

# **Phase-Frequency Detector**

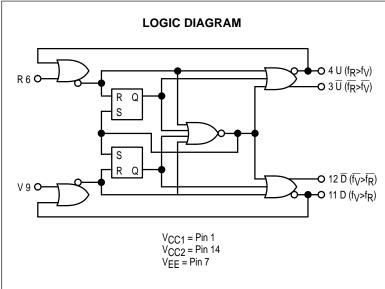
The MC12040 is a phase–frequency detector intended for use in systems requiring zero phase and frequency difference at lock. In combination with a voltage controlled oscillator (such as the MC1648, MC12147, MC12148 or MC12149), it is useful in a broad range of phase–locked loop applications.

• Operating Frequency = 80 MHz Typical

#### **Pin Conversion Table**

14 PIN DIP	1	2	3	4	5	6	7	8	9	10	11	12	13	14
20 PIN PLCC	2	3	4	6	8	9	10	12	13	14	16	18	19	20

Inp	uts		Outputs								
R	v	U	D	Ū	D						
0 0 1 0	0 1 1 1	X X X X	X X X X	X X X X	X X X X						
1 0 1 1	1 1 1 0	1 1 1	0 0 0 0	0 0 0	1 1 1						
1 1 1 1	1 0 1 0	0 0 0 0	0 0 1 1	1 1 1	1 1 0 0						
1 0 1	1 1 1	0 0 0	1 1 0	1 1 1	0 0 1						



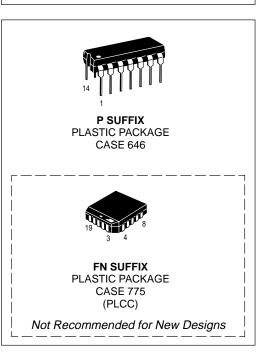
#### TRUTH TABLE

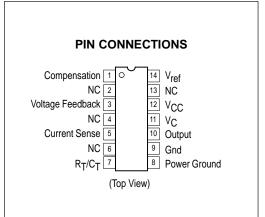
This is not strictly a functional truth table; i.e., it does not cover all possible modes of operation. However, it gives a sufficient number of tests to ensure that the device will function properly in all modes of operation.



## PHASE-FREQUENCY DETECTOR

SEMICONDUCTOR TECHNICAL DATA



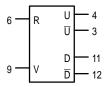


## ORDERING INFORMATION

Device	Operating Temperature Range	Package
MC12040P	$T_A = 0^\circ$ to +75°C	Plastic

### **ELECTRICAL CHARACTERISTICS**

The MC12040 has been designed to meet the dc equilibrium has been established. Outputs are terminated through a 50 ohm resistor to +3.0 V for +5.0 V tests and through a 50 ohm resistor to -2.0 V for -5.2 V tests.



TEST VOLTAGE VALUES

(Volts)

NOTE: For more information on using an ECL device in a +5V system, refer to Motorola Application Note AN1406/D, "Designing with PECL (ECL at +5.0V)"

	7 NINI +00/D, DC	Jigimig	WILLI I L	-0- (		0.0 v )						· ·			1
							0	Test Tem	perature	V <sub>IHmax</sub>	V <sub>ILmin</sub>	V <sub>IHAmin</sub>	V <sub>ILAmax</sub>	V <sub>EE</sub>	
									0°C	-0.840	-1.870	-1.145	-1.490	-5.2	
									25°C	-0.810	-1.850	-1.105	-1.475	-5.2	
Supply Vo	oltage = -5.2V								75°C	-0.720	-1.830	-1.045	-1.450	-5.2	
						MC12040				7507				0.14	
		Pin Under	0	°C	25	°C	75	°C		IESI	VOLTAGE	E APPLIED T	O PINS BEL	Ow	N
Symbol	Characteristics	Test	Min	Max	Min	Max	Min	Max	Unit	V <sub>IHmax</sub>	V <sub>ILmin</sub>	V <sub>IHAmin</sub>	VILAmax	VEE	(V <sub>CC</sub> Gnd
ΙE	Power Supply Drain	7			-120	-60			mAdc					7	1,14
IINH	Input Current	6 9				350 350			μAdc	6 9				7 7	1,14 1,14
V <sub>OH</sub> 1	Logic "1" Output Voltage	3 4 11 12	-1.000	-0.840	-0.960	-0.810	-0.900	-0.720	Vdc					7	1,14
V <sub>OL</sub> 1	Logic "0" Output Voltage	3 4 11 12	-1.870	-1.635	-1.850	-1.620	-1.830	-1.595	Vdc					7	1,14
V <sub>OHA</sub> <sup>2</sup>	Logic "1" Input Voltage	3 4 11 12	-1.020		-0.980		-0.920		Vdc			6.9		7	1,14
V <sub>OLA</sub> <sup>2</sup>	Logic "0" Input Voltage	3 4 11 12		-1.615		-1.600		-1.575	Vdc			9 6 9 6	6 9 6 9	7	1,14

