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DS3021-2.2

## **ZN426E8**8-BIT D-A CONVERTER

The ZN426 is a monolithic 8-bit D-A converter containing an R-2R ladder network of diffused resistors with precision bipolar switches, and a 2.5V precision voltage reference.

#### **FEATURES**

- $\pm \frac{1}{2}$  LSB Linearity Error
- Guaranteed Monolithic over the Full Operating Temperature Range
- 0°C to +70°C
- TTL and 5V CMOS Compatible
- Single 5V Supply
- Settling Time 1 microsecond Typical
- Only Reference Capacitor and Resistor Required

#### **ABSOLUTE MAXIMUM RATINGS**

Supply voltage +7.0V Max.voltage, logic and V<sub>REF</sub> inputs+5.5V Operating temperature range0°C to +70°C Storage temperature range-55°C to +125°C

#### ORDERING INFORMATION

Ambient operating temperature 0°C to +70°C Package, ZN426E8DP14

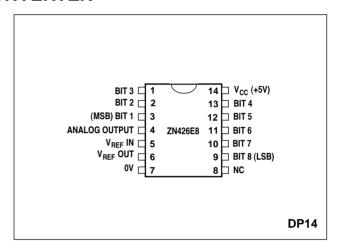


Fig.1 Pin connections (not to scale) - top view

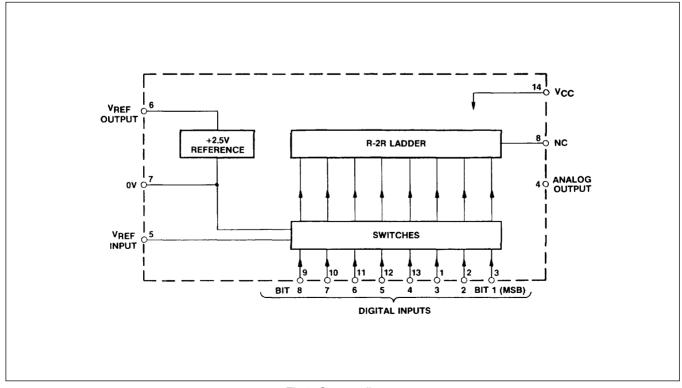


Fig.2 System diagram

#### **ZN426**

#### **ELECTRICAL CHARACTERISTICS**

(at  $T_{amb} = 25$ °C and  $V_{CC} = +5V$  unless otherwise specified)

Parameter	Symbol	Min.	Тур.	Max.	Units	Conditions
Converter Resolution		8	-	-	bits	
Non-linearity		-	-	±0.5	LSB	
Differential non-linearity		-	±0.5	-	LSB	Note 1
Settling time to 0.5LSB		-	1.0	-	μs	1LSB step
Settling time to 0.5LSB		-	2.0	-	μs	All bits ON to OFF or OFF to ON
Offset voltage ZN426E8 and D	Vos	-	3.0	5.0	mV	All bits OFF
V <sub>OS</sub> temperature coefficient		-	5	-	μV/°C	
Full-scale output		2.545	2.550	2.555	V	All bits ON Ext. V <sub>REF</sub> = 2.560V
Full-scale temp. coefficient		-	3	-	ppm/°C	Ext. V <sub>REF</sub> = 2.560V
Non-linearity temp. coefficient		-	7.5	-	ppm/°C	Relative to F.S.R.
Analog output resistance	R <sub>O</sub>	-	10	-	kΩ	
External reference voltage		0	-	3.0	V	
Supply voltage	V <sub>CC</sub>	4.5	-	5.5	V	
Supply current	Is	-	5	9	mA	
High level input voltage	V <sub>IH</sub>	2.0	-	-	V	
Low level input voltage	V <sub>IL</sub>	-	-	0.7	V	
High level input current	I <sub>IH</sub>	-	-	10	μΑ	V <sub>CC</sub> = max. V <sub>I</sub> = 2.4V
		-	-	100	μΑ	$V_{CC} = max.$ $V_{I} = 5.5V$
Low level input current	I <sub>IL</sub>	-	-	-0.18	mA	$V_{CC} = max.$ $V_{I} = 0.3V$
Internal voltage reference Output voltage	V <sub>REF</sub>	2.475	2.55	2.626	V	Note 2 R <sub>REF</sub> = 390Ω
Slope resistance	R <sub>S</sub>	-	1	2	Ω	R <sub>REF</sub> = 390Ω
V <sub>REF</sub> temperature coefficient		-	40	-	ppm/°C	$R_{REF} = 390\Omega$

NOTE 1: Monotonic over full temperature range. NOTE 2: The internal reference requires a  $1\mu F$  stabilising capacitor between  $V_{REF}$  out and  $V_{REF}$  out  $V_{CC}$  and  $V_{REF}$  out.

