160 \text{kHz},  f_{\text{IN}} = 5 \text{MHz to } 65 \text{MHz}, \\ A_{\text{V}} = -26 \text{dB to } 27 \text{dB (Note 3)}$			-59	dBc



### AC ELECTRICAL CHARACTERISTICS—MAX3516 (continued)

(MAX3516 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Transmit Enable Transient Duration		TXEN input rise/fall time < 0.1µs, T <sub>A</sub> = +25°C (Note 3)			2	μs
Transmit Disable Transient Duration		TXEN input rise/fall time < 0.1µs, T <sub>A</sub> = +25°C (Note 3)			2	μs
Transmit Disable/Transmit Enable		D7 = 1, gain code = 119, (A <sub>V</sub> = 28dB), T <sub>A</sub> = +25°C		30	100	ma)/m m
Transient Step Size		D7 = 0, gain code = 94, (A <sub>V</sub> = 0.5 dB), T <sub>A</sub> = +25°C		1		mVp-p
Input Impedance	Z <sub>IN</sub>	f <sub>IN</sub> = 5MHz to 65MHz, single-ended (Note 3)	1	1.5		kΩ
Output Return Loss		$f_{IN}$ = 5MHz to 65MHz in 75 $\Omega$ system D7 = 1, gain code = 125, (A <sub>V</sub> = 31dB) (Note 4)		10		dB
Output Return Loss in Transmit- Disable Mode		$f_{\text{IN}}$ = 5MHz to 65MHz in 75 $\Omega$ system, TXEN = low (Note 4)		10		dB
Two-Tone Third-Order Distortion	11.40	Input tones at 42MHz and 42.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone		-53.5		10
(Note 3)	IM3	Input tones at 65MHz and 65.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone		-48.8		dBc
		f <sub>IN</sub> = 33MHz, V <sub>OUT</sub> = +61dBmV		-55	-53	
2nd-Harmonic Distortion (Note 3)	HD2	f <sub>IN</sub> = 33MHz, V <sub>OUT</sub> = +64dBmV		-55		dBc
(14010-0)		f <sub>IN</sub> = 65MHz, V <sub>OUT</sub> = +61dBmV		-55	-52	
		f <sub>IN</sub> = 22MHz, V <sub>OUT</sub> = +61dBmV		-55	-50.5	
3rd-Harmonic Distortion	HD3	$f_{IN} = 22MHz$ , $V_{OUT} = +64dBmV$		-50		dBc
		$f_{IN} = 65MHz$ , $V_{OUT} = +61dBmV$		-55	-50.5	
AM to AM	AM/AM	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 42MHz		0.1		dB
AM to AM	AM/AM	A <sub>V</sub> = 27dB, V <sub>IN</sub> = +34dBmV to +38dBmV, f <sub>IN</sub> = 65MHz		0.1		dB
AM to PM	AM/PM	A <sub>V</sub> = 27dB, V <sub>IN</sub> = +34dBmV to +38dBmV, f <sub>IN</sub> = 42MHz		1		degrees
AM to PM	AM/PM	A <sub>V</sub> = 27dB, V <sub>IN</sub> = +34dBmV to +38dBmV, f <sub>IN</sub> = 65MHz		1		degrees

### **AC ELECTRICAL CHARACTERISTICS—MAX3517**

(MAX3517 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		D7 = 1, gain code = 125, T <sub>A</sub> = 0°C to +85°C	26.7	27.7	28.7	
		D7 = 1, gain code = 110, T <sub>A</sub> = 0°C to +85°C	19.2	20.2	21.2	
		D7 = 1, gain code = 90, T <sub>A</sub> = 0°C to +85°C	9.2	10.2	11.2	
		D7 = 1, gain code = 70, T <sub>A</sub> = 0°C to +85°C	-0.8	0.2	1.2	
Voltage Gain, f <sub>IN</sub> = 5MHz	Av	D7 = 1, gain code = 115, T <sub>A</sub> = 0°C to +85°C	6.7	7.7	8.7	dB
		D7 = 1, gain code = 100, T <sub>A</sub> = 0°C to +85°C	-0.8	0.2	1.2	
		D7 = 1, gain code = 80, T <sub>A</sub> = 0°C to +85°C	-10.8	-9.8	-8.8	
		D7 = 0, gain code = 60, T <sub>A</sub> = 0°C to +85°C	-20.8	-19.8	-18.8	
		D7 = 0, gain code = 48, T <sub>A</sub> = 0°C to +85°C	-27.0	-26.0	-25.0	
Gain Step Size		$f_{IN}$ = 5MHz to 65MHz, A <sub>V</sub> = -26dB to +27dB		0.5		dB
Transmit-Disable Mode Noise		TXEN = low, BW = 160kHz, f <sub>IN</sub> = 5MHz to 65MHz		-71		dBmV
Isolation in Transmit-Disable Mode		TXEN = low, f <sub>IN</sub> = 5MHz to 65MHz	50	58		dB
Transmit Mode Noise		$BW = 160 \text{kHz},  f_{\text{IN}} = 5 \text{MHz}  \text{to}  65 \text{MHz}, \\ A_{\text{V}} = -26 \text{dB}  \text{to}  +27 \text{dB};  \text{Note}  3$		-60	-59	dBc
Transmit Enable Transient Duration		TXEN input rise/fall time < 0.1µs, T <sub>A</sub> = +25°C; Note 3			2	μs

