General Description

The MAX9945 operational amplifier features an excellent combination of low operating power and low input voltage noise. In addition, MOS inputs enable the MAX9945 to feature low input bias currents and low input current noise. The device accepts a wide supply voltage range from 4.75V to 38V and draws a low 400µA quiescent current. The MAX9945 is unity-gain stable and is capable of rail-to-rail output voltage swing.

The MAX9945 is ideal for portable medical and industrial applications that require low noise analog front-ends for performance applications such as photodiode transimpedance and chemical sensor interface circuits.

The MAX9945 is available in both an 8-pin μ MAX[®] and a space-saving, 6-pin TDFN package, and is specified over the automotive operating temperature range (-40°C to +125°C).

Applications

Medical Pulse Oximetry

Photodiode Sensor Interface

Industrial Sensors and Instrumentation

Chemical Sensor Interface

High-Performance Audio Line Out

Active Filters and Signal Processing

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_Features

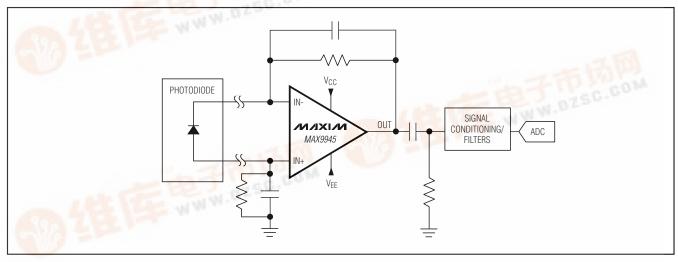
- +4.75V to +38V Single-Supply Voltage Range
- ±2.4V to ±19V Dual-Supply Voltage Range
- Rail-to-Rail Output Voltage Swing
- 400µA Low Quiescent Current
- 50fA Low Input Bias Current
- ◆ 1fA/√Hz Low Input Current Noise
- ♦ 15nV/√Hz Low Noise
- ♦ 3MHz Unity-Gain Bandwidth
- ♦ Wide Temperature Range from -40°C to +125°C
- Available in Space-Saving, 6-Pin TDFN Package (3mm x 3mm)

Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX9945ATT+	-40°C to +125°C	6 TDFN-EP*	AUE
MAX9945AUA+	-40°C to +125°C	8 µMAX	

+Denotes a lead(Pb)-free/RoHS-compliant package. *EP = Exposed pad.

Typical Operating Circuit



_ Maxim Integrated Products 1

Propericing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V _{CC} to V _{EE})0 IN+, IN-, OUT Voltage(V _{EE} - 0.3V) to (V	(CC + 0.3V)
IN+ to IN-	
OUT Short Circuit to Ground Duration	10S
Continuous Input Current into Any Pin	±20mA
Continuous Power Dissipation ($T_A = +70^{\circ}C$)	
6-Pin TDFN-EP (derate 23.8mW/°C above +70°C)	
Multilayer Board	.1904.8mW
Package Thermal Resistance (Note 1)	
θ」Α	42°C/W
θ」C	