#### **INTRODUCTION**

The ZN426 is an 8-bit D-A converter. It contains an advanced design of R-2R ladder network and an array of precision bipolar switches plus a 2.5V precision voltage reference on a single monolithic chip.

The special design of the ladder network results in full 8-bit accuracy using normal defused resistors.

The use of on-chip voltage reference is pin optional to retain flexibility. An external fixed or varying reference may therefore be substituted. In this case there is no need to supply power to the internal reference so  $R_{\text{REF}}$  and  $C_{\text{REF}}$ can be omitted.

The converter is of the voltage switching type and uses an R-2R resistor ladder network as shown in Fig.3.

Each 2R element is connected either to 0V or  $V_{REF}$  by transistor switches specially designed for low offset voltage (typically 1mV).

Binary weighted voltages are produced at the output of the R-2R ladder, the value depending on the digital number applied to the bit inputs.

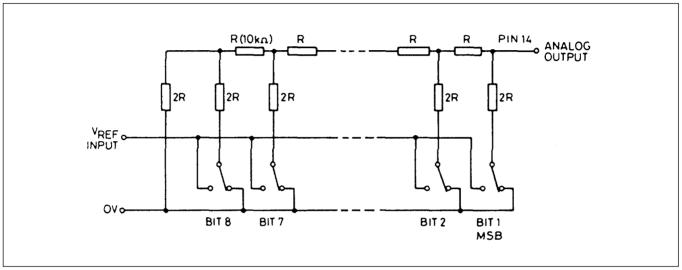


Fig.3 The R-2R ladder network

#### **APPLICATIONS**

#### 8-bit D-A Converter

The ZN426 gives an analog voltage output directly from pin 4 therefore the usual current to voltage converting amplifier is not required. The output voltage drift, due to the temperature coefficient of the analog output resistance  $R_O$ , will be less than 0.004% per  $^{\circ}C$  (or 1LSB/100°C) if  $R_L$  is chosen to be  $\geq 650 k\Omega$ .

In order to remove the offset voltage and to calibrate the converter a buffer amplifier is necessary. Fig.4 shows a typical scheme using the internal reference voltage. To minimise temperature drift in this and similar applications the source resistance to the inverting input of the operational amplifier should be approximately  $6k\Omega$ . The calibration procedure is as follows:

- i. Set all bits to OFF (low) and adjust  $R_2$  until  $V_{OUT} = 0.000V$ .
- ii. Set all bits to ON (high) and adjust  $R_1$  until  $V_{OUT}$  = Normal full-scale reading 1LSB
- iii. Repeat i. and ii.

e.g. Set F.S.R. to + 3.840V - 1LSB  
=3.825V  
(1LSB = 
$$\frac{3.84}{256}$$
 =15.0mV)

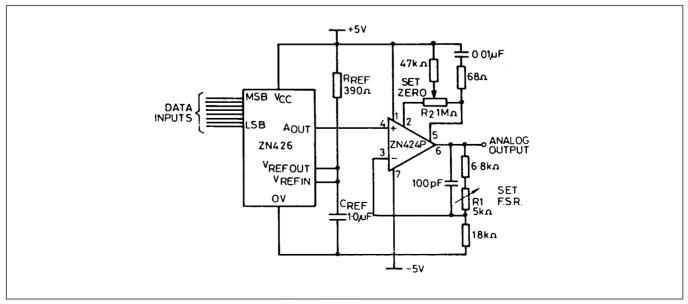


Fig.4 8-bit D-A converter

#### **ZN426**

#### **Alternative Output Buffer Using the 741**

The circuit of Fig.5, employing the 741 operational amplifier, may be used as the output buffer.

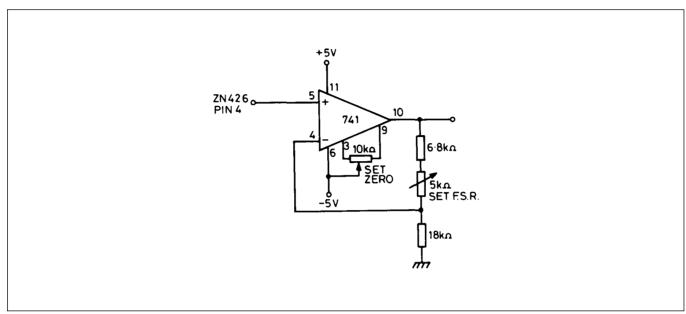


Fig.5 The ZLD741 as output buffer



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