

# **Programmable Low Power Gyroscope**

## ADIS16260/ADIS16265

## Preliminary Technical Data 1250

#### **FEATURES**

Yaw rate gyroscope with range scaling ±80°/sec, ±160°/sec, and ±320°/sec settings SPI°-compatible serial interface Calibrated sensitivity and bias ADIS16260: +25°C ADIS16265: -40°C to +85°C

Operating temperature range: –40°C to +105°C
Digital temperature sensor output

In-system, auto-zero for bias drift calibration Digitally controlled sample rate, up to 2048SPS

Digitally controlled frequency response 50/300Hz Sensor bandwidth selection

Programmable Bartlett Window FIR filter
Dual alarm settings with configurable operation
Embedded integration for short-term angle estimates
Digitally activated self-test

Digitally activated low power mode
Interrupt-driven wake-up

Auxiliary 12-bit ADC input and 12-bit DAC output
Auxiliary digital input/output

Single-supply operation: 4.75 V to 5.25 V 2000 g powered shock survivability

#### **GENERAL DESCRIPTION**

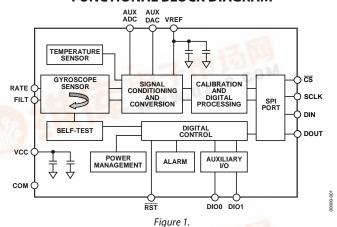
df.dzsc.com

The ADIS16260/ADIS16265 are complete angular rate measurement systems available in a single compact package enabled by Analog Devices, Inc. *i*Sensor™ integration. By enhancing Analog Devices *i*MEMS® sensor technology with an embedded signal processing solution, the ADIS16260/ADIS16265 provide factory-calibrated and tunable digital sensor data in a convenient format that can be accessed using a simple SPI serial interface. The ADIS16265 additionally provides an extended temperature calibration. The SPI interface provides access to measurements for the gyroscope, temperature, power supply, and one auxiliary analog input. Easy access to calibrated digital sensor data provides developers with a system-ready device, reducing development time, cost, and program risk.

#### **APPLICATIONS**

Instrumentation control
Platform control and stabilization
Motion control and analysis
Avionics instrumentation
Navigation
Image stabilization
Robotics

#### FUNCTIONAL BLOCK DIAGRAM



The device range can be digitally selected from three different settings: ±80°/sec, ±160°/sec, and ±320°/sec. Unique characteristics of the end system are accommodated easily through several built-in features, including a single-command auto-zero recalibration function, as well as configurable sample rate and frequency response. Additional features can be used to further reduce system complexity, including:

- Configurable alarm function
- Auxiliary 12-bit ADC and DAC
- Two configurable digital I/O ports
- Digital self-test function
- Digital sensor bandwidth selection

System power dissipation can be optimized via the ADIS16260/ ADIS16265 power management features, including an interrupt-driven wake-up. The ADIS16260/ADIS16265 are available in an  $11 \text{ mm} \times 11 \text{ mm} \times 5.5 \text{ mm}$ , laminate-based land grid array (LGA) package with a temperature range of  $-40^{\circ}\text{C}$  to  $+105^{\circ}\text{C}$ .

Rev. PrB
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10/06—Revision 0: Initial Version

## **SPECIFICATIONS**

 $T_A = -40$ °C to +105°C,  $V_{CC} = 5.0$  V, angular rate = 0°/sec,  $\pm 1$  g,  $\pm 320$ °/sec range setting, unless otherwise noted.

Table 1.

Parameter	Conditions	Min	Тур	Max	Unit
SENSITIVITY <sup>1</sup>	Clockwise rotation is positive output				
	25°C, dynamic range = $\pm 320^{\circ}/\text{sec}^2$		0.07326		°/sec/LSB
	25°C, dynamic range = ±160°/sec		0.03663		°/sec/LSB
	25°C, dynamic range = ±80°/sec		0.01832		°/sec/LSB
Initial Tolerance	25°C, dynamic range = ±320°/sec		±0.2	±1	%
Temperature Coefficient	ADIS16260		TBD		ppm/°C
	ADIS16265		25		ppm/°C
Nonlinearity	Best fit straight line		0.1		% of FS
BIAS					
In Run Bias Stability	25°C, 1σ		0.007		°/sec
Turn-On-to-Turn-On Bias Stability	25°C, 1σ		0.05		°/sec
Angular Random Walk	25°C, 1σ		2		°/√hour
Temperature Coefficient	ADIS16260		TBD		°/sec/°C
•	ADIS16265		0.005		°/sec/°C
Linear Acceleration Effect	Any axis		0.2		°/sec/g
Voltage Sensitivity	$V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$		1.0		°/sec/V
NOISE PERFORMANCE					
Output Noise	At 25°C, ±320°/sec range, no filtering		TBD		°/sec rms
	At 25°C, ±160°/sec range, 4-tap filter setting		TBD		°/sec rms
	At 25°C, ±80°/sec range, 16-tap filter setting		TBD		°/sec rms
Rate Noise Density	At 25°C, $f = 25$ Hz, $\pm 320^{\circ}$ /sec range, no filtering		0.05		°/sec/√Hz rms
FREQUENCY RESPONSE					,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,
3 dB Bandwidth	SENS_AVG[7] = 1		50		Hz
	SENS_AVG[7] = 0		300		
Sensor Resonant Frequency	52.15_7.115[2]		14		kHz
SELF-TEST STATE					
Change for Positive Stimulus	320°/sec dynamic range setting	TBD	TBD	TBD	LSB
Change for Negative Stimulus	320°/sec dynamic range setting	TBD	TBD	TBD	LSB
Internal Self-Test Cycle Time	320 / see dynamic range seeming	100	20		ms
ADC INPUT					
Resolution			12		Bits
Integral Nonlinearity			±2		LSB
Differential Nonlinearity			±1		LSB
Offset Error			±4		LSB
Gain Error			±2		LSB
Input Range		0	<u></u>	2.5	V
Input Capacitance	During acquisition		20	۷.5	pF
ON-CHIP VOLTAGE REFERENCE	During acquisition		2.5		V
Accuracy	At 25°C	-10	۷.5	+10	mV
Temperature Coefficient	N(2) C	-10	±40	Ŧ10	ppm/°C
Output Impedance			±40 70		Ω
Output impedance			70	_	1.2