### AC ELECTRICAL CHARACTERISTICS—MAX3517 (continued)

(MAX3517 EV kit;  $V_{CC}$  = +4.75V to +5.25V,  $V_{GND}$  = 0,  $P_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $T_A$  = -40°C to +85°C. Typical parameters are at  $V_{CC}$  = +5V,  $T_A$  = +25°C, unless otherwise specified.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Transmit Disable Transient Duration		TXEN input rise/fall time $< 0.1 \mu s$ , $T_A = +25^{\circ}C$			2	μs
Transmit Disable/Transmit Enable		D7 = 1, gain code = 125, (A <sub>V</sub> = 27dB), T <sub>A</sub> = +25°C		30	100	m\/n n
Transient Step Size		D7 = 0, gain code = 100, (A <sub>V</sub> = 0.2 dB), T <sub>A</sub> = +25°C		1		mVp-p
Input Impedance	Z <sub>IN</sub>	f <sub>IN</sub> = 5MHz to 65MHz, single ended; Note 3	1	1.5		kΩ
Output Return Loss		$f_{IN}$ = 5MHz to 65MHz in 75 $\Omega$ system D7 = 1, gain code = 125, (Ay = 27dB); Note 4		8.3		dB
Output Return Loss in Transmit- Disable Mode		$f_{\text{IN}} = 42 \text{MHz}$ , in 75 $\Omega$ system TXEN = low; Note 4		10.5		dB
Two-Tone Third-Order Distortion	IM3	Input tones at 42MHz and 42.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone		-49.5		dBc
(Note 2)	IIVIS	Input tones at 65MHz and 65.2MHz, both +31dBmV, V <sub>OUT</sub> = +58dBmV/tone		-46.3		авс
2nd-Harmonic Distortion	HD2	$f_{IN} = 33MHz$ , $V_{OUT} = +61dBmV$		-55		dBc
2110-1 IaiTHOTHC Distortion	TIDZ	$f_{IN} = 65MHz$ , $V_{OUT} = +61dBmV$		-55		abc
3rd-Harmonic Distortion	HD3	$f_{IN} = 22MHz$ , $V_{OUT} = +61dBmV$		-55		dBc
ord Flatmonic Biotortion	1100	$f_{IN} = 65MHz$ , $V_{OUT} = +61dBmV$		-55		аво
AM to AM	AM/AM	$A_V = 27$ dB, $V_{IN} = +34$ dBmV to +38dBmV, $f_{IN} = 42$ MHz		0.1		dB
AM to AM	AM/AM	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 65MHz		0.1		dB
AM to PM	AM/PM	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 42MHz		1		degrees
AM to PM	AM/PM	$A_V$ = 27dB, $V_{IN}$ = +34dBmV to +38dBmV, $f_{IN}$ = 65MHz		1		degrees

#### TIMING CHARACTERISTICS

(VCC = 4.75V to 5.25V, V<sub>GND</sub> = 0, TXEN = SHDN = high, D7 = X, T<sub>A</sub> = +25°C, unless otherwise specified.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SEN to SCLK Setup Time	tsens		20			ns
SEN to SCLK Hold Time	tsenh		10			ns
SDA to SCLK Setup Time	tsdas		10			ns
SDA to SCLK Hold Time	tsdah		20			ns
SDA Pulse Width High	T <sub>DATAH</sub>		50			ns
SDA Pulse Width Low	T <sub>DATAL</sub>		50			ns
SCLK Pulse Width High	tsclkh		50			ns
SCLK Pulse Width Low	tsclkl		50			ns

Note 1: Guaranteed by design and characterization to ±3 sigma for T<sub>A</sub> < +25°C, unless otherwise specified.

Note 2: AC Gain correlated to DC Gain measurements to ±3 sigma.

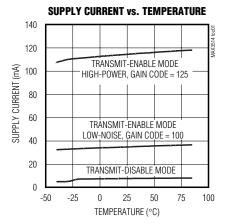
Note 3: Guaranteed by design and characterization to ±6 sigma.

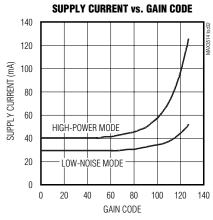
Note 4: Does not include output matching; see Output Match in the Applications section.

