



SGLS305D – JULY 2005 – REVISED SEPTEMBER 2010

HIGH-SPEED PWM CONTROLLER

Check for Samples: UC2825A-EP

FEATURES

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- Improved Version of the UC2825 PWM
- Compatible With Voltage-Mode or Current-Mode Control Methods
- Practical Operation at Switching Frequencies to 1 MHz
- 50-ns Propagation Delay to Output
- High-Current Dual Totem-Pole Outputs
 (2-A Peak)
- Trimmed Oscillator Discharge Current
- Low 100-µA Startup Current
- Pulse-by-Pulse Current-Limiting Comparator
- Latched Overcurrent Comparator With Full
 Cycle Restart

SUPPORTS DEFENSE, AEROSPACE, AND MEDICAL APPLICATIONS • Controlled Baseline

- One Assembly/Test Site
- One Fabrication Site
- Available in Military (–55°C/125°C) Temperature Range⁽¹⁾
- Extended Product Life Cycle
- Extended Product-Change Notification
- Product Traceability
- (1) Additional temperature ranges are available contact factory

DESCRIPTION/ORDERING INFORMATION

The UC2825A-EP pulse width modulation (PWM) controller is an improved version of the standard UC2825. Performance enhancements have been made to several of the circuit blocks. Error amplifier gain bandwidth product is 12 MHz, while input offset voltage is 2 mV. Current-limit threshold is specified to a tolerance of 5%. Oscillator discharge current is specified at 10 mA for accurate dead-time control. Frequency accuracy is improved to 6%. Startup supply current, typically 100 µA, is ideal for off-line applications. The output drivers are redesigned to actively sink current during undervoltage lockout (UVLO) at no expense to the startup current specification. In addition, each output is capable of 2-A peak currents during transitions.

Functional improvements also have been implemented in this family. The UC2825A-EP shutdown comparator is now a high-speed overcurrent comparator with a threshold of 1.2 V. The overcurrent comparator sets a latch that ensures full discharge of the soft-start capacitor before allowing a restart. While the fault latch is set, the outputs are in the low state. In the event of continuous faults, the soft-start capacitor is fully charged before discharge to ensure that the fault frequency does not exceed the designed soft-start period. The UC2825 CLOCK pin is CLK/LEB in the UC2825A-EP. This pin combines the functions of clock output and leading-edge blanking adjustment and has been buffered for easier interfacing.

The UC2825A-EP has dual alternating outputs and the same pin configuration as UC2825. UVLO thresholds are identical to the original UC2825.

Consult the application report, *The UC3823A,B and UC2825A,B Enhanced Generation of PWM Controllers*, literature number SLUA125, for detailed technical and applications information.



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Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



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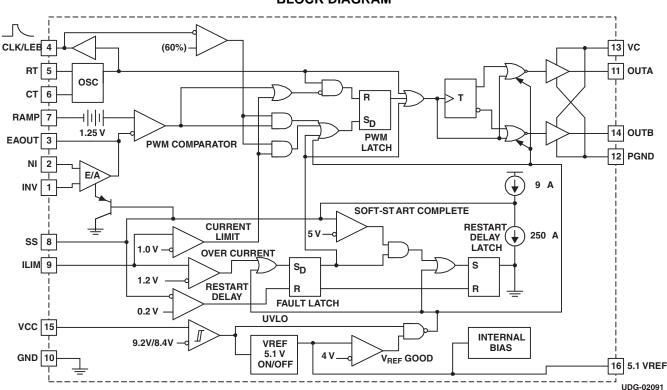
This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

T _A	PACKAGE ⁽¹⁾	ORDERABLE PART NUMBER	TOP-SIDE MARKING						
-40°C to 125°C	SOIC – DW	UC2825AQDWREP	UC2825AQEP						
–55°C to 125°C	SOIC – DW	UC2825AMDWREP	UC2825AMEP						

ORDERING INFORMATION

 Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.