8-Pin µMAX (derate 4.8mW/°C above +70	O°C)
Multilayer Board	
Package Thermal Resistance (Note 1)	
θ」Α	206.3°C/W
θJC	42°C/W
Operating Temperature Range	40°C to +125°C
Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to <u>www.maxim-ic.com/thermal-tutorial</u>.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +15V, V_{EE} = -15V, V_{IN+} = V_{IN-} = GND = 0, R_{OUT} = 100k\Omega$ to GND, $T_A = -40^{\circ}C$ to $+125^{\circ}C$, typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	COND	MIN	ТҮР	MAX	UNITS	
DC ELECTRICAL CHARACTERIS	STICS		-				
Input Voltage Range	V _{IN+} , V _{IN-}	Guaranteed by	T _A = +25°C	MIN TYP MAX UNITS V_{EE} V_{CC}^{-} V_{CC}^{-} V_{CC}^{-} V_{EE} V_{CC}^{-} V_{CC}^{-} V_{CC}^{-} ± 0.6 ± 5 mV ± 0.6 ± 5 mV ± 50.6 ± 5 mV 50 fA $MV^{\circ}C$ 78 94 MB			
input voltage nange	v IIN+, v IIN-	CMRR	$T_A = T_{MIN}$ to T_{MAX}	V_{EE}			V
Input Offset Voltage	Vee	$T_A = +25^{\circ}C$			±0.6	±5	
Input Onset Voltage	Vos	$T_A = T_{MIN}$ to T_{MAX}				±8	IIIV
Input Offset Voltage Drift	V _{OS} - T _C				2		µV/°C
Input Bias Current (Note 3)	ΙB				50		fA
Common Made Dejection Datio	CMRR	$V_{CM} = V_{EE}$ to V_{CC} - 1. T _A = +25°C	2V,	78	94		٩D
Common-Mode Rejection Ratio		$V_{CM} = V_{EE}$ to V_{CC} - 1. T _A = T _{MIN} to T _{MAX}	78	94		uБ	
	A	$V_{EE} + 0.3V \le V_{OUT} \le V_{ROUT} = 100k\Omega$ to GNE		110	130		dD
Open-Loop Gain	A _{OL}	$V_{EE} + 0.75V \le V_{OUT} \le R_{OUT} = 10k\Omega$ to GND	V _{CC} - 0.75V,	110	130		dB
Output Short-Circuit Current	ISC				25		mA

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +15V, V_{EE} = -15V, V_{IN+} = V_{IN-} = GND = 0, R_{OUT} = 100k\Omega$ to GND, $T_A = -40^{\circ}C$ to $+125^{\circ}C$, typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.) (Note 2)

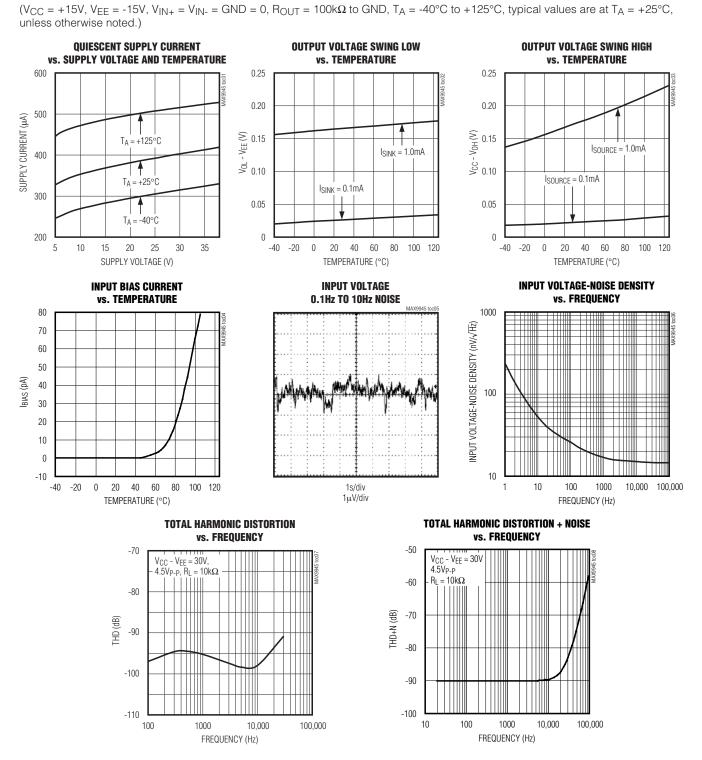
PARAMETER	SYMBOL	CONDI	TIONS	MIN	ТҮР	MAX	UNITS	
		$R_{OUT} = 10k\Omega$ to GND	$T_A = T_{MIN}$ to T_{MAX}		V _{EE} + 0.26	V _{EE} + 0.45	v	
Output Voltage Low	V _{OL}	$R_{OUT} = 100k\Omega$ to GND	$T_A = T_{MIN}$ to T_{MAX}		V _{EE} + 0.05	V _{EE} + 0.15	V	
Quite ut Valta en Llista		$R_{OUT} = 10k\Omega$ to GND	$T_A = T_{MIN}$ to T_{MAX}	V _{CC} - V _{CC} - 0.45 0.24				
Output Voltage High	VOH	$R_{OUT} = 100k\Omega$ to GND	$T_A = T_{MIN}$ to T_{MAX}	V _{CC} - 0.15	V _{CC} - 0.03		V	
AC ELECTRICAL CHARACTERIS	TICS							
Input Current-Noise Density	IN	f = 1kHz			1		fA/√Hz	
Input Voltage Noise	V _{NP-P}	f = 0.1Hz to $10Hz$			2		μV _{P-P}	
	V _N	f = 100Hz	25			nV/√Hz		
Input Voltage-Noise Density		f = 1kHz	16.5					
		f = 10kHz	15					
Gain Bandwidth	GBW			3			MHz	
Slew Rate	SR				2.2		V/µs	
Capacitive Loading (Note 4)	CLOAD	No sustained oscillatio	ns	120			pF	
Total Harmonic Distortion	THD	$V_{O,III} = 4.5V_{D,D} \Delta v = 1V/V$		dB				
POWER-SUPPLY ELECTRICAL	HARACTER	ISTICS						
Power-Supply Voltage Range	V _{CC} - V _{EE}	Guaranteed by PSRR,	$V_{EE} = 0$	+4.75		+38	V	
Power-Supply Rejection Ratio	PSRR	$V_{CC} - V_{EE} = +4.75V$ to	+38V	82	100		dB	
Ourisses and Oursely Ourrest		T _A = +25°C		400	700			
Quiescent Supply Current	Icc	$T_A = T_{MIN}$ to T_{MAX}		850		μA		