Parameter	Conditions	Min	Тур	Max	Unit
DAC OUTPUT	5 kΩ/100 pF to GND				
Resolution			12		Bits
Relative Accuracy	For Code 101 to Code 4095		4		LSB
Differential Nonlinearity			1		LSB
Offset Error			±5		mV
Gain Error			±0.5		%
Output Range			0 to 2.5		V
Output Impedance			2		Ω
Output Settling Time			10		μs
LOGIC INPUTS	Internal 3.3 V Interface				•
Input High Voltage, V <sub>INH</sub>		2.0			V
Input Low Voltage, V <sub>INL</sub>				0.8	V
	For $\overline{CS}$ signal when used to wake up from sleep mode			0.55	V
Logic 1 Input Current, I <sub>INH</sub>	V <sub>H</sub> = 3.3 V		±0.2	±10	μA
Logic 0 Input Current, IINL	$V_{IL} = 0.5$ $V$				,
All except RST			-40	-60	μΑ
RST <sup>3</sup>			-1	00	mA
			10		pF
Input Capacitance, C <sub>IN</sub> DIGITAL OUTPUTS	Internal 3.3 V Interface		10		рг
		2.4			
Output High Voltage, VoH	Isource = 1.6 mA	2.4		0.4	V
Output Low Voltage, Vol	I <sub>SINK</sub> = 1.6 mA			0.4	V
SLEEP TIMER		0.5		400	
Timeout Period <sup>4</sup>		0.5		128	sec
START-UP TIME					
Initial			160		ms
Sleep Mode Recovery			2.5		ms
FLASH MEMORY					
Endurance <sup>5</sup>		10,000			Cycles
Data Retention <sup>6</sup>	T <sub>J</sub> = 55°C	20			Years
CONVERSION RATE					
Minimum Conversion Time			0.488		ms
Maximum Conversion Time			7.75		sec
Maximum Throughput Rate			2048		SPS
Minimum Throughput Rate			0.129		SPS
POWER SUPPLY		_			
Operating Voltage Range, V <sub>CC</sub>		4.75	5.0	5.25	V
Power Supply Current	Normal mode at 25°C		TBD		mA
•	Fast mode at 25°C		TBD		mA
	Sleep mode at 25°C		TBD		μΑ

 $<sup>^1</sup>$  ADIS16265 characterization data represents  $\pm 4\sigma$  to fall within the  $\pm 1\%$  limit.

 $<sup>^2</sup>$  The <u>sens</u>or is capable of  $\pm 600^\circ$ /sec, but the specifications herein are for  $\pm 320^\circ$ /sec only.

 $<sup>^3</sup>$  The  $\overline{\text{RST}}$  pin has an internal pull-up.

<sup>&</sup>lt;sup>4</sup> Guaranteed by design.

<sup>&</sup>lt;sup>5</sup> Endurance is qualified as per JEDEC Standard 22 Method A117 and measured at -40°C, +25°C, +85°C, and +125°C.

<sup>&</sup>lt;sup>6</sup> Retention lifetime equivalent at junction temperature (T.) 55°C, as per JEDEC Standard 22 Method A117. Retention lifetime decreases with junction temperature.

#### **TIMING SPECIFICATIONS**

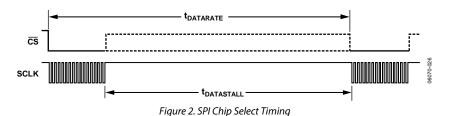
 $T_A = -40$ °C to +85°C,  $V_{CC} = 5.0$  V, unless otherwise noted.

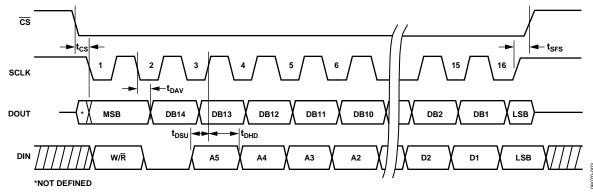
Table 2.

Parameter	Description	Min <sup>1</sup>	Тур	Max <sup>1</sup>	Unit
f <sub>SCLK</sub>	Fast mode, SMPL_PRD $\leq$ 0x07 (f <sub>s</sub> $\geq$ 64 Hz)	0.01		2.5	MHz
	Normal mode, SMPL_PRD $\geq$ 0x08 (f <sub>s</sub> $\leq$ 56.9 Hz)	0.01		1.0	MHz
<b>t</b> DATARATE	Data rate period, fast mode, SMPL_PRD $\leq$ 0x07 (f <sub>S</sub> $\geq$ 64 Hz)	32			μs
	Data rate period, normal mode, SMPL_PRD $\geq$ 0x08 (f <sub>s</sub> $\leq$ 56.9 Hz)	42			μs
<b>t</b> stall	Stall period, fast mode, SMPL_PRD $\leq$ 0x07 (f <sub>S</sub> $\geq$ 64 Hz)	9			μs
	Stall period, normal mode, SMPL_PRD $\geq$ 0x08 (f <sub>s</sub> $\leq$ 56.9 Hz)	12			μs
<b>t</b> cs	Chip select to clock edge	48.8			ns
t <sub>DAV</sub>	Data output valid after SCLK falling edge <sup>2</sup>			100	ns
t <sub>DSU</sub>	Data input setup time before SCLK rising edge	24.4			ns
t <sub>DHD</sub>	Data input hold time after SCLK rising edge	48.8			ns
t <sub>DF</sub>	Data output fall time		5	12.5	ns
t <sub>DR</sub>	Data output rise time		5	12.5	ns
t <sub>SFS</sub>	CS high after SCLK edge <sup>3</sup>	5			ns
	Flash update time (power supply must be within range)	50			ms

<sup>&</sup>lt;sup>1</sup> Guaranteed by design; not production tested.

<sup>&</sup>lt;sup>3</sup> This parameter may need to be expanded to allow for proper capture of the LSB. After CS goes high, the DOUT line goes into a high impedance state.





Figure~3.~SPI~Timing~(Using~SPI~Settings~Typically~Identified~as~Phase=1, Polarity=1)

<sup>&</sup>lt;sup>2</sup> The MSB presents an exception to this parameter. The MSB clocks out on the falling edge of CS. The rest of the DOUT bits are clocked after the falling edge of SCLK and are governed by this specification.

## **ABSOLUTE MAXIMUM RATINGS**

Table 3.