BLOCK DIAGRAM



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PIN ASSIGNMENTS

DW PACKAGE (TOP VIEW)								
INV	[1	16] VREF					
NI	[2	15] VCC					
EAOUT	[3	14] OUTB					
CLK/LEB	[4	13] VC					
RT	[5	12] PGND					
CT	[6	11] OUTA					
RAMP	[7	10] GND					
SS	[8	9] ILIM					

TERMINAL FUNCTIONS

TERMI	NAL	1/0	DESCRIPTION
NAME	NO.	1/0	DESCRIPTION
CLK/LEB	4	0	Clock/leading-edge blanking. Output of the internal oscillator.
СТ	6	I	Capacitor timing. Timing capacitor connection for oscillator frequency programming. The timing capacitor should be connected to the device ground using minimal trace length.
EAOUT	3	0	Output of the error amplifier for compensation
GND	10		Analog ground return
ILIM	9	I	Input to the current-limit comparator
INV	1	I	Inverting input to the error amplifier
NI	2	I	Noninverting input to the error amplifier
OUTA	11	0	High-current totem-pole output A of the on-chip drive stage
OUTB	14	0	High-current totem-pole output B of the on-chip drive stage
PGND	12		Ground return for the output driver stage
RAMP	7	I	Noninverting input to the PWM comparator, with 1.25-V internal input offset. In voltage-mode operation, this serves as the input voltage feed-forward function by using the CT ramp. In peak current mode operation, this serves as the slope compensation input.
RT	5	I	Resistor timing. Timing resistor connection for oscillator frequency programming.
SS	8	I	Soft-start. SS also doubles as the maximum duty cycle clamp.
VC	13		Power-supply for the output stage. This pin should be bypassed with a 0.1 - μ F monolithic ceramic low-ESL capacitor with minimal trace lengths.
VCC	15	0	Power supply for the device. This pin should be bypassed with a 0.1 - μ F monolithic ceramic low-ESL capacitor with minimal trace lengths.
VREF	16		5.1-V reference. For stability, the reference should be bypassed with a 0.1- μ F monolithic ceramic low-ESL capacitor and minimal trace length to the ground plane.

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Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			UNIT
V _{IN}	Supply voltage	VC, VCC	22 V
I _O	Source or sink current, DC	OUTA, OUTB	0.5 A
I _O	Source or sink current, pulse (0.5 µs)	OUTA, OUTB	2.2 A
	Anolog inpute	INV, NI, RAMP	–0.3 V to 7 V
	Analog inputs	ILIM, SS	–0.3 V to 6 V
	Power ground	PGND	±0.2 V
	Outputs	OUTA, OUTB limits	PGND -0.3 V to VC +0.3 V
I _{CLK}	Clock output current	CLK/LEB	–5 mA
I _{O(EA)}	Error amplifier output current	EAOUT	5 mA
I _{SS}	Soft-start sink current	SS	20 mA
I _{OSC}	Oscillator charging current	RT	–5 mA
TJ	Operating virtual junction temperature range		–55°C to 150°C
T _{stg}	Storage temperature ⁽²⁾		-65°C to 150°C
	Lead temperature 1,6 mm (1/16 in) from case for 10 s		–55°C to 150°C
t _{stg}	Storage temperature ⁽²⁾		-65°C to 150°C
	Lead temperature 1,6 mm (1/16 in) from case for 10 s		300°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Long-term high-temperature storage and/or extended use at maximum recommended operating conditions may result in a reduction of overall device life. See http://www.ti.com/ep_quality for additional information on enhanced plastic packaging.

Table 1. DISSIPATION RATING TABLE – FREE-AIR TEMPERATURE

PACKAGE	AIR FLOW	T _A ≤ 25°C	DERATING FACTOR	T _A = 70°C	T _A = 85°C	T _A = 125°C
	(CFM)	POWER RATING	ABOVE T _A = 25°C	POWER RATING	POWER RATING	POWER RATING
DW	0	1.105 W	9.62 mW/°C	673 mW	528 mW	144 mW