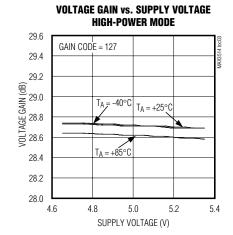
### Typical Operating Characteristics

(Typical operating circuit;  $V_{CC}$  = +5V,  $V_{IN}$  = +34dBmV, TXEN =  $\overline{SHDN}$  = high,  $f_{IN}$  = 20MHz,  $Z_{LOAD}$  = 75 $\Omega$ ,  $T_A$  = +25°C, unless otherwise noted.)

#### MAX3514/MAX3517







### **Typical Operating Characteristics (continued)**

(Typical operating circuit;  $V_{CC} = +5V$ ,  $V_{IN} = +34dBmV$ ,  $TXEN = \overline{SHDN} = high$ ,  $f_{IN} = 20MHz$ ,  $Z_{LOAD} = 75\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

#### MAX3514/MAX3517 (continued) **VOLTAGE GAIN vs. TEMPERATURE VOLTAGE GAIN vs. TEMPERATURE VOLTAGE GAIN vs. SUPPLY VOLTAGE LOW-NOISE MODE LOW-NOISE MODE HIGH-POWER MODE** -9.0 292 -25.0 GAIN CODE = 80 GAIN CODE = 48 GAIN CODE = 127 -9.2 -25.2 29.1 -9.4 -25.4 29.0 T<sub>A</sub> = -40°C (qp) N-25.8 -26.0 -26.2 -26.4 -9.6 28.9 $V_{CC} = +5.00V$ $V_{CC} = +5.25V$ **/OLTAGE GAIN (dB)** VOLTAGE GAIN (dB) -9.8 28.8 V<sub>CC</sub> = +4.75V -10.0 28.7 -10.2 28.6 $V_{CC} = +5.25V$ = +5.00V T<sub>A</sub> = +25°C V<sub>CC</sub> : -10.4 -26.4 28.5 -10.6 -26.6 28.4 T<sub>A</sub> = +85°C -10.8 -26.8 28.3 -27.0 -11.0 28.2 -40 85 4.6 5.0 -40 -15 10 35 60 85 35 TEMPERATURE (°C) SUPPLY VOLTAGE (V) TEMPERATURE (°C) **VOLTAGE GAIN vs. FREQUENCY VOLTAGE GAIN vs. FREQUENCY VOLTAGE GAIN vs. GAIN CODE HIGH-POWER MODE LOW-NOISE MODE** 30 40 25 20 20 30 10 15 HIGH-POWER MODE В 10 20 (gB) VOLTAGE GAIN (dB) VOLTAGE GAIN (dB) 5 -10 VOLTAGE GAIN 0 10 D -5 -20 0 -10 -30 LOW- NOISE MODE -15 GAIN CODE: -40 -10 -20 A = 125-50 GAIN CODE B = 100 -30 -20 A = 120, B = 115, C = 95 C = 87-35 D = 75, E = 55, F = 35D = 60-30 -70 10 1000 30 40 50 60 70 80 90 100 110 120 100 10 100 1000 FREQUENCY (MHz) FREQUENCY (MHz) GAIN CODE **GAIN STEP vs. GAIN CODE GAIN STEP vs. GAIN CODE LOW-NOISE MODE** TRANSMIT NOISE vs. GAIN CODE **HIGH-POWER MODE** 1.0 -20 1.0 0.9 0.9 -25 OUTPUT NOISE (dBmV IN 160kHz) 0.8 8.0 HIGH-POWER MODE 0.7 -30 0.7 GAIN STEP (dB) SAIN STEP (dB) 0.6 0.6 -35 0.5 0.5 -40 0.4 0.4 0.3 0.3 -45 0.2 LOW-NOISE MODE 0.2 -50 0.1 0.1 0 -55 30 40 50 60 70 80 90 100 110 120 60 70 80 90 100 110 30 50 70 90 110 130

GAIN CODE

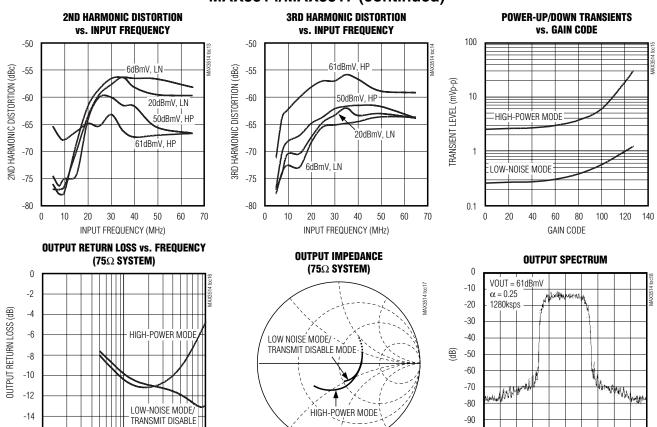
GAIN CODE

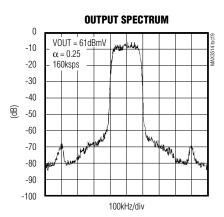
GAIN CODE

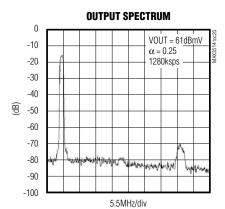
### **Typical Operating Characteristics (continued)**