1 110 10 00	
Parameter	Rating
Acceleration (Any Axis, Unpowered, 0.5 ms)	2000 g
Acceleration (Any Axis, Powered, 0.5 ms)	2000 <i>g</i>
V <sub>CC</sub> to COM	-0.3 V to +6.0 V
Digital Input/Output Voltage to COM	−0.3 V to +5.5 V
Analog Inputs to COM	-0.3 V to +3.5 V
Operating Temperature Range <sup>1</sup>	-40°C to +105°C
Storage Temperature Range <sup>1</sup>	−65°C to +150°C

<sup>&</sup>lt;sup>1</sup> Extended exposure to temperatures outside of the specified temperature range of  $-40^{\circ}$ C to  $+85^{\circ}$ C can adversely affect the accuracy of the factory calibration. For best accuracy, store the parts within the specified operating range of  $-40^{\circ}$ C to  $+85^{\circ}$ C.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

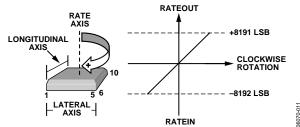


Figure 4. RATEOUT Level Increase with Clockwise Rotation Increase

#### **ESD CAUTION**



**ESD** (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

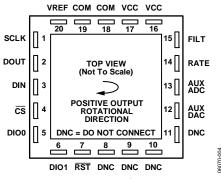


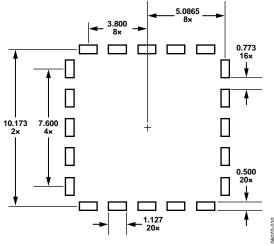
Figure 5. Pin Configuration

**Table 4. Pin Function Descriptions** 

Pin No.	Mnemonic	Type <sup>1</sup>	Description
1	SCLK	1	SPI, Serial Clock.
2	DOUT	0	SPI, Data Output.
3	DIN	1	SPI, Data Input.
4	<u>CS</u>	1	SPI, Chip Select, Active Low.
5, 6	DIO0, DIO1	I/O	Multifunction Digital Input/Output Pin.
7	RST	1	Reset, Active Low. This resets the sensor signal conditioning circuit and initiates a start-up sequence.
8, 9, 10, 11	DNC	-	Do Not Connect.
12	AUX DAC	0	Auxiliary DAC Analog Output Voltage.
13	AUX ADC	1	Auxiliary ADC Analog Input Voltage.
14	RATE	0	Analog Rate Signal Output (Uncalibrated).
15	FILT	1	Analog Amplifier Summing Junction. This is used for setting the analog bandwidth. See the Analog Bandwidth section for more details.
16, 17	VCC	S	5.0 V Power Supply.
18, 19	COM	S	Common. Reference point for all circuitry in the ADIS16260/ADIS16265.
20	VREF	0	Precision Reference Output.

<sup>&</sup>lt;sup>1</sup> S = supply; O = output; I = input.

#### **RECOMMENDED LAYOUT**



11mm × 11mm STACKED LGA PACKAGE

Figure 6. Recommended Pad Layout (Units in Millimeters)

## TYPICAL PERFORMANCE CHARACTERISTICS

# BD

Figure 7. Root Allan Variance vs. TAU, ±320°/sec Range

# THEORY OF OPERATION OVERVIEW

The core angular rate sensor integrated inside the ADIS16260/ ADIS16265 is based on the Analog Devices *i*MEMS technology. This sensor operates on the principle of a resonator gyroscope. Two polysilicon sensing structures each contain a dither frame electrostatically driven to resonance. This provides the necessary velocity element to produce a Coriolis force during rotation. At two of the outer extremes of each frame, orthogonal to the dither motion, are movable fingers placed between fixed fingers to form a capacitive pickoff structure that senses Coriolis motion. The resulting signal is fed to a series of gain and demodulation stages that produce the electrical rate signal output.

The base sensor output signal is sampled using an ADC, and then the digital data is fed into a proprietary digital calibration circuit. This circuit contains calibration coefficients from the factory calibration, along with user-defined calibration registers that can be used to calibrate system-level errors.

The calibrated gyroscope data (GYRO\_OUT) is made available through output data registers along with temperature, power supply, auxiliary ADC, and relative angle output calculations.

#### **RELATIVE ANGLE ESTIMATE**

The ANGL\_OUT register offers the integration of the GYRO\_OUT data. In order for this information to be useful, the reference angle must be known. This can be accomplished by reading the register contents at the initial time, before starting the monitoring, or by setting its contents to zero. This number is reset to zero when the NULL command is used, after a RESET command is used, and during power-up. This function can be used to estimate change in angle over a period. The user is cautioned to fully understand the stability requirements and the time period over which to use this estimated relative angle position.

#### **FACTORY CALIBRATION**

The ADIS16260/ADIS16265 provide a factory calibration that includes correction for initial tolerance and power supply variation. In addition, the ADIS16265 provides correction for temperature variation. This calibration includes individual sensor characterization and custom correction coefficient calculation.

#### **AUXILIARY ADC FUNCTION**

The auxiliary ADC function integrates a standard 12-bit ADC into the ADIS16260/ADIS16265 to digitize other system-level analog signals. The output of the ADC can be monitored through the AUX\_ADC control register, as defined in Table 6. The ADC is a 12-bit successive approximation converter. The output data is presented in straight binary format with the full-scale range extending from 0 V to 2.5 V. The 2.5 V upper limit is derived from the on-chip precision internal reference.

Figure 8 shows the equivalent circuit of the analog input structure of the ADC. The input capacitor (C1) is typically 4 pF and can be attributed to parasitic package capacitance. The two diodes provide ESD protection for the analog input. Care must be taken to ensure that the analog input signals never exceed the range of -0.3~V to +3.5~V. This causes the diodes to become forward-biased and to start conducting. The diodes can handle 10 mA without causing irreversible damage. The resistor is a lumped component that represents the on resistance of the switches. The value of this resistance is typically  $100~\Omega$ . Capacitor C2 represents the ADC sampling capacitor and is typically  $16~\rm pF$ .

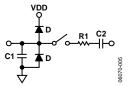


Figure 8. Equivalent Analog Input Circuit Conversion Phase: Switch Open Track Phase: Switch Closed

For ac applications, it is recommended to remove high frequency components from the analog input signal by using a low-pass filter on the analog input pin.

In applications where harmonic distortion and signal-to-noise ratio are critical, the analog input must be driven from a low impedance source. Large source impedances significantly affect the ac performance of the ADC. This can necessitate the use of an input buffer amplifier. When no input amplifier is used to drive the analog input, the source impedance should be limited to values lower than  $1~\mathrm{k}\Omega$ .