4



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Electrical Characteristics

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Referen	ce, V _{REF}		l			
Vo	Output voltage range	$T_{J} = 25^{\circ}C, I_{O} = 1 \text{ mA}$	5.05	5.1	5.15	V
-	Line regulation	12 V ≤ VCC ≤ 20 V		2	15	mV
	Load regulation	$1 \text{ mA} \le I_{O} \le 10 \text{ mA}$		5	20	mV
	Total output variation	Line, load, temperature	5.03		5.17	V
	Temperature stability ⁽¹⁾	$T_{(min)} < T_A < T_{(max)}$		0.2	0.4	mV/°C
	Output noise voltage ⁽¹⁾	10 Hz < f < 10 kHz		50		μV_{RMS}
	Long-term stability ⁽¹⁾	T _J = 125°C, 1000 h		5	25	mV
	Short-circuit current	V _{REF} = 0 V	30	60	90	mA
Oscillat	or		I			
,	(1)	$T_J = 25^{\circ}C$	375	400	425	kHz
fosc	Initial accuracy ⁽¹⁾	R _T = 6.6 kΩ, C _T = 220 pF, T _A = 25°C	0.9	1	1.1	MHz
	T - (-),,,	Line, temperature	350		450	kHz
	Total variation ⁽¹⁾	$R_{T} = 6.6 \text{ k}\Omega, C_{T} = 220 \text{ pF}$	0.85		1.15	MHz
	Voltage stability	12 V ≤ VCC ≤ 20 V			1%	
	Temperature stability ⁽¹⁾	$T_{(min)} < T_A < T_{(max)}$		5%		
	High-level output voltage, clock		3.7	4		V
	Low-level output voltage, clock			0	0.2	V
	Ramp peak		2.6	2.8	3	V
	Ramp valley		0.7	1	1.25	V
	Ramp valley to peak		1.6	1.8	2	V
losc	Oscillator discharge current	$R_T = OPEN, V_{CT} = 2 V$	9	10	11	mA
Error Ar	nplifier					
	Input offset voltage			2	10	mV
	Input bias current			0.6	3	μA
	Input offset current			0.1	1	μA
	Open-loop gain	1 V < V _O < 4 V	60	95		dB
CMRR	Common-mode rejection ratio	1.5 V < V _{CM} < 5.5 V	75	95		dB
PSRR	Power-supply rejection ratio	12 V < VCC < 20 V	85	110		dB
I _{O(sink)}	Output sink current	V _{EAOUT} = 1 V	1	2.5		mA
I _{O(src)}	Output source current	V _{EAOUT} = 4 V		-1.3	-0.5	mA
	High-level output voltage	$I_{EAOUT} = -0.5 \text{ mA}$	4.5	4.7	5	V
	Low-level output voltage	$I_{EAOUT} = -1 \text{ mA}$	0	0.5	1	V
	Gain bandwidth product	f = 200 kHz	6	12		MHz
	Slew rate ⁽¹⁾		6	9		V/µs
PWM Co	omparator					
I _{BIAS}	Bias current, RAMP	V _{RAMP} = 0 V		-1	-8	μA
	Minimum duty cycle				0%	
	Maximum duty cycle		85%			
t _{LEB}	Leading-edge blanking time	$R_{LEB} = 2 \text{ k}\Omega, C_{LEB} = 470 \text{ pF}$	300	375	450	ns
R _{LEB}	Leading-edge blanking resistance	$V_{CLK/LEB} = 3 V$	8.5	10	11.5	kΩ
V _{ZDC}	Zero dc threshold voltage, EAOUT	V _{RAMP} = 0 V	1.1	1.25	1.4	V
t _{DELAY}	Delay-to-output time ⁽¹⁾	$V_{EAOUT} = 2.1 \text{ V}, V_{ILIM} = 0 \text{-V to } 2 \text{-V step}$		50	80	ns

(1) Specified by design. Not production tested.

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Electrical Characteristics (continued)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Curren	t Limit/Start Sequence/Fault				1	
I _{SS}	Soft-start charge current	V _{SS} = 2.5 V	8	14	20	μA
V _{SS}	Full soft-start threshold voltage		4.3	5		V
I _{DSCH}	Restart discharge current	V _{SS} = 2.5 V	100	250	350	μA
I _{SS}	Restart threshold voltage			0.3	0.5	V
I _{BIAS}	ILIM bias current	V _{ILIM} = 0-V to 2-V step			15	μA
I _{CL}	Current-limit threshold voltage		0.95	1	1.05	V
	Overcurrent threshold voltage		1.14	1.2	1.26	V
t _d	Delay-to-output time, ILIM ⁽²⁾	V _{ILIM} = 0-V to 2-V step		50	80	ns
Output						
		I _{OUT} = 20 mA		0.25	0.4	V
	Low-level output saturation voltage	I _{OUT} = 200 mA		1.2	2.2	V
		I _{OUT} = 20 mA		1.9	2.9	V
	High-level output saturation voltage	I _{OUT} = 200 mA		2	3	V
t _r , t _f	Rise/fall time ⁽²⁾	C _L = 1 nF		20	45	ns
Underv	oltage Lockout (UVLO)					
	Start threshold voltage		8.4	9.2	9.6	V
	UVLO hysteresis		0.4	0.8	1.2	V
Supply	Current					
l _{su}	Startup current	$V_{C} = VCC = V_{TH}(start) - 0.5 V$		100	300	μA
I _{CC}	Input current			28	36	mA

(2) Specified by design. Not production tested.