(Typical operating circuit;  $V_{CC} = +5V$ ,  $V_{IN} = +34dBmV$ , TXEN =  $\overline{SHDN} = high$ ,  $f_{IN} = 20MHz$ ,  $Z_{LOAD} = 75\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

### MAX3514/MAX3517 (continued)







-100

500kHz/div

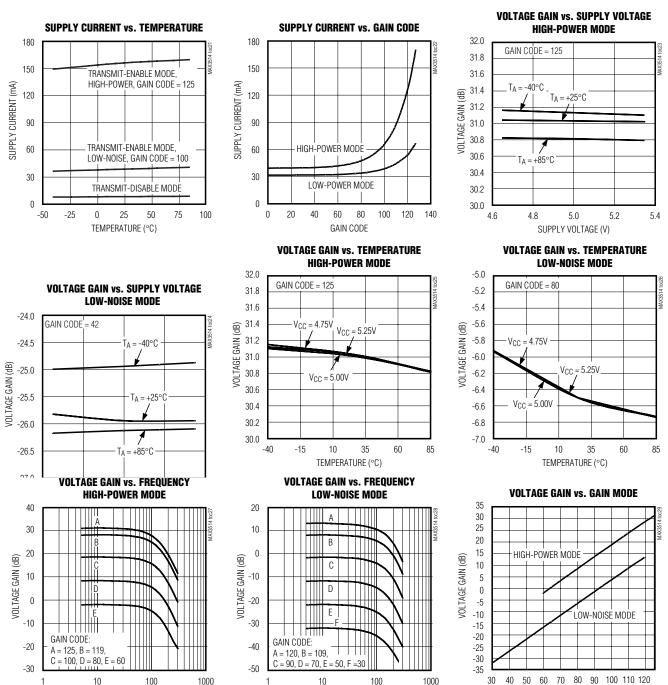
-16

FREQUENCY (MHz)

### **Typical Operating Characteristics (continued)**

(Typical operating circuit;  $V_{CC} = +5V$ ,  $V_{IN} = +34dBmV$ ,  $TXEN = \overline{SHDN} = high$ ,  $f_{IN} = 20MHz$ ,  $Z_{LOAD} = 75\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

#### **MAX3516**



FREQUENCY (MHz)

GAIN CODE

FREQUENCY (MHz)

### Typical Operating Characteristics (continued)

(Typical operating circuit;  $V_{CC} = +5V$ ,  $V_{IN} = +34dBmV$ , TXEN =  $\overline{SHDN} = high$ ,  $f_{IN} = 20MHz$ ,  $Z_{LOAD} = 75\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

MAX3516 (continued)

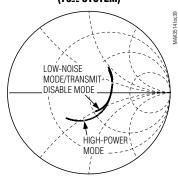
#### **GAIN STEP vs. GAIN CODE GAIN STEP vs. GAIN MODE HIGH-POWER MODE LOW-NOISE MODE** TRANSMIT NOISE vs. GAIN CODE 1.0 1.0 -20 0.9 0.9 -25 HIGH-POWER MODE OUTPUT NOISE (dBmV IN 160kHz) 8.0 8.0 0.7 -30 0.7 GAIN STEP (dB) 0.6 0.6 -35 GAIN STEP 0.5 0.5 -40 0.4 0.4 0.3 0.3 -45 LOW-NOISE MODE 0.2 0.2 -50 0.1 0.1 0 0 -55 100 80 100 110 120 60 70 80 90 40 50 70 90 30 50 70 90 110 130 GAIN CODE GAIN CODE GAIN CODE 2ND HARMONIC DISTORTION **3RD HARMONIC DISTORTION** 2ND HARMONIC DISTORTION vs. INPUT FREQUENCY, V<sub>CC</sub> = 5.0V vs. INPUT FREQUENCY, V<sub>CC</sub> = 5.0V vs. INPUT FREQUENCY, V<sub>CC</sub> = 7.0V -50 -45 -50 64dBmV, HP 20dBmV, LN 20dBmV, LN -50 -55 2ND HARMONIC DISTORTION (dBc) 3RD HARMONIC DISTORTION (dBc) 2ND HARMONIC DISTORTION (dBc) -55 50dBmV, HP 61dBmV, HP -60 -55 50dBmV, HP -60 50dBmV, HP -60 -65 61dBmV, HP -65 61dBmV, HP -70 -65 -70 -75 -70 64dBmV, HP -75 -80 -75 64dBmV, HP -85 -80 -80 30 40 30 INPUT FREQUENCY (MHz) INPUT FREQUENCY (MHz) INPUT FREQUENCY (MHz) 3RD HARMONIC DISTORTION **POWER-UP/DOWN TRANSIENTS OUTPUT RETURN LOSS vs. FREQUENCY** vs. GAIN CODE vs. INPUT FREQUENCY, $V_{CC} = 7.0V$ (75 $\Omega$ SYSTEM) 100 -50 64dBmV, HP -2 -55 3RD HARMONIC DISTORTION (dBc) -4 (mVp-p) OUTPUT RETURN LOSS (dB) 61dBmV, HP -60 10 50dBmV, HP HIGH-POWER MODE -6 HIGH-POWER MODE 20dBmV, LN -65 -8 -10 -70 -12 LOW-NOISE MODE LOW-NOISE MODE/ -14 TRANSMIT-DISABLE MODE -80 -16 0 20 30 40 60 20 80 100 120 10 100 INPUT FREQUENCY (MHz) GAIN CODE FREQUENCY (MHz)