#### BASIC OPERATION

The ADIS16260/ADIS16265 are designed for simple integration into industrial system designs, requiring only a 5.0 V power supply and a 4-wire, industry standard serial peripheral interface (SPI). All outputs and user-programmable functions are handled by a simple register structure. Each register is 16 bits in length and has its own unique bit map. The 16 bits in each register consist of an upper (D8 to D15) byte and a lower (D0 to D7) byte, each of which has its own 6-bit address.

#### **SERIAL PERIPHERAL INTERFACE (SPI)**

The ADIS16260/ADIS16265 serial peripheral interface (SPI) port includes four signals: chip select ( $\overline{\text{CS}}$ ), serial clock (SCLK), data input (DIN), and data output (DOUT). The  $\overline{\text{CS}}$  line enables the ADIS16260/ADIS16265 SPI port and frames each SPI event, which consists of single or multiple data frames. When this signal is high, the DOUT lines are in a high impedance state and the signals on DIN and SCLK have no impact on operation. A complete data frame contains 16 clock cycles. Because the SPI port operates in full duplex mode, it supports simultaneous, 16-bit receive (DIN) and transmit (DOUT) functions during the same data frame.

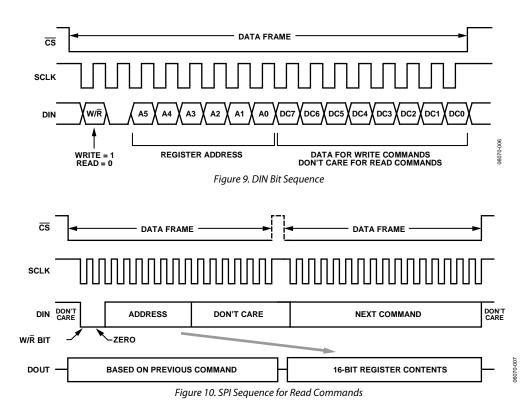
Refer to Table 2, Figure 2, and Figure 3 for detailed timing and operation of the SPI port.

#### **Writing to Registers**

Figure 9 displays a typical data frame for writing a command to a control register. In this case, the first bit of the DIN sequence is a 1, followed by a 0, the 6-bit address, and the 8-bit data command. Because each write command covers a single byte of data, two data frames are required when writing the entire 16-bit space of a register.

#### **Reading from Registers**

Reading the contents of a register requires a modification to the sequence in Figure 9. In this case, the first two bits in the DIN sequence are 0, followed by the address of the register. Each register has two addresses (upper, lower), but either one can be used to access its entire 16 bits of data. The final eight bits of the DIN sequence are irrelevant and can be counted as don't cares during a read command. During the next data frame, the DOUT sequence contains the register's 16-bit data, as shown in Figure 10. Although a single read command requires two separate data frames, the full duplex mode minimizes this overhead, requiring only one extra data frame when continuously sampling.



#### Preliminary Technical Data 查询"ADIS16260BCCZ"供应商

#### **DATA OUTPUT REGISTER ACCESS**

The ADIS16260/ADIS16265 provide access to calibrated rotation measurements, relative angle estimates, power supply measurements, temperature measurements, and an auxiliary 12-bit ADC channel. This output data is continuously updating internally, regardless of user read rates. The following bit map describes the structure of all output data registers, except ENDURANCE, in the ADIS16260/ADIS16265.

Table 5. Register Bit Map

MSB							LSB
ND	EA	D13	D12	D11	D10	D9	D8
D7	D6	D5	D4	D3	D2	D1	D0

Bit D13 and Bit D12 are assigned don't care status.

The output data register map is located in Table 6 and provides all of the necessary details for accessing each register's data.

Table 7 displays the output coding for the GYRO\_OUT register.

Figure 11 provides an example SPI read cycle for this register.

The MSB holds the new data (ND) indicator. When the output registers are updated with new data, the ND bit goes to a 1 state.

After the output data is read, it returns to a 0 state. The EA bit is

result from a number of conditions, such as a power supply that

used to indicate a system error or an alarm condition that can

Diagnostics section for more details. The output data is either

12 bits or 14 bits in length. For all of the 12-bit output data,

is out of the specified operating range. See the Status and

**Table 6. Data Output Register Information** 

Name	Function	Address	Resolution (Bits)	Data Format	Scale Factor (per LSB)					
ENDURANCE	Flash Memory Write Counter	0x01, 0x00	16	Binary	1 count					
SUPPLY_OUT	Power Supply Data	0x03, 0x02	12	Binary	1.8315 mV					
GYRO_OUT	Gyroscope Data	0x05, 0x04	14	Twos Complement	0.07326°/sec <sup>1</sup>					
AUX_ADC	Auxiliary Analog Input Data	0x0B, 0x0A	12	Binary	0.6105 mV					
TEMP_OUT	Sensor Temperature Data	0x0D, 0x0C	12	Twos Complement	0.1453℃					
ANGL_OUT	Angle Output	0x0F, 0x0E	14	Binary	0.03663°					

<sup>&</sup>lt;sup>1</sup> Assumes that the scaling is set to 320°/sec.

Table 7. Output Coding Example, GYRO\_OUT<sup>1,2</sup>

	Rate of Rotation				
±320°/sec Range	±160°/sec Range	±80°/sec Range	Binary Output	Hex Output	Decimal
600°/sec	300°/sec	150°/sec	01 1111 1111 1111	0x1FFF	8191
320°/sec	160°/sec	80°/sec	01 0001 0001 0000	0x1110	4368
80°/sec	40°/sec	20°/sec	00 0100 0100 0100	0x0444	1092
40°/sec	20°/sec	10°/sec	00 0010 0010 0010	0x0222	546
0.07326°/sec	0.03663°/sec	0.018315°/sec	00 0000 0000 0001	0x0001	1
0°/sec	0°/sec	0°/sec	00 0000 0000 0000	0x0000	0
-0.07326°/sec	-0.03663°/sec	-0.018315°/sec	11 1111 1111 1111	0x3FFF	-1
-40°/sec	-20°/sec	-10°/sec	11 1101 1101 1110	0x3DDE	-546
-80°/sec	-40°/sec	-20°/sec	11 1011 1011 1100	0x3BBC	-1092
-320°/sec	-160°/sec	-80°/sec	10 1110 1111 0000	0x2EF0	-4368
-600°/sec	-300°/sec	-150°/sec	10 0000 0000 0000	0x2000	-8192

 $<sup>^{\</sup>mbox{\tiny 1}}$  Two MSBs have been masked off and are not considered in the coding.