		TEST	VOLTAGE VA	LUES	
		_	(Volts)	_	
@ Test Temperature	V <sub>IHmax</sub>	V <sub>ILmin</sub>	V <sub>IHAmin</sub>	V <sub>ILAmax</sub>	٧ <sub>EE</sub>
0°C	+4.160	+3.130	+3.855	+3.510	+5.0
25°C	+4.190	+3.150	+3.895	+3.525	+5.0
75°C	+4.280	+3.170	+3.955	+3.550	+5.0

Supply Vo	Supply Voltage = +5.0V         75°C										+3.170	+3.955	+3.550	+5.0		
			MC12040								TEST VOLTAGE APPLIED TO PINS BELOW					
		Pin Under	0°C		25°C		75°C			TEST VOLTAGE APPLIED TO PINS				0₩	(Vee)	
Symbol	Characteristics	Test	Min	Max	Min	Max	Min	Max	Unit	V <sub>IHmax</sub>	V <sub>ILmin</sub>	V <sub>IHAmin</sub>	V <sub>ILAmax</sub>	VEE	(V <sub>CC</sub> ) Gnd	
ΙE	Power Supply Drain	7			-115	-60			mAdc					1,14	7	
INH	Input Current	6 9				350 350			μAdc	6 9				1,14 1,14	7 7	
V <sub>OH</sub> 1	Logic "1" Output Voltage	3 4 11 12	4.000	4.160	4.040	4.190	4.100	4.280	Vdc					1,14	7	
V <sub>OL</sub> 1	Logic "0" Output Voltage	3 4 11 12	3.190	3.430	3.210	3.440	3.230	3.470	Vdc					1,14	7	
V <sub>OHA</sub> <sup>2</sup>	Logic "1" Input Voltage	3 4 11 12	3.980		4.020		4.080		Vdc			6.9		1,14	7	
V <sub>OLA</sub> <sup>2</sup>	Logic "0" Input Voltage	3 4 11 12		3.450		3.460		3.490	Vdc			9 6 9 6	6 9 6 9	1,14	7	

## MC12040

## 查询"MC12040"供应商

Pulse

Gen 1

Pulse

Gen 2

Output

Output

Waveform A

Waveform B

t+

t++

To Scope Channel A G T Pulse (0) Gen 1 PRF = 5.0MHz Duty Cycle = 50% t+ = t- = 1.5ns ±0.2ns Pulse Gen 2 О

90%

20ns

50%

- t++

50%

t+

50%

10%

10%

t-

80%

t+

20%

20%

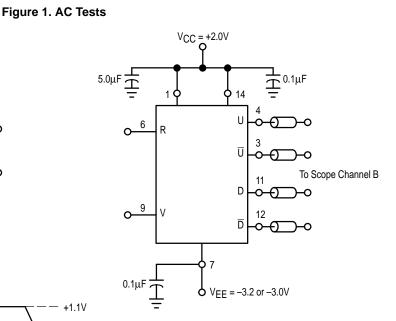
80%

90%

50%

t+

t+





+0.3V

+1.1V

+0.3V

- 1 All input and output cables to the scope are equal lengths of  $50\Omega$ coaxial cable.
- 2 Unused input and outputs are connected to a  $50\Omega$  resistor to ground.
- 3 The device under test must be preconditioned before performing the ac tests. Preconditioning may be accomplished by applying pulse generator 1 for a minimum of two pulses prior to pulse gen-erator 2. The device must be preconditioned again when inputs to pins 6 and 9 are interchanged. The same technique applies.