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UC2825A-EP

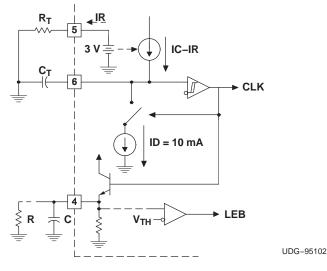
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APPLICATION INFORMATION

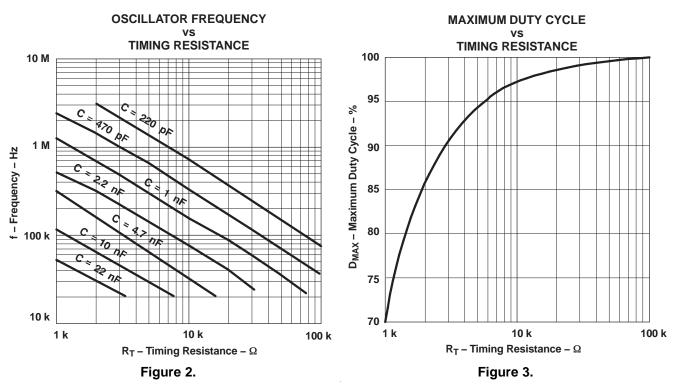
The oscillator of the UC2825A-EP is a sawtooth. The rising edge is governed by a current controlled by the RT pin and value of capacitance at the CT pin (C_{CT}). The falling edge of the sawtooth sets dead time for the outputs. Selection of RT should be done first, based on desired maximum duty cycle. CT then can be chosen, based on the desired frequency (RT) and D_{MAX} . The design equations are:

$$R_{T} = \frac{3 V}{(10 \text{ mA}) \times (1 - D_{\text{MAX}})} \qquad C_{T} = \frac{(1.6 \times D_{\text{MAX}})}{(R_{T} \times f)}$$
(1)

Recommended values for R_T range from 1 k Ω to 100 k Ω . Control of D_{MAX} less than 70% is not recommended.







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Leading-Edge Blanking (LEB)

The UC2825A-EP performs fixed-frequency PWM control. The UC2825A-EP outputs are alternately controlled. During every other cycle, one output is off. Each output then switches at one-half the oscillator frequency, varying in duty cycle from 0 to less than 50%.

To limit maximum duty cycle, the internal clock pulse blanks both outputs low during the discharge time of the oscillator. On the falling edge of the clock, the appropriate output(s) is driven high. The end of the pulse is controlled by the PWM comparator, current-limit comparator, or the overcurrent comparator.

Normally the PWM comparator senses a ramp crossing a control voltage (error-amplifier output) and terminates the pulse. LEB causes the PWM comparator to be ignored for a fixed amount of time after the start of the pulse. This allows noise inherent with switched-mode power conversion to be rejected. The PWM ramp input may not require any filtering as a result of LEB.

To program an LEB period, connect a capacitor, C, to CLK/LEB. The discharge time set by C and the internal 10-k Ω resistor determines the blanked interval. The 10-k Ω resistor has a 10% tolerance. For more accuracy, an external 2-k Ω 1% resistor (R) can be added, resulting in an equivalent resistance of 1.66 k Ω with a tolerance of 2.4%. The design equation is:

 $\mathbf{t}_{\mathsf{LEB}} = \, \mathbf{0.5} \times \big(\mathsf{R} \parallel \mathbf{10} \ \mathsf{k}\Omega\big) \times \,\mathsf{C}$

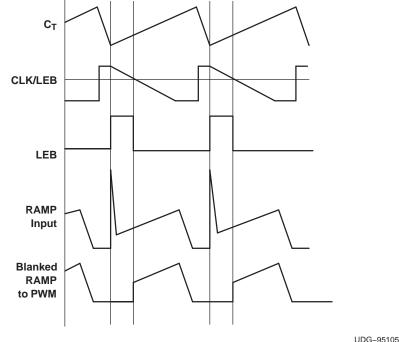
(2)

Values of R less than 2 k Ω should not be used.