### Typical Operating Characteristics (continued)

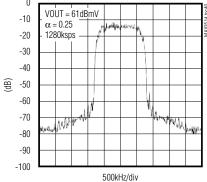
(Typical operating circuit;  $V_{CC} = +5V$ ,  $V_{IN} = +34dBmV$ , TXEN =  $\overline{SHDN} = high$ ,  $f_{IN} = 20MHz$ ,  $Z_{LOAD} = 75\Omega$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

### MAX3516 (continued)

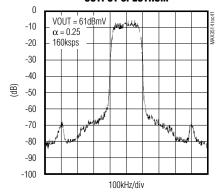
### OUTPUT IMPEDANCE, 5MHz-65MHz (75 $\Omega$ SYSTEM)



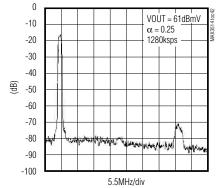
### OUTPUT SPECTRUM



#### OUTPUT SPECTRUM



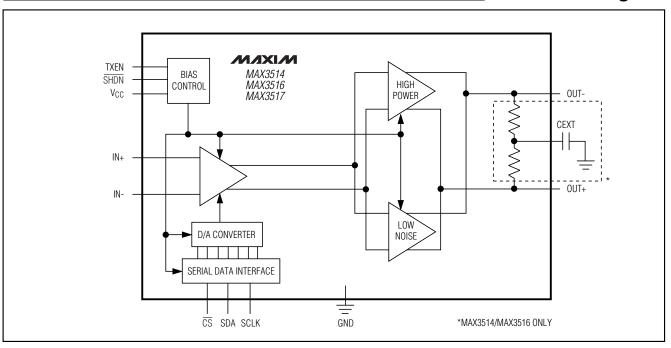
#### OUTPUT SPECTRUM



### Pin Description

PIN	NAME	FUNCTION
1, 3, 7, 11	GND	Ground
2	Vcc	Programmable-Gain Amplifier (PGA) +5V Supply. Bypass to pin 4 with a decoupling capacitor as close to the part as possible.
4	GND	PGA RF Ground. As with all ground connections, maintain the shortest possible (low-inductance) length to the ground plane.
5	IN+	Positive PGA Input. Along with IN-, this port forms a high-impedance differential input to the PGA. Driving this port differentially increases the rejection of second-order distortion at low output levels.
6	IN-	Negative PGA Input. When not used, this port must be AC-coupled to ground. See IN+.
8	CS	Serial-Interface Enable. TTL-compatible input. See Serial Interface section.
9	SDA	Serial-Interface Data. TTL-compatible input. See Serial Interface section.
10	SCLK	Serial-Interface Clock. TTL-compatible input. See Serial Interface section.
12	SHDN	Shutdown. When SHDN is set low, all functions (including the serial interface) are disabled.
13, 17	N.C.	No Connection
14	CEXT	RF Output Bypass. Bypass to ground with a 0.1µF capacitor. (N.C. for MAX3517.)
15	OUT-	Negative Output. Along with OUT+, this port forms a $300\Omega$ impedance output. This port is matched to a $75\Omega$ load using a 2:1 (voltage ratio) transformer.
16	OUT+	Positive Output. See OUT
18	TXEN	Transmit Enable. Drive TXEN high to place the device in transmit-enable mode.
19	Vcc	Output Amplifier Bias, +5V Supply. Bypass to pin 20 with a decoupling capacitor as close to the part as possible.
20	GND	Output Amplifier Bias Ground. As with all ground connections, maintain the shortest possible (low-inductance) length to the ground plane.
Exposed Paddle	GND	Ground (MAX3516 only)

### **Functional Diagram**



### **Detailed Description**

#### **Programmable-Gain Amplifier**

The PGA consists of the variable-gain amplifier (VGA) and the digital-to-analog converter (DAC), which provide better than 56dB of output level control in 0.5dB steps. The PGA is implemented as a programmable Gilbert-cell attenuator. The gain of the PGA is determined by a 7-bit word (D6–D0) programmed through the serial data interface (Tables 1 and 2).