<sup>&</sup>lt;sup>2</sup> Nominal sensitivity and zero offset null performance are assumed.

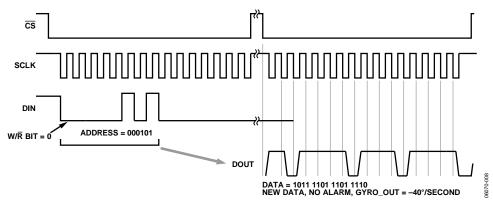


Figure 11. Example Read Cycle, ±320°/sec Setting

# PROGRAMMING AND CONTROL CONTROL REGISTER OVERVIEW

The ADIS16260/ADIS16265 offer many programmable features controlled by writing commands to the appropriate control registers using the SPI. Table 8 provides a summary of these control registers, which controls the operation of the following parameters:

- Calibration
- Global commands
- Operational control
  - Sample rate
  - · Power management
  - Digital filtering
  - Dynamic range
  - DAC output
  - Digital I/O
- Operational status and diagnostics
  - Self-test
  - Status conditions
  - Alarms

#### **CONTROL REGISTER STRUCTURE**

The ADIS16260/ADIS16265 uses a temporary, RAM-based memory structure to facilitate the control registers displayed in Table 8. The start-up configuration is stored in a flash memory structure that automatically loads into the control registers during the start-up sequence. Each nonvolatile register has a corresponding flash memory location, for storing the latest configuration contents. Since flash memory has endurance limitations, the contents of each nonvolatile register must be manually stored to flash (note that the contents of the control register contents are only nonvolatile when they are stored to flash). The manual flash update command, made available in the COMMAND register, provides this function. The ENDURANCE register provides a counter that allows for memory reliability management against the 20,000-write cycle specification.

Table 8. Control Register Memory Map

Register Name	Туре	Volatility	ility Address Bytes Function		Reference Table	
GYRO_OFF	R/W	Nonvolatile	0x15, 0x14	2	Gyroscope bias offset factor	Table 9, Table 10
GYRO_SCALE	R/W	Nonvolatile	0x17, 0x16	2	Gyroscope scale factor	Table 11, Table 12
			0x18 to 0x1F	8	Reserved	
ALM_MAG1	R/W	Nonvolatile	0x21, 0x20	2	Alarm 1 amplitude threshold and polarity	Table 31, Table 32
ALM_MAG2	R/W	Nonvolatile	0x23, 0x22	2	Alarm 2 amplitude threshold and polarity	Table 35, Table 36
ALM_SMPL1	R/W	Nonvolatile	0x25, 0x24	2	Alarm 1 sample period	Table 33, Table 34
ALM_SMPL2	R/W	Nonvolatile	0x27, 0x26	2	Alarm 2 sample period	Table 37, Table 38
ALM_CTRL	R/W	Nonvolatile	0x29, 0x28	2	Alarm control register	Table 39, Table 40
			0x2A to 0x2F	6	Reserved	
AUX_DAC	R/W	Volatile	0x31, 0x30	2	Auxiliary DAC data	Table 21, Table 22
GPIO_CTRL	R/W	Volatile	0x33, 0x32	2	Auxiliary digital I/O control register	Table 23, Table 24
MSC_CTRL	R/W	Nonvolatile <sup>1</sup>	0x35, 0x34	2	Miscellaneous control register	Table 26, Table 27
SMPL_PRD	R/W	Nonvolatile	0x37, 0x36	2	ADC sample period control	Table 15, Table 16
SENS/AVG	R/W	Nonvolatile	0x39, 0x38	2	Defines the dynamic range (sensitivity setting) and the number of taps for the digital filter	Table 19, Table 20
SLP_CNT	R/W	Volatile	0x3B, 0x3A	2	Counter used to determine length of power-down mode	Table 17, Table 18
STATUS	R	Volatile	0x3D, 0x3C	2	System status register	Table 28, Table 29
COMMAND	W	N/A	0x3F, 0x3E	2	System command register	Table 13, Table 14

<sup>&</sup>lt;sup>1</sup> The contents of the upper byte are nonvolatile; the contents of the lower byte are volatile.

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#### **CALIBRATION**

The ADIS16260/ADIS16265 are factory-calibrated for sensitivity and bias. It also provides several user calibration functions for simplifying field-level corrections. The calibration factors are stored in nonvolatile memory and are applied using the following linear calibration equation:

$$y = mx + b$$

where:

y is the calibrated output data.
x is the precalibration data.
m is the sensitivity scale factor.
b is the offset scale factor.

There are three options for system-level calibrations of the bias in the ADIS16260/ADIS16265: auto-null, factory calibration restore, and manual calibration updates. The auto-null and factory reset options are described in the Global Commands section. Optional field-level calibrations use the preceding equation and require two steps:

- 1. Characterize the behavior of the ADIS16260/ADIS16265 at predefined critical operating conditions.
- 2. Use this characterization data to calculate and load the contents of GYRO\_OFF (b) and GYRO\_SCALE (m).

The GYRO\_OFF provides a calibration range of ±37.5°/sec, and its contents are nonvolatile. The GYRO\_SCALE register provides a calibration range of 0 to 1.9995, and its contents are also nonvolatile.

Table 9. GYRO\_OFF Register Definition

Address	Scale <sup>1</sup>	Default	Format	Access
0x15,	0.018315°/sec	0x0000	Twos	R/W
0x14			complement	

<sup>&</sup>lt;sup>1</sup> Scale is the weight of each LSB.

Table 10. GYRO\_OFF Bit Descriptions

Bit	Description
15:12	Not used
11:0	Data bits

Table 11. GYRO\_SCALE Register Definition

Address	Scale <sup>1</sup>	Default <sup>2</sup>	Format	Access
0x17, 0x16	0.0487%	0x0800	Binary	R/W

<sup>&</sup>lt;sup>1</sup> Scale is the weight of each LSB.

Table 12. GYRO SCALE Bit Descriptions

Bit	Description
15:12	Not used
11:0	Data bits

#### **GLOBAL COMMANDS**

The ADIS16260/ADIS16265 provide global commands for common operations such as auto-null, factory calibration restore, manual flash update, auxiliary DAC latch, and software reset. Each of these global commands has a unique control bit assigned to it in the COMMAND register and is initiated by writing a 1 to its assigned bit.

