FAIRCHILD

SEMICONDUCTOR

MM74HC221A Dual Non-Retriggerable Monostable Multivibrator

Features

■ Typical propagation delay: 40 ns

■ Wide power supply range: 2V–6V

■ Low input current: 1 µA maximum

■ Simple pulse width formula T = RC

■ Wide pulse range: 400 ns to ∞ (typ)

■ Part to part variation: ±5% (typ)

■ Fanout of 10 LS-TTL loads

rise or fall times

■ Low quiescent current: 80 µA maximum (74HC Series)

Schmitt Trigger A & B inputs enable infinite signal input

General Description

The MM74HC221A high speed monostable multivibrators (one shots) utilize advanced silicon-gate CMOS technology. They feature speeds comparable to low power Schottky TTL circuitry while retaining the low power and high noise immunity characteristic of CMOS circuits.

Each multivibrator features both a negative, A, and a positive, B, transition triggered input, either of which can be used as an inhibit input. Also included is a clear input that when taken low resets the one shot. The MM74HC221A can be triggered on the positive transition of the clear while A is held LOW and B is held HIGH.

The MM74HC221A is a non-retriggerable, and therefore cannot be retriggered until the output pulse times out.

Pulse width stability over a wide range of temperature and supply is achieved using linear CMOS techniques. The output pulse equation is simply: PW = (R_{EXT}) (C_{EXT}); where PW is in seconds, R is in ohms, and C is in farads. All inputs are protected from damage due to static discharge by diodes to V_{CC} and ground.

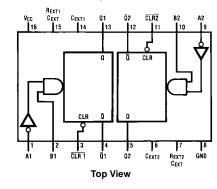
September 1983 Revised January 2004

MM74HC221A Dual Non-Retriggerable Monostable Multivibrator

Order Number	Package Number	Package Description
MM74HC221AM (Note 1)	M16A	16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150" Narrow
MM74HC221ASJ	M16D	16-Lead Small Outline Package (SOP), EIAJ TYPE II, 5.3mm Wide
MM74HC221AN	N16E	16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300" Wide

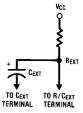


Ordering Code

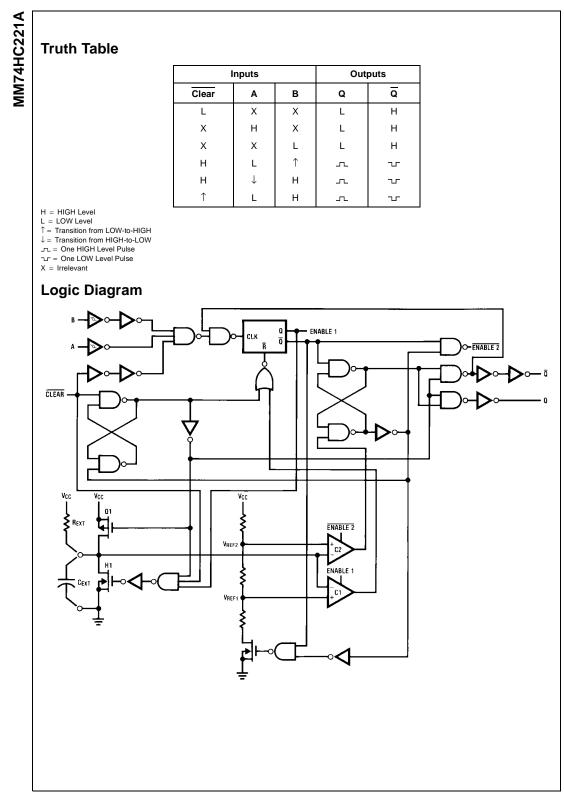


Note: Pin 6 and Pin 14 must be hard-wired to GND.