					MC12040	_		TEST VOLTAGES/WAVEFORMS				
				0°C	25°C	85°C		APP	LIED TO	PINS LIST	ED	
Symbol	Characteristic	Pin Under Test	Output Waveform	Мах	Мах	Мах	Unit	Pulse Gen 1	Pulse Gen 2	V <sub>EE</sub> –3.0 or –3.2V	V <sub>CC</sub> +2.0V	
$t_{6+4+}$ $t_{6+12+}$ $t_{6+3-}$ $t_{6+11-}$ $t_{9+11+}$ $t_{9+3+}$ $t_{9+12-}$ $t_{9+4-}$	Propagation Delay	6,4 6,12 6,3 6,11 9,11 9,3 9,12 9,4	B A B B A B B	4.6 6.0 4.5 6.4 4.6 6.0 4.5 6.4	4.6 6.0 4.5 6.4 4.6 6.0 4.5 6.4	5.0 6.6 4.9 7.0 5.0 6.6 4.9 7.0	ns	6 9 9 9 6 9 6	9696969	7	1,14	
t3+ t4+ t11+ t14+	Output Rise Time	3 4 11 14	A B B A	3.4	3.4	3.8	ns	6 6 9 9	996 66	7	1,14	
t3_ t4_ t <sub>11</sub> _ t <sub>14</sub> _	Output Fall Time	3 4 11 14	A B B A	3.4	3.4	3.8	ns	6 6 9 9	9 9 6 6	7	1,14	

## MC12040 APPLICATIONS INFORMATION

查证 MC120440 学供应该 network designed for use as a phase comparator for MECL-compatible input signals. It determines the "lead" or "lag" phase relationship and the time difference between the leading edges of the waveforms. Since these edges occur only once per cycle, the detector has a range of  $\pm 2\pi$  radians.

Operation of the device may be illustrated by assuming two waveforms, R and V (Figure 2), of the same frequency but differing in phase. If the logic had established by past history that R was leading V, the U output of the detector (pin 4) would produce a positive pulse width equal to the phase difference and the D output (pin 11) would simply remain low.

On the other hand, it is also possible that V was leading R (Figure 2), giving rise to a positive pulse on the D output and a constant low level on the U output pin. Both outputs for the sample condition are valid since the determination of lead or lag is dependent on past edge crossing and initial conditions at start–up. A stable phase–locked loop will result from either condition.

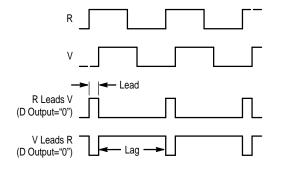
Phase error information is contained in the output duty cycle–that is, the ratio of the output pulse width to total period. By integrating or low–pass filtering the outputs of the detector and shifting the level to accommodate ECL swings, usable analog information for the voltage controlled oscillator can be developed. A circuit useful for this function is shown in Figure 3.

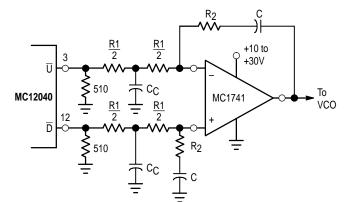
Figure 2. Timing Diagram

Proper level shifting is accomplished by differentially driving the operational amplifier from the normally high outputs of the phase detector (U and D). Using this technique the quiescent differential voltage to the operational amplifier is zero (assuming matched "1" levels from the phase detector). The U and D outputs are then used to pass along phase information to the operational amplifier. Phase error summing is accomplished through resistors R1 connected to the inputs of the operational amplifier. Some R–C filtering imbedded within the input network (NO TAG) may be very beneficial since the very narrow correctional pulses of the MC12040 would not normally be integrated by the amplifier. Phase detector gain for this configuration is approximately 0.16 volts/radian.

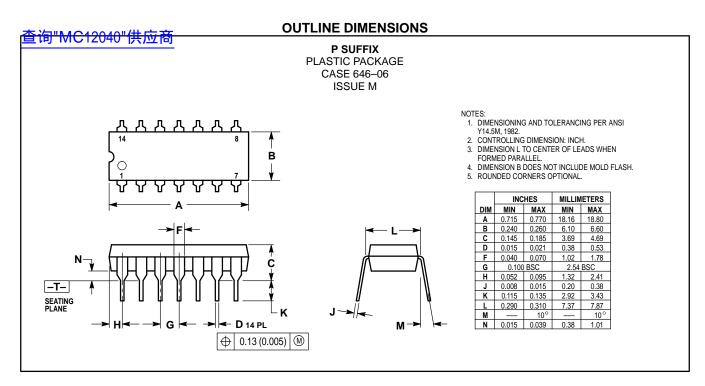
System phase error stems from input offset voltage in the operational amplifier, mismatching of nominally equal resistors, and mismatching of phase detector "high" states between the outputs used for threshold setting and phase measuring. All these effects are reflected in the gain constant. For example, a 16mV offset voltage in the amplifier would cause an error of 0.016/0.16 = 0.1 radian or 5.7 degrees of error. Phase error can be trimmed to zero initially by trimming either input offset or one of the threshold resistors (R1 in Figure 3). Phase error over temperature depends on how much the offending parameters drift.

#### Figure 3. Typical Filter and Summing Network





## MC12040



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