The auto-null function does two things: it resets the contents of the ANGL\_OUT register to zero, and it adjusts the GYRO\_OUT register to zero. This automated adjustment takes two steps:

- 1. Read GYRO\_OUT.
- Write the opposite of this value into the GRYO\_OFF register. Sensor noise influences the accuracy of this step.

For optimal calibration accuracy, set the number of filtering taps to its maximum, wait for the appropriate number of samples to process through the filter, and then exercise this option.

The factory calibration restore command sets the contents of GYRO\_OFF to 0x0000 and GYRO\_SCALE to 0x0800, erasing any field-level calibration contents. The manual flash update writes the contents of each nonvolatile register into flash memory for storage. This process takes approximately 50 ms and requires the power supply voltage to be within specification for the duration of the event. It is worth noting that this operation also automatically follows the auto-null and factory reset commands.

The DAC latch command loads the contents of AUX\_DAC into the DAC latches. Since the AUX\_DAC contents must be updated one byte at a time, this command ensures a stable DAC output voltage during updates. Finally, the software reset command sends the ADIS16260/ADIS16265 digital processor into a restart sequence, effectively doing the same thing as the RST line.

**Table 13. COMMAND Register Definition** 

Address	Default	Format	Access
0x3F, 0x3E	N/A	N/A	Write only

**Table 14. COMMAND Bit Descriptions** 

Bit	Description
15:8	Not used
7	Software reset command
6:4	Not used
3	Manual flash update command
2	Auxiliary DAC data latch
1	Factory calibration restore command
0	Auto-null command

<sup>&</sup>lt;sup>2</sup> Equates to a scale factor of one.

#### **OPERATIONAL CONTROL**

#### Internal Sample Rate

The internal sample rate defines how often data output variables are updated, independent of the rate at which they are read out on the SPI port. The SMPL\_PRD register controls the ADIS16260/ADIS16265 internal sample rate and has two parts: a selectable time base and a multiplier. When SMPL\_PRD is greater than zero, the sample period can be calculated using the following equation:

$$T_S = T_B \times (N_S + 1)$$

where:

 $T_s$  is the sample period.

 $T_B$  is the time base.

 $N_S$  is the increment setting.

When SMPL\_PRD[7:0] = 0x00, the internal sample setting is 2048SPS. Since this mode bypasses an internal filtering network, it will result in a total noise increase of approximately 25%. This SENS\_AVG register provides opportunity for noise reduction. The default value is the maximum 256 samples per second, and the contents of this register are nonvolatile.

Table 15. SMPL\_PRD Register Definition

Address	Default	Format	Access
0x37, 0x36	0x0001	N/A	R/W

Table 16. SMPL\_PRD Bit Descriptions

Bit	Description
15:8	Not used
7	Time base, 0 = 1.953 ms, 1 = 60.54 ms
6:0	Multiplier

The following is an example calculation of the sample period for the ADIS16260/ADIS16265:

If 
$$SMPL\_PRD = 0x0007$$
,  $B7...B0 = 00000111$   
 $B7 = 0 \rightarrow T_B = 1.953$  ms  
 $B6...B0 = 000000111 \rightarrow N_S = 7$   
 $T_S = T_B \times (N_S + 1) = 1.953$  ms  $\times (7 + 1) = 15.624$  ms  
 $f_S = 1/T_S = 64$  SPS

The sample rate setting has a direct impact on the SPI data rate capability. For sample rates of 64 SPS and above, the SPI SCLK can run at a rate up to 2.5 MHz. For sample rates below 64 SPS, the SPI SCLK can run at a rate up to 1 MHz.

The sample rate setting also affects the power dissipation. When the sample rate is set below 64 SPS, the power dissipation reduces by a factor of 60%. The two different modes of operation offer a system-level trade-off between performance (sample rate, serial transfer rate) and power dissipation.

#### **Power Management**

In addition to offering two different performance modes for power optimization, the ADIS16260/ADIS16265 offer a programmable shutdown period. Writing the appropriate sleep

time to the SLP\_CNT register shuts the device down for the specified time. The following example provides an illustration of this relationship:

$$B7 \dots B0 = 00000110$$

$$Sleep period = 3 sec$$

After completing the sleep period, the ADIS16260/ADIS16265 return to normal operation. If measurements are required before sleep period completion, the ADIS16260/ADIS16265 can be awakened by putting the  $\overline{\text{CS}}$  line in a zero logic state. Otherwise, the  $\overline{\text{CS}}$  line must be kept high to maintain sleep mode.

Table 17. SLP\_CNT Register Definition

Address	Scale <sup>1</sup>	Default	Format	Access
0x3B, 0x3A	0.5 sec	0x0000	Binary	R/W

<sup>&</sup>lt;sup>1</sup> Scale is the weight of each LSB.

Table 18. SLP\_CNT Bit Descriptions

Bit	Description
15:8	Not used
7:0	Data bits

#### **Analog Bandwidth**

The analog bandwidth of the ADIS16260/ADIS16265 is 50 Hz. This bandwidth can be reduced by placing an external capacitor across the RATE and FILT pins. In this case, the analog bandwidth can be calculated using the following equation:

$$f_{OUT} = 1/(2 \times \pi \times R_{OUT} \times (C_{OUT} + C_{IN}))$$

where:

 $R_{OUT} = 90 \text{ k}\Omega.$ 

 $C_{OUT}$  is the external capacitance.