Timing Component



© 2004 Fairchild Semiconductor Corporation DS005325



www.fairchildsemi.com

2

(Note 3)

Absolute Maximum Ratings(Note 2)

Recommended Operating Conditions

Supply Voltage (V _{CC})	-0.5V to +7.0V	
DC Input Voltage (V _{IN})	–1.5V to $V_{CC} \mbox{+} 1.5 \mbox{V}$	Supply Voltag
DC Output Voltage (V _{OUT})	–0.5V to $V_{CC} \mbox{+} 0.5 \mbox{V}$	DC Input or O
Clamp Diode Current (I _{IK} , I _{OK})	±20 mA	(V _{IN} , V _{OUT})
DC Output Current, per pin (I _{OUT})	±25 mA	Operating Ten
DC V_{CC} or GND Current, per pin (I _{CC})	±50 mA	Maximum Inp
Storage Temperature Range (T _{STG})	$-65^{\circ}C$ to $+150^{\circ}C$	Time (Clear Ir
Power Dissipation (P _D)		Vc
(Note 4)	600 mW	Vc
S.O. Package only	500 mW	V _C
Lead Temperature (T _L) (Soldering 10 seconds)	260°C	Note 2: Maximur device may occur
(concerning to boothab)	200 0	Note 3: Unless of

	Min	Max	Units
Supply Voltage (V _{CC})	2	6	V
DC Input or Output Voltage			
(V _{IN} , V _{OUT})	0	V _{CC}	V
Operating Temperature Range (T_A)	-40	+85	°C
Maximum Input Rise and Fall			
Time (Clear Input)			
$V_{CC} = 2.0V$		1000	ns
$V_{CC} = 4.5V$		500	ns
$V_{CC} = 6.0V$		400	ns

MM74HC221A

Note 2: Maximum Ratings are those values beyond which damage to the device may occur.

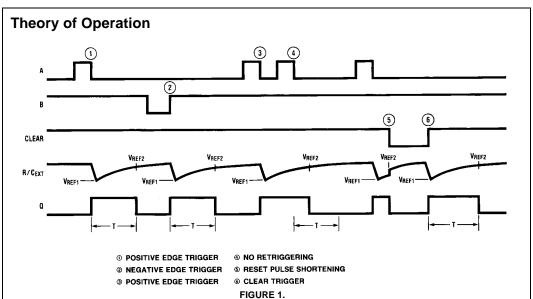
Note 3: Unless otherwise specified all voltages are referenced to ground. Note 4: Power Dissipation temperature derating — plastic "N" package: – 12 mW/°C from 65°C to 85°C.

DC Electrical Characteristics (Note 5)

Symbol	Parameter	Conditions	V _{cc}	T _A = 25°C		$T_A = -40$ to $85^{\circ}C$	$T_A = -55$ to $125^{\circ}C$	Units
Symbol	Parameter	Conditions	VCC	Тур		Guaranteed L	imits	Units
V _{IH}	Minimum HIGH Level		2.0V		1.5	1.5	1.5	
	Input Voltage		4.5V		3.15	3.15	3.15	۷ ۷ ۷ ۷ ۹ ۹
			6.0V		4.2	4.2	4.2	
V _{IL}	Maximum LOW Level		2.0V		0.3	0.3	0.3	
	Input Voltage		4.5V		0.9	0.9	0.9	V
			6.0V		1.2	1.2	1.2	
V _{OH}	Minimum HIGH Level	$V_{IN} = V_{IH}$ or V_{IL}						ν ν ν ν ν
	Output Voltage	$ I_{OUT} \le 20 \ \mu A$	2.0V	2.0	1.9	1.9	1.9	
			4.5V	4.5	4.4	4.4	4.4	
			6.0V	6.0	5.9	5.9	5.9	
		$V_{IN} = V_{IH} \text{ or } V_{IL}$						
		$ I_{OUT} \le 4.0 \text{ mA}$	4.5V	4.2	3.98	3.84	3.7	
		$ I_{OUT} \le 5.2 \text{ mA}$	6.0V	5.7	5.48	5.34	5.2	
V _{OL}	Maximum LOW Level	$V_{IN} = V_{IH}$ or V_{IL}						v
	Output Voltage	$ I_{OUT} \le 20 \; \mu A$	2.0V	0	0.1	0.1	0.1	
			4.5V	0	0.1	0.1	0.1	
			6.0V	0	0.1	0.1	0.1	
		$V_{IN} = V_{IH} \text{ or } V_{IL}$						
		$ I_{OUT} \le 4.0 \text{ mA}$	4.5V	0.2	0.26	0.33	0.4	
		$ I_{OUT} \le 5.2 \text{ mA}$	6.0V	0.2	0.26	0.33	0.4	
I _{IN}	Maximum Input Current	$V_{IN} = V_{CC}$ or GND	6.0V		±0.5	±5.0	±5.0	μΑ
	(Pins 7, 15)							۷ ۷ بلم بلم
I _{IN}	Maximum Input Current	$V_{IN} = V_{CC}$ or GND	6.0V		±0.1	±1.0	±1.0	μΑ
	(all other pins)							
I _{CC}	Maximum Quiescent Supply	$V_{IN} = V_{CC}$ or GND	6.0V		8.0	80	160	μΑ
	Current (standby)	$I_{OUT} = 0 \ \mu A$						
I _{CC}	Maximum Active Supply	$V_{IN} = V_{CC}$ or GND	2.0V	36	80	110	130	μΑ
	Current (per monostable)	$R/C_{EXT} = 0.5V_{CC}$	4.5V	0.33	1.0	1.3	1.6	mA
			6.0V	0.7	2.0	2.6	3.2	mA