Specified performance is achieved when the input is driven differentially. The device may be driven single ended. To drive the device in this manner, one of the input pins must be capacitively coupled to ground. Use a capacitor value large enough to allow for a low-impedance path to ground at the lowest frequency of operation. For operation down to 5MHz, a  $0.001\mu F$  capacitor is suggested.

#### **Output Amplifiers**

The output amplifiers are Class A differential amplifiers capable of driving +61dBmV (QPSK, MAX3514) differentially. This architecture provides superior even-order distortion performance but requires that a transformer be used to convert to a single-ended output. In transmit-disable mode, the output amplifiers are powered

down. A resistor is across the output so that the output impedance remains matched when the amplifier is in transmit-disable mode. Disabling the output devices also results in low output noise.

#### MAX3514/MAX3516

To match the output impedance to a  $75\Omega$  load, the transformer must have a turns ratio (voltage ratio) of 2:1 (4:1 impedance ratio). The differential amplifier is biased directly from the +5V supply using the center tap of the output transformer. This provides a significant benefit when switching between transmit mode and transmit-disable mode. Stored energy due to bias currents will cancel within the transformer and prevent switching transients from reaching the load.

### MAX3517

The MAX3517 uses external matching resistors to allow matching to various load impedances through suitable values of matching resistors and transformer turns ratios.

#### Serial Interface

The serial interface has an active-low enable  $(\overline{CS})$  to bracket the data, with data clocked in MSB first on the rising edge of SCLK. Data is stored in the storage latch on the rising edge of  $\overline{CS}$ . The serial interface controls

the state of the PGA and the output amplifiers. Tables 1 and 2 show the register format. Serial-interface timing is shown in Figure 1.

### Applications Information

### **High-Power and Low-Noise Modes**

The MAX3514/MAX3516/MAX3517 have two transmit modes, high power (HP) and low noise (LN). Each of these modes is actuated by the high-order bit D7 of the 8-bit programming word. When D7 is a logic 1, HP mode is enabled. When D7 is a logic 0, LN mode is enabled.

Each of these modes is characterized by the activation of a distinct output stage. In HP mode, the output stage exhibits 15dB higher gain than LN mode. The lower gain of the LN output stage allows for significantly lower

**Table 1. Serial-Interface Control Word** 

BIT	MNEMONIC	DESCRIPTION		
MSB 7	D7	High-power/low-noise mode select		
6	D6	Gain code, bit 6		
5	D5	Gain code, bit 5		
4	D4	Gain code, bit 4		
3	D3	Gain code, bit 3		
2	D2	Gain code, bit 2		
1	D1	Gain code, bit 1		
LSB 0	D0	Gain code, bit 0		

output noise and lower transmit/transmit-disable transients.

The full range of gain codes (D6–D0) may be used in either mode. For DOCSIS applications, HP mode is recommended for output levels at or above +42dBmV (MAX3514, D7 = 1, gain code = 87), LN mode when the output level is below +42dBmV (MAX3514, D7 = 0, gain code = 115).

#### **Shutdown Mode**

In normal operation, the shutdown pin (SHDN) is held high. When SHDN is taken low, all circuits within the IC are disabled. Only leakage currents flow in this state. Data stored within the serial-data interface latches will

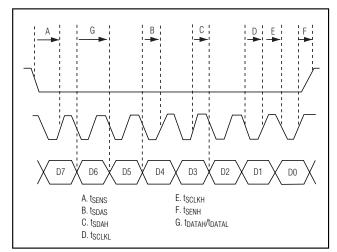


Figure 1. Serial-Interface Timing Diagram

### **Table 2. Chip State Control Bits**

SHDN	TXEN	D7	D6	D5	D4	D3	D2	D1	D0	GAIN CODE (DECIMAL)	GAIN* (DB)	STATES
0	Χ	Χ	Χ	Χ	Χ	Χ	Χ	Χ	Χ			Shutdown Mode
1	0	Χ	Χ	Χ	Χ	Χ	Χ	Χ	Χ			Transmit-Disable Mode
1	1	1	Χ	Χ	Χ	Χ	Χ	Χ	Χ			Transmit-Enable Mode, High Power
1	1	0	Χ	Χ	Χ	Χ	Χ	Χ	Χ			Transmit- Enable Mode, Low Noise
1	1	0	0	1	1	0	0	0	0	48	-26	
1	1	0	1	0	1	0	0	0	0	80	-10	
1	1	0	1	1	1	0	0	1	1	115	8	
1	1	1	1	0	1	0	1	1	1	87	9	
1	1	1	1	1	0	1	1	1	0	110	20	
1	1	1	1	1	1	1	1	0	1	125	28	

<sup>\*</sup>Typical Gain at +25°C, Vcc = 5.0V

be lost upon entering this mode. Current consumption is reduced to 10µA (typ) in shutdown mode.