LEB also is applied to the current-limit comparator. After LEB, if the ILIM pin exceeds the 1-V threshold, the pulse is terminated. The overcurrent comparator, however, is not blanked. It catches catastrophic overcurrent faults without a blanking delay. Any time the ILIM pin exceeds 1.2 V, the fault latch is set, and the outputs are driven low. For this reason, some noise filtering may be required on the ILIM pin.



8







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Undervoltage Lockout (UVLO), Soft-Start, and Fault Management

Soft-start is programmed by a capacitor on the soft-start (SS) pin. At power up, SS is discharged. When SS is low, the error-amplifier output also is forced low. While the internal $9-\mu A$ source charges SS, the error-amplifier output follows until closed-loop regulation takes over.

Anytime ILIM exceeds 1.2 V, the fault latch is set and the output pins are driven low. The soft-start cap then is discharged by a $250-\mu$ A current sink. No more output pulses are allowed until soft-start is fully discharged and ILIM is below 1.2 V. At this point, the fault latch resets and the chip executes a soft-start.

Should the fault latch get set during soft-start, the outputs are terminated immediately, but the soft-start capacitor does not discharge until it has been fully charged first. This results in a controlled hiccup interval for continuous fault conditions.

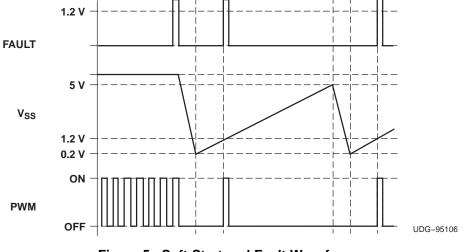


Figure 5. Soft-Start and Fault Waveforms

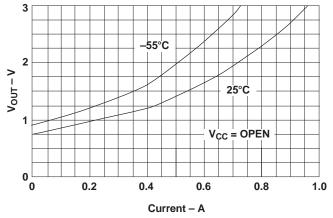
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Active-Low Outputs During UVLO

The UVLO function forces the outputs to be low and considers both VCC and VREF before allowing the chip to operate.



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Figure 6. Output Voltage vs Output Current

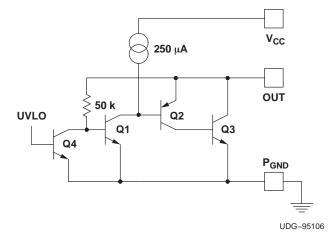
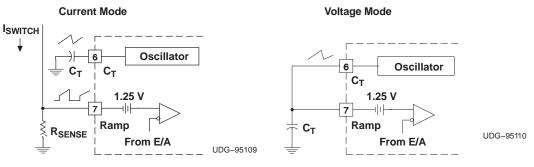


Figure 7. Output V and I During UVLO

Control Methods





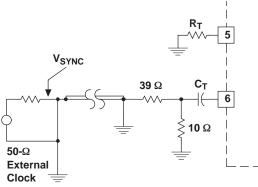


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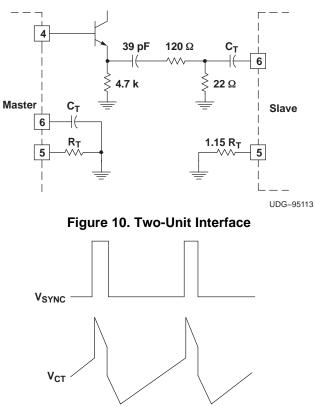
Synchronization

The oscillator can be synchronized by an external pulse inserted in series with the timing capacitor. Program the free-running frequency of the oscillator to be 10% to 15% slower than the desired synchronous frequency. The pulse width should be greater than 10 ns and less than half the discharge time of the oscillator. The rising edge of the CLK/LEB pin can be used to generate a synchronizing pulse for other chips. Note that CLK/LEB no longer accepts an incoming synchronizing signal.