Note 5: For a power supply of 5V \pm 10% the worst-case output voltages (V_{OH}, and V_{OL}) occur for HC at 4.5V. Thus the 4.5V values should be used when designing with this supply. Worst-case V_{IH} and V_{IL} occur at V_{CC} = 5.5V and 4.5V respectively. (The V_{IH} value at 5.5V is 3.85V.) The worst-case leakage current (I_{IN}, I_{CC}, and I_{O2}) occur for CMOS at the higher voltage and so the 6.0V values should be used.

$v_{CC} = 0$	5V, $T_A = 25^{\circ}C$, $C_L = 15 \text{ pF}$, t	$r = t_f = 6 \text{ ns}$							
Symt	bol Parame	eter		Condit	tions		Тур	uaranteed Limit	Uı
t _{PLH} Maximum Trigger Propagai Delay A, B or Clear to Q							22	36	r
t _{PHL}	Maximum Trigger Pro						25	42	r
"" TIL	Delay A, B or Clear to								
t _{PHL}	Maximum Propagation						20	31	r
t _{PLH}	Maximum Propagation						22	33	ı
tw	Minimum Pulse Width	•					14	26	
t _{REM}	Minimum Clear Remo							0	1
	Minimum Output Puls		C _{EXT} = 28	۶DF			400	-	r
			R _{EXT} = 2 k						
t _{WQ}	Output Pulse Width		C _{EXT} = 10				10		ļ
			$R_{EXT} = 10$						
-	Electrical Charac			11					
Symbol	Parameter	Conditi	ons	v _{cc}		25°C	$T_{A} = -40$ to 85°		125°u
					Тур	100	Guaranteed		
t _{PLH}	Maximum Trigger Propagation	1		2.0V	77	169	194	210	
	Delay A, B or Clear to Q			4.5V	26	42	51	57	
	Maximum Triagor Propagation			6.0V 2.0V	21 88	32 197	39 229	44 250	
t _{PHL}	Maximum Trigger Propagation Delay A, B or Clear to Q	1		2.0V 4.5V	88 29	48	60	250 67	
	Delay A, D UI Cicai IU Q			4.5V 6.0V	29 24	48 38	60 46	51	
t _{PHL}	Maximum Propagation			2.0V	54	114	132	143	
PHL	Delay Clear to Q			4.5V	23	34	41	45	
				6.0V	19	28	33	36	
t _{PLH}	Maximum Propagation			2.0V	56	116	135	147	
	Delay Clear to Q			4.5V	25	36	42	46	
				6.0V	20	29	34	37	
t _W	Minimum Pulse Width			2.0V	57	123	144	157	
	A, B, Clear			4.5V	17	30	37	42	
				6.0V	12	21	27	30	
t _{REM}	Minimum Clear			2.0V		0	0	0	
	Removal Time			4.5V		0	0	0	
				6.0V		0	0	0	
t _{TLH} , t _{THL}				2.0V 4.5V	30 8	75	95	110 22	
	Rise and Fall Time			4.5V 6.0V	8 7	15 13	19 16	19	
+	Minimum Output	C _{EXT} = 28 pF		2.0V		13	10	19	
^L WQ(MIN)	Pulse Width	$R_{EXT} = 26 \text{ pr}$ $R_{EXT} = 2 \text{ k}\Omega$		4.5V	1.5 450				
		$R_{EXT} = 6 k\Omega (V)$	$\sqrt{2} = 2V$	6.0V	380				
t _{WQ}	Output Pulse Width	C _{EXT} = 0.1 μF	Min	5.0V	1	0.9	0.86	0.85	
-1100		$R_{EXT} = 10 \ k\Omega$							
			Max	5.0V	1	1.1	1.14	1.15	
C _{PD}	Power Dissipation				87				
<u> </u>	Capacitance (Note 6)				40	20		20	
2 _{IN}	Maximum Input Capacitance (Pins 7 & 15)				12	20	20	20	
CIN	Maximum Input				6	10	10	10	