 $CIN = 0.0377\mu F$  when  $SENS\_AVG[7] = 0$ 

 $CIN = 0.0047\mu F$  when  $SENS\_AVG[7] = 1$ 

#### **Digital Filtering**

The ADIS16260/ADIS16265 GYRO\_OUT signal path has a nominal analog bandwidth of 50 Hz. The ADIS16260 provides a Bartlett Window FIR filter for additional noise reduction on all of the output data registers. The SENS/AVG register stores the number of taps in this filter in seven power-of-two step sizes (that is,  $2^{M} = 1, 2, 4, 16, 32, 64$ , and 128). Filter setup requires one simple step: write the appropriate M factor to the assigned bits in the SENS/AVG register. The bit assignments are listed in Table 20. The following equation offers a frequency response relationship for this filter:

$$H_B(f) = H_A^2(f) \Rightarrow H_A(f) = \frac{\sin(\pi \times N \times f \times t_S)}{N \times \sin(\pi \times f \times t_S)}$$

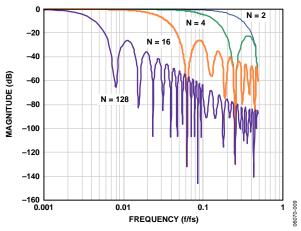


Figure 12. Bartlett Window FIR Frequency Response

#### **Dynamic Range**

The ADIS16260/ADIS16265 provide three dynamic range settings:  $\pm 80^{\circ}/\text{sec}$ ,  $\pm 160^{\circ}/\text{sec}$ , and  $\pm 320^{\circ}/\text{sec}$ . The lower dynamic range settings (80, 160) limit the minimum filter tap sizes in order to maintain the resolution as the maximum rate measurements decrease. The recommended order for programming the SENS/AVG register is (1) dynamic range and then (2) filtering response. The contents of the SENS/AVG register are nonvolatile.

Table 19. SENS/AVG Register Definition

Address	Default	Format	Access
0x39, 0x38	0x0402	Binary	R/W

Table 20. SENS/AVG Bit Descriptions

		<u>.                                      </u>
Bit	Value	Description
15:11		Not used
10:8		Sensitivity selection bits
	100	320°/sec (default condition)
	010	160°/sec, filter taps ≥ 4 (Bits[3:0] ≥ 0x02)
	001	80°/sec, filter taps ≥16 (Bits[3:0] ≥ 0x04)
7		Sensor bandwidth selection:
		1 = 300Hz, 0 = 50Hz
6:4		Not used
3:0		Filter tap setting, M = binary number
		(number of taps, $N = 2^{M}$ )

#### **Auxiliary DAC**

The auxiliary DAC provides a 12-bit level adjustment function. The AUX\_DAC register controls the operation of this feature. It offers a rail-to-rail buffered output that has a range of 0 V to 2.5 V. The DAC can drive its output to within 5 mV of the ground reference when it is not sinking current. As the output approaches ground, the linearity begins to degrade (100 LSB beginning point). As the sink current increases, the nonlinear range increases. The DAC output latch function, contained in the COMMAND register, provides continuous operation while writing each byte of this register. The contents of this register

are volatile, which means that the desired output level must be set after every reset and power cycle event.

Table 21. AUX DAC Register Definition

Address	Default	Format	Access
0x31, 0x30	0x0000	Binary	R/W

Table 22. AUX\_DAC Bit Descriptions

Bit	Description
15:12	Not used
11:0	Data bits

#### General-Purpose I/O

The ADIS16260/ADIS16265 provide two general-purpose pins that enable digital I/O control using the SPI. The GPIO\_CTRL control register establishes the configuration of these pins and handles the SPI-to-pin controls. Each pin provides the flexibility of both input (read) and output (write) operations. For example, writing a 0x0202 to this register establishes Line 0 as an output and sets its level as a one. Writing 0x0000 to this register establishes both lines as inputs, and their status can be read through Bit 0 and Bit 1 of this register.

The digital I/O lines are also available for data-ready and alarm/error indications. In the event of conflict, the following priority structure governs the digital I/O configuration:

- MSC\_CTRL
- 2. ALM\_CTRL
- 3. GPIO\_CTRL

Table 23. GPIO CTRL Register Definition

Address	Default	Format	Access
0x33, 0x32	0x0000	N/A	R/W

Table 24. GPIO CTRL Bit Descriptions

Table 24. GITO_CIRE bit Descriptions		
Bit	Description	
15:10	Not used	
9	General-purpose I/O Line 1 polarity  1 = high  0 = low	
8	General-purpose I/O Line 0 polarity  1 = high  0 = low	
7:2	Not used	
1	General-purpose I/O Line 1, data direction control  1 = output  0 = input	
0	General-purpose I/O Line 0, data direction control  1 = output  0 = input	

#### STATUS AND DIAGNOSTICS

The ADIS16260/ADIS16265 provide a number of status and diagnostic functions. Table 25 provides a summary of these functions, along with their appropriate control registers.

**Table 25. Status and Diagnostic Functions** 

Function	Register
Data-ready I/O indicator	MSC_CTRL
Self-test, mechanical check for MEMS sensor	MSC_CTRL
Status, check for predefined error conditions	STATUS
Flash memory endurance	ENDURANCE
Alarms, configure and check for user-specific	ALM_MAG1/2
conditions	ALM_SMPL1/2
	ALM CTRL

#### Data-Ready I/O Indicator

The data-ready function provides an indication of updated output data. The MSC\_CTRL register provides the opportunity to configure either of the general-purpose I/O pins (DIO0 and DIO1) as a data-ready indicator signal. After each output register update, the digital I/O changes states, then returns to its original state, creating a pulsed waveform. The duty cycle of that waveform is in between 15% and 35%.

Table 26. MSC\_CTRL Register Definition

Address	Default	Format	Access
0x35, 0x34	0x0000	N/A	R/W

Table 27. MSC\_CTRL Bit Descriptions

1 more 27 ( 1/10 0 _ 0 1/10 2 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0		
Bit	Description	
15:11	Not used	
10	Internal self-test enable 1 = enabled	
	0 = disabled	
9	External negative rotation self-test enable  1 = enabled	
	0 = disabled	
8	External positive rotation self-test enable	
	1 = enabled 0 = disabled	
7:3	Not used	
2	Data-ready enable	
	1 = enabled	
	0 = disabled	
1	Data-ready polarity	
	1 = active high	
	0 = active low	
0	Data-ready line select	
	1 = DIO1	
	0 = DIO0	

#### Self-Test

The MSC\_CTRL register also provides a self-test function, which verifies the MEMS sensor's mechanical integrity. There are two different self-test options: (1) internal self-test and (2) external self-test. The internal test provides a simple, two-step process

for checking the MEMS sensor: (1) start the process by writing a 1 to Bit 10 in the MSC\_CTRL register and (2) check the result by reading Bit 5 of the STATUS register, after 35 ms.

The external self-test is a static condition that can be enabled and disabled. In this test, both positive and negative MEMS sensor movements are available. After writing to the appropriate control bit, the GYRO\_OUT register reflects the changes after a delay that reflects the sensor signal chain response time.