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Figure 11. Operational Waveforms

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High-Current Outputs

Each totem-pole output of the UC2825A-EP can deliver a 2-A peak current into a capacitive load. The output can slew a 1000-pF capacitor by 15 V in approximately 20 ns. Separate collector supply (VC) and power ground (PGND) pins help decouple the device analog circuitry from the high-power gate drive noise. The use of 3-A Schottky diodes (1N5120, USD245, or equivalent) (see Figure 13) from each output to both VC and PGND are recommended. The diodes clamp the output swing to the supply rails, necessary with any type of inductive/capacitive load, typical of a MOSFET gate. Schottky diodes must be used because a low forward voltage drop is required. Do not use standard silicon diodes.

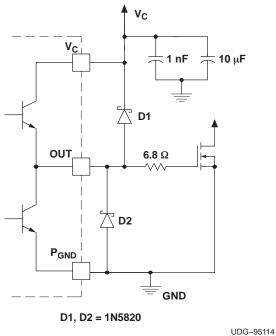


Figure 12. Power MOSFET Drive Circuit

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Ground Planes

Each output driver of these devices is capable of 2-A peak currents. Careful layout is essential for correct operation of the chip. A ground plane must be employed. A unique section of the ground plane must be designated for high di/dt currents associated with the output stages. This point is the power ground to which the PGND pin is connected. Power ground can be separated from the rest of the ground plane and connected at a single point, although this is not necessary if the high di/dt paths are well understood and accounted for. VCC should be bypassed directly to power ground with a good high-frequency capacitor. The sources of the power MOSFET should connect to power ground as should the return connection for input power to the system and the bulk input capacitor. The output should be clamped with a high-current Schottky diode to both VCC and PGND. Nothing else should be connected to power ground.

VREF should be bypassed directly to the signal portion of the ground plane with a good high-frequency capacitor. Low-ESR/ESL ceramic 1-mF capacitors are recommended for both VCC and VREF. All analog circuitry likewise, should be bypassed to the signal ground plane.

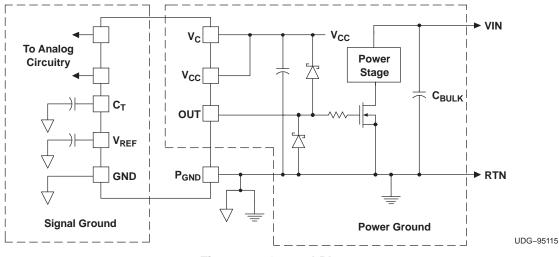


Figure 13. Ground Planes

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Open-Loop Test Circuit

This test fixture is useful for exercising many functions of this device family and measuring their specifications. As with any wideband circuit, careful grounding and bypass procedures should be followed. The use of a ground plane is highly recommended.

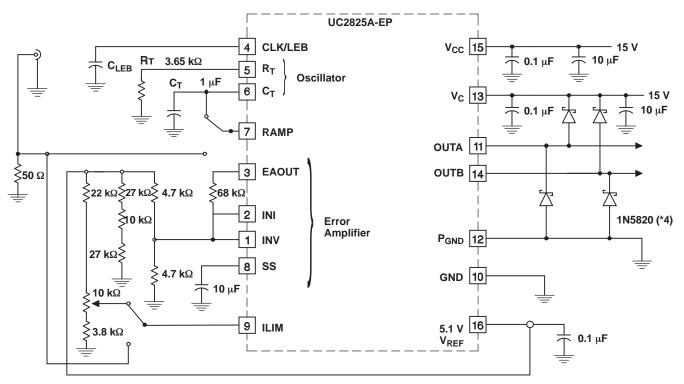


Figure 14. Open-Loop Test Circuit



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PACKAGING INFORMATION

(1) Package T	Drawing		Package Qty		Lead/ Ball Finish	MSL Pe
VE SOIC	DW	16	2000	0		
			2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
VE SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
VE SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
VE SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
VE SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
VE SOIC	DW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260
	VE SOIC VE SOIC VE SOIC	VE SOIC DW VE SOIC DW VE SOIC DW	VE SOIC DW 16 VE SOIC DW 16 VE SOIC DW 16	VE SOIC DW 16 2000 VE SOIC DW 16 2000 VE SOIC DW 16 2000	WE SOIC DW 16 2000 Green (RoHS & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & no Sb/Br)	WE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br) VE SOIC DW 16 2000 Green (RoHS & CU NIPDAU & no Sb/Br)

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www. information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retard in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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OTHER QUALIFIED VERSIONS OF UC2825A-EP :