### Output Match MAX3514/MAX3516

When used in conjunction with a 2:1 voltage-ratio transformer, the MAX3514/MAX3516 are internally resistively matched to 75 $\Omega$ . This internal resistor is across the OUT+ and OUT- terminals.

To improve the match at the high end of the frequency range (65MHz), a reactive match may be employed as part of the ensuing diplex filter. A series inductor (typ 180nH) followed by a shunt capacitor (typ 33µF) can be placed directly after the output transformer. This match will also improve the gain flatness substantially.

As mentioned above, the matching components may be incorporated into the diplex filter design. Optimize the input impedance of the diplex filter to be 35 + j35 (typ) at 65MHz when using the specified output transformer.

#### MAX3517

The MAX3517 does not have an internal matching resistor. This allows the device performance to be optimized for various load impedances.

When  $300\Omega$  resistors are placed across the output terminals of the device, performance identical to the MAX3514 will result. If an impedance higher than  $300\Omega$  is used, additional gain will result.

Note also that a 2:1 voltage-ratio output transformer is not needed.

When operating the device with arbitrary output resistance and XFMR turns ratio, take care not to exceed the allowable power dissipation (see *Absolute Maximum Ratings*).

#### **Transformer**

To match the output of the MAX3514/MAX3516 to a  $75\Omega$  load, a 2:1 voltage-ratio transformer is required. This transformer must have adequate bandwidth to cover the intended application. Note that most RF transformers specify bandwidth with a  $50\Omega$  source on the primary and a matching resistance on the secondary winding. Operating in a  $75\Omega$  system will tend to shift the low-frequency edge of the transformer bandwidth specification up by a factor of 1.5 due to primary inductance. Keep this in mind when specifying a transformer.

Bias to the output stage is provided through the center tap on the transformer primary. This greatly diminishes the on/off transients present at the output when switching between transmit and transmit-disable modes. Commercially available transformers typically have adequate balance between half-windings to achieve substantial transient cancellation.

Finally, keep in mind that transformer core inductance varies proportionally with temperature. If the application requires low temperature extremes (less than 0°C), adequate primary inductance must be present to sustain low-frequency output capability as temperatures drop. In general, this will not be a problem as modern RF transformers have adequate bandwidth.

#### **Input Circuit**

To achieve rated performance, the inputs of the MAX3514/MAX3516/MAX3517 must be driven differentially with an appropriate input level. The differential input impedance is approximately 1.5k $\Omega$ . Most applications will require a differential low-pass filter preceding the device. The filter design will dictate a terminating impedance of a specified value. Place this load impedance across the AC-coupled input pins (see *Typical Operating Circuit*).

The MAX3514/MAX3517 have sufficient gain to produce an output level of +61dBmV (QPSK through a 2:1 transformer) when driven with a +34dBmV input signal. The MAX3516 provides an additional 3dB of gain and output level. When a lower input level is present, the maximum output level will be reduced proportionally and output linearity will increase. If an input level greater than +34dBmV is used, the 3rd-order distortion performance will degrade slightly.

If single-ended sources drive the MAX3514/MAX3516/ MAX3517, one of the input terminals must be capacitively coupled to ground (IN+ or IN-). The value of this capacitor must be large enough to look like a short circuit at the lowest frequency of interest. For operation at 5MHz with a  $75\Omega$  source impedance, a value of  $0.001\mu F$  will suffice.

### \_Layout Issues

A well-designed PC board is an essential part of an RF circuit. For best performance, pay attention to power-supply layout issues, as well the output circuit layout.

#### **Output Circuit Layout**

The differential implementation of the MAX3514/MAX3516/MAX3517s' output has the benefit of significantly reducing even-order distortion, the most significant of which is 2nd-harmonic distortion. The degree of distortion cancellation depends on the amplitude and phase balance of the overall circuit. It is important to keep the trace lengths from the output pins equal.

#### **Power-Supply Layout**

For minimal coupling between different sections of the IC, the ideal power-supply layout is a star configuration. This configuration has a large-value decoupling capacitor at the central power-supply node. The power-supply traces branch out from this node, each going to a separate power-supply node in the circuit. At the end of each of these traces is a decoupling capacitor that provides a very low impedance at the frequency of interest. This arrangement provides local power-supply decoupling at each power-supply pin.

The power-supply traces must be made as thick as practical.