Note 2: All devices are 100% production tested at $T_A = +25$ °C. All temperature limits are guaranteed by design.

Note 3: IN+ and IN- are internally connected to the gates of CMOS transistors. CMOS GATE leakage is so small that it is impractical to test in production. Devices are screened during production testing to eliminate defective units.

Note 4: Specified over all temperatures and process variation by circuit simulation.

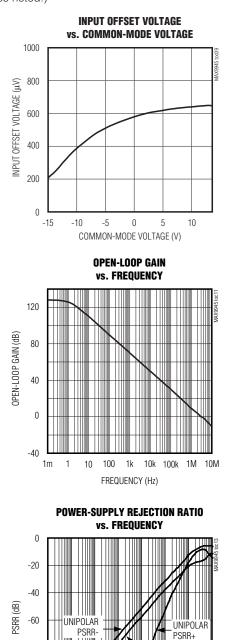
MAX9945



Typical Operating Characteristics

Typical Operating Characteristics (continued)

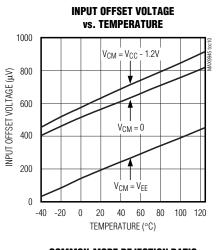
 $(V_{CC} = +15V, V_{EE} = -15V, V_{IN+} = V_{IN-} = GND = 0, R_{OUT} = 100k\Omega$ to GND, $T_A = -40^{\circ}C$ to $+125^{\circ}C$, typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.)



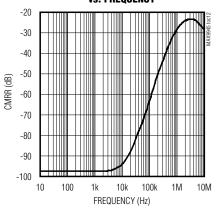
BIPOLAR

SRR

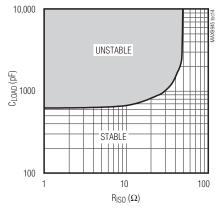
FREQUENCY (Hz)



COMMON-MODE REJECTION RATIO vs. FREQUENCY



RESISTOR ISOLATION vs. Capacitive Load



-80

-100

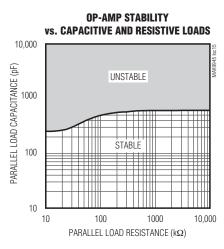
-120

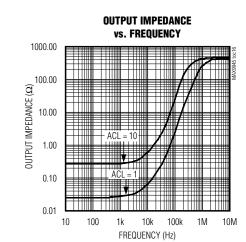
1

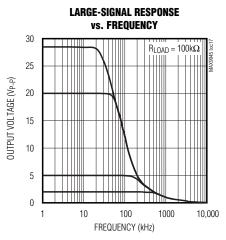
10 100 1k 10k 100k 1M 10M

Typical Operating Characteristics (continued)