For example, the standard 52 Hz bandwidth reflects an exponential response with a time constant of 3.2 ms. If the bandwidth is reduced externally (capacitor across RATE and FILT) or internally (increasing the number of filter taps, SENS/AVG), this time constant increases. For the internal self-test option, increasing the delay can produce false alarms, since the internal timing for this function is optimized for maximum bandwidth. The appropriate bit definitions for self-test are listed in Table 26 and Table 27.

#### **Status Conditions**

The STATUS register contains the following error-condition flags: Alarm conditions, self-test status, angular rate over range, SPI communication failure, control register update failure, and power supply out of range. See Table 28 and Table 29 for the appropriate register access and bit assignment for each flag.

The bits assigned for checking power supply range and angular rate over range automatically reset to 0 when the error condition no longer exists. The remaining error-flag bits in the STATUS register require a read in order to return them to 0. Note that a STATUS register read clears all of the bits to 0.

**Table 28. STATUS Register Definition** 

Address	Default	Format	Access
0x3D, 0x3C	0x0000	N/A	Read-only

**Table 29. STATUS Bit Descriptions** 

Table 25. 51111 05 Dit Descriptions		
Bit	Description	
15:10	Not used	
9	Alarm 2 status	
	1 = active, 0 = inactive	
8	Alarm 1 status	
	1 = active, 0 = inactive	
7:6	Not used	
5	Self-test diagnostic error flag	
	1 = error condition, 0 = normal operation	
4	Angular rate over range	
	1 = error condition, 0 = normal operation	
3	SPI communications failure	
	1 = error condition, 0 = normal operation	
2	Control register update failed	
	1 = error condition, 0 = normal operation	
1	Power supply above 5.25 V	
	1 = above 5.25 V, 0 = below 5.25 V (normal)	
0	Power supply below 4.75 V	
	1 = below 4.75 V, 0 = above 4.75 V (normal)	

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#### Flash Memory Endurance

The ENDURANCE register maintains a running count of writes to the flash memory. It provides up to 32,768 counts. Note that if this count is exceeded, the register wraps around, and goes back to zero, before beginning to increment again.

Table 30. ENDURANCE Register Definition

Address	Default	Format	Access
0x01, 0x00	N/A	Binary	Read-only

#### **Alarms**

The ADIS16260/ADIS16265 provide two independent alarm options for event detection. Event detections occur when output register data meets the configured conditions. Configuration options are:

- All output data registers are available for monitoring as the source data.
- The source data can be filtered or unfiltered.
- Comparisons can be static or dynamic (rate of change).
- The threshold levels and times are configurable.
- Comparison can be greater than or less than.

The ALM\_MAG1 register and the ALM\_MAG2 register both establish the threshold level for detecting events. They take on the format of the source data and provide a bit for establishing the greater than/less than comparison direction. When making dynamic comparisons, the ALM\_SMPL1 register and the ALM\_SMPL2 register establish the number of averages taken for the source data as a reference for comparison. In this configuration, each subsequent source data sample is subtracted from the previous one, establishing an instantaneous delta. The ALM\_CTRL register controls the source data selection, static/dynamic selection, filtering selection, and digital I/O usage for the alarms.

The rate of change calculation is

$$Y_C = \frac{1}{N_{DS}} \sum_{n=1}^{N_{DS}} y(n+1) - y(n)$$

*Rate of change alarm*  $\Rightarrow$  is  $Y_C > \text{or} < M_C$ ?

#### where:

 $N_{DS}$  is the number of samples in ALM\_SMPL1/2. y(n) is the sampled output data.  $M_C$  is the magnitude for comparison in ALM\_MAG1/2.  $Y_C$  is the factor to be compared with  $M_C$ . > or < is determined by the MSB in ALM\_MAG1/2.

#### Table 31. ALM\_MAG1 Register Definition

Address	Default	Format	Access
0x21, 0x20	0x0000	N/A	R/W

#### Table 32. ALM\_MAG1 Bit Descriptions

Bit	Description
15	Comparison polarity: 1 = greater than, 0 = less than
14	Not used
13:0	Data bits: format matches source data format

#### Table 33. ALM\_SMPL1 Register Definition

Address	Default	Format	Access
0x25, 0x24	0x0000	Binary	R/W

#### Table 34. ALM\_SMPL1 Bit Descriptions

Bit	Description
15:8	Not used
7:0	Data bits

#### Table 35. ALM\_MAG2 Register Definition

Address	Default	Format	Access
0x23, 0x22	0x0000	N/A	R/W

#### Table 36. ALM\_MAG2 Bit Descriptions

Bit	Description	
15	Comparison polarity	
	1 = greater than	
	0 = less than	
14	Not used	
13:0	Data bits: format matches source data format	

#### Table 37. ALM SMPL2 Register Definition

Address	Default	Format	Access
0x27, 0x26	0x0000	Binary	R/W

#### Table 38. ALM\_SMPL2 Bit Designations

Bit	Description		
15:8	Not used		
7:0	Data bits		

#### Table 39. ALM\_CTRL Register Definition

Tuble 35, TEME CITE REGISTED Delimition			
Address	Default	Format	Access
0x29, 0x28	0x0000	N/A	R/W

#### Table 40. ALM\_CTRL Bit Descriptions

2
1
1

## **OUTLINE DIMENSIONS**

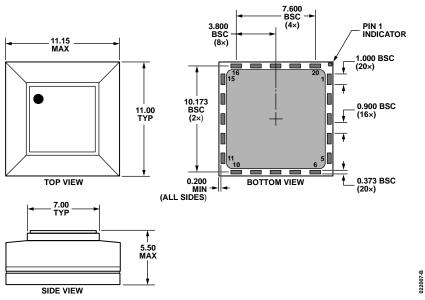


Figure 13. 20-Terminal Stacked Land Grid Array [LGA] (CC-20-1) Dimensions shown in millimeters

#### **ORDERING GUIDE**

Model Temperature Range Page		Package Description	Package Option
ADIS16260BCCZ <sup>1</sup>	-40°C to +105°C	20-Terminal Stacked Land Grid Array [LGA]	CC-20-1
ADIS16265BCCZ <sup>1</sup>	-40°C to +105°C	20-Terminal Stacked Land Grid Array [LGA]	CC-20-1
ADIS16260/PCBZ <sup>1</sup>		Evaluation Board for the ADIS16260	
ADIS16265/PCBZ <sup>1</sup>		Evaluation Board for the ADIS16265	

<sup>&</sup>lt;sup>1</sup> Z = RoHS Compliant Part.