- Catalog: UC2825A
- Automotive: UC2825A-Q1

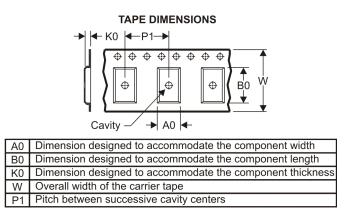
NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Automotive Q100 devices qualified for high-reliability automotive applications targeting zero defects

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



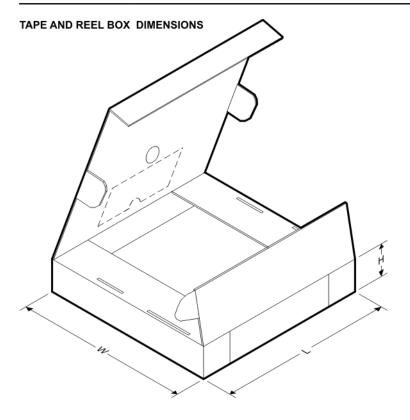
*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UC2825AMDWREP	SOIC	DW	16	2000	330.0	16.4	10.85	10.8	2.7	12.0	16.0	Q1



PACKAGE MATERIALS INFORMATION

30-Aug-2010

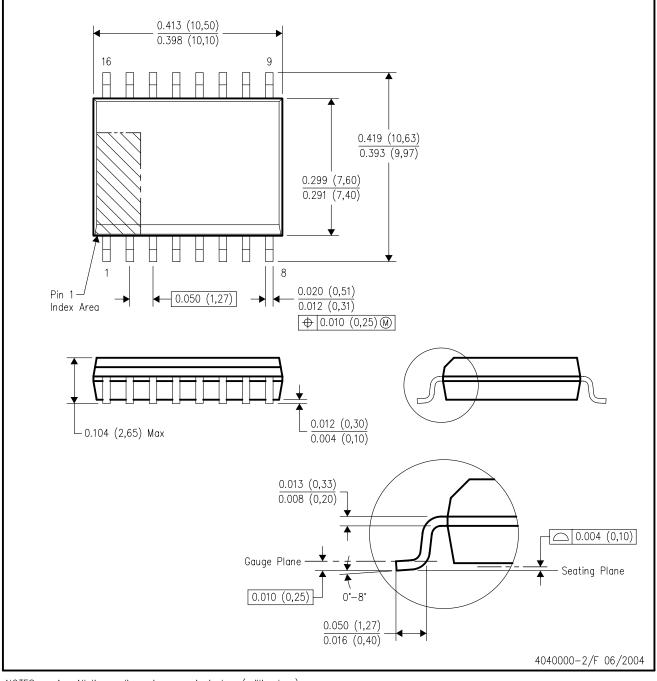


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UC2825AMDWREP	SOIC	DW	16	2000	346.0	346.0	33.0

DW (R-PDSO-G16)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).

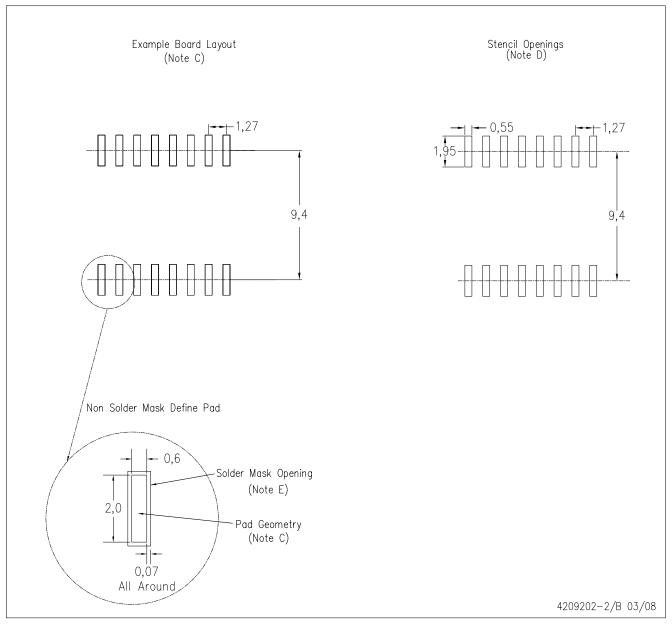
D. Falls within JEDEC MS-013 variation AA.



LAND PATTERN

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NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Refer to IPC7351 for alternate board design.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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