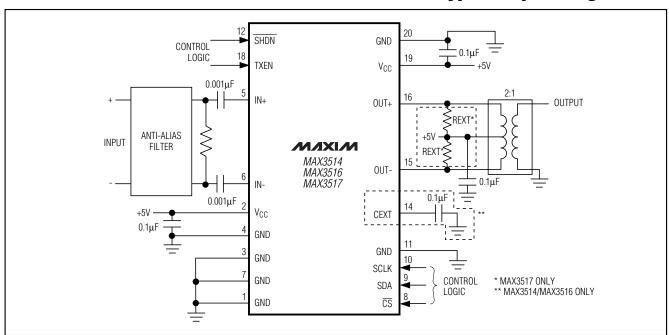
Ground inductance degrades distortion performance. Therefore, ground plane connections to pin 4 and pin 20 should be made with multiple vias if necessary.

#### **Exposed Paddle Thermal Considerations**

The exposed paddle (EP) of the MAX3516's 20-pin TSSOP-EP package provides a low thermal-resistance path to the die. It is important that the PC board on which the MAX3516 is mounted, be designed to conduct heat from this contact. In addition, the EP should be provided with a low inductance path to electrical ground.

It is recommended that the EP be soldered to a ground plane on the PC board, either directly or through an array of plated via holes.

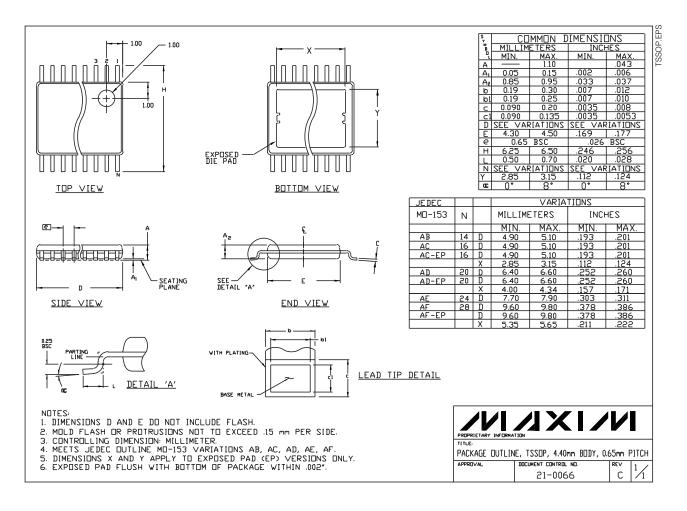
### **Typical Operating Circuit**



**Chip Information** 

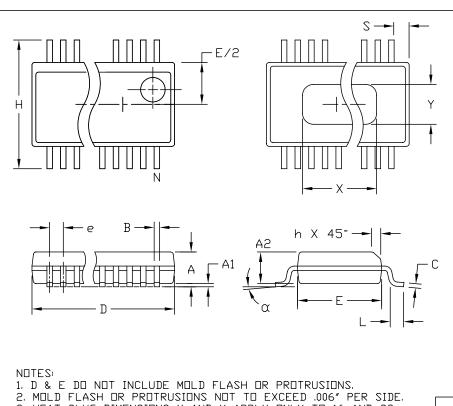
TRANSISTOR COUNT: 1006

### Package Information



### Package Information (continued)

TNICHES



	INCH	F2	MILLIME LEKS					
DIM	MIN	MAX	MIN	MAX				
Α	.061	.068	1.55	1.73				
A1	.004	.0098	0.102	0.249				
A2	.055	.061	1.40	1.55				
В	.008	.012	0.20	0.31				
С	.0075	.0075 .0098 0.191		0.249				
D	SEE VARIATIONS							
E	.150	.157	3.81	3.99				
е	.025	BSC	0.635 BSC					
Н	.230	.244	5.84	6.20				
h	.010	.016	0.25	0.41				
L	.016	.035	0.41	0.89				
N	SEE VARIATIONS							
X	SEE VARIATIONS							
Υ	.071	.087	1.803	2.209				
α	0, 8,		0*	8*				

MILL IMPTEDS

#### VARIATIONS:

	INCHE	S	MILLIM			
	MIN.	MAX.	MIN.	MAX.	N	
D	.189	.196	4.80	4.98	16	АΑ
S	.0020	.0070	0.05	0.18		
Х	.107	.123	2.72	3.12		
D	.337	.344	8.56	8.74	20	ΑВ
S	.0500	.0550	1.270	1.397		
D	.337	.344	8.56	8.74	24	AC
S	.0250	.0300	0.635	0.762		
D	.386	.393	9.80	9.98	28	ΑD
S	.0250	.0300	0.635	0.762		
Х	.271	.287	6.88	7.29		

# PROPRIETARY INFORMATION

C 1 21-0055

- 3. HEAT SLUG DIMENSIONS X AND Y APPLY ONLY TO 16 AND 28 LEAD POWER-QSOP PACKAGES.
- 4. CONTROLLING DIMENSIONS: INCHES. 5. MEETS JEDEC MO137.

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