MM74HC221A

TRIGGER OPERATION

As shown in Figure 1 and the logic diagram before an input trigger occurs, the monostable is in the guiescent state with the Q output LOW, and the timing capacitor C_{EXT} completely charged to $\mathsf{V}_{\mathsf{CC}}.$ When the trigger input A goes from V_{CC} to GND (while inputs B and clear are held to V_{CC}) a valid trigger is recognized, which turns on comparator C1 and N-channel transistor N11. At the same time the output latch is set. With transistor N1 on, the capacitor C_{EXT} rapidly discharges toward GND until V_{REE1} is reached. At this point the output of comparator C1 changes state and transistor N1 turns off. Comparator C1 then turns off while at the same time comparator C2 turns on. With transistor N1 off, the capacitor $\mathbf{C}_{\mathsf{EXT}}$ begins to charge through the timing resistor, R_{EXT} , toward V_{CC} . When the voltage across C_{EXT} equals V_{REF2}, comparator C2 changes state causing the output latch to reset (Q goes LOW) while at the same time disabling comparator C2. This ends the timing cycle with the monostable in the quiescent state, waiting for the next triaaer.

A valid trigger is also recognized when trigger input B goes from GND to V_{CC} (while input A is at GND and input clear is at V_{CC}2). The MM74HC221 can also be triggered when clear goes from GND to V_{CC} (while A is at Gnd and B is at V_{CC}6).

It should be noted that in the quiescent state C_{EXT} is fully charged to V_{CC} causing the current through resistor R_{EXT} to be zero. Both comparators are "off" with the total device current due only to reverse junction leakages. An added feature of the MM74HC221 is that the output latch is set via the input trigger without regard to the capacitor voltage. Thus, propagation delay from trigger to Q is independent of the value of $C_{\text{EXT}}, R_{\text{EXT}},$ or the duty cycle of the input waveform.

The MM74HC221 is non-retriggerable and will ignore input transitions on A and B until it has timed out 3 and 4.

RESET OPERATION

These one shots may be reset during the generation of the output pulse. In the reset mode of operation, an input pulse on clear sets the reset latch and causes the capacitor to be fast charged to V_{CC} by turning on transistor Q1 5. When the voltage on the capacitor reaches V_{REF2}, the reset latch will clear and then be ready to accept another pulse. If the clear input is held LOW, any trigger inputs that occur will be inhibited and the Q and $\overline{\rm Q}$ outputs of the output latch will not change. Since the Q output is reset when an input low level is detected on the Clear input, the output pulse T can be made significantly shorter than the minimum pulse width specification.