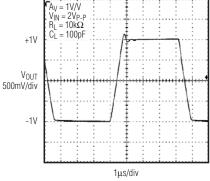
 $(V_{CC} = +15V, V_{EE} = -15V, V_{IN+} = V_{IN-} = GND = 0, R_{OUT} = 100k\Omega$ to GND, $T_A = -40^{\circ}C$ to $+125^{\circ}C$, typical values are at $T_A = +25^{\circ}C$, unless otherwise noted.)



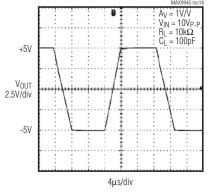




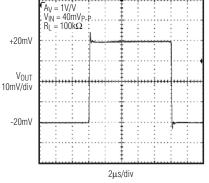
LARGE SIGNAL-STEP RESPONSE













_Pin Description

PIN	I	NAME	FUNCTION
TDFN-EP	μΜΑΧ	NAME	FUNCTION
1	6	OUT	Amplifier Output
2	4	V_{EE}	Negative Power Supply. Bypass V _{EE} with 0.1 μ F ceramic and 4.7 μ F electrolytic capacitors to quiet ground plane if different from V _{EE} .
3	3	IN+	Noninverting Amplifier Input
4	2	IN-	Inverting Amplifier Input
5	1, 5, 8	N.C.	No Connection. Not internally connected.
6	7	V _{CC}	Positive Power Supply. Bypass V_{CC} with 0.1µF ceramic and 4.7µF electrolytic capacitors to quiet ground plane or $V_{EE}.$
	_	EP	Exposed Pad. Connect to V_{EE} externally. Connect to a large copper plane to maximize thermal performance. Not intended as an electrical connection (TDFN only).

Detailed Description

The MAX9945 features a combination of low input current and voltage noise, rail-to-rail output voltage swing, wide supply voltage range, and low-power operation. The MOS inputs on the MAX9945 make it ideal for use as transimpedance amplifiers and high-impedance sensor interface front-ends in medical and industrial applications. The MAX9945 can interface with small signals from either current-sources or high-output impedance voltage sources. Applications include photodiode pulse oximeters, pH sensors, capacitive pressure sensors, chemical analysis equipment, smoke detectors, and humidity sensors.

A high 130dB open-loop gain (typ) and a wide supply voltage range, allow high signal-gain implementations prior to signal conditioning circuitry. Low quiescent supply current makes the MAX9945 compatible with portable systems and applications that operate under tight power budgets. The combination of excellent THD, low voltage noise, and MOS inputs also make the MAX9945 ideal for use in high-performance active filters for data acquisition systems and audio equipment.

Low-Current, Low-Noise Input Stage

The MAX9945 features a MOS-input stage with only 50fA (typ) of input bias current and a low 1fA/ \sqrt{Hz} (typ) input current-noise density. The low-frequency input voltage noise is a low 2µV_{P-P} (typ). The input stage accepts a wide common-mode range, extending from the negative supply, V_{EE}, to within 1.2V of the positive supply, V_{CC}.

Rail-to-Rail Output Stage

The MAX9945 output stage swings to within 50mV (typ) of either power-supply rail with a 100k Ω load and provides a 3MHz GBW with a 2.2V/µs slew rate. The device is unity-gain stable, and unlike other devices with a low quiescent current, can drive a 120pF capacitive load without compromising stability.

Applications Information

High-Impedance Sensor Front Ends

High-impedance sensors can output signals of interest in either current or voltage form. The MAX9945 interfaces to both current-output sensors such as photodiodes and potentiostat sensors, and high-impedance voltage sources such as pH sensors.

For current-output sensors, a transimpedance amplifier is the most noise-efficient method for converting the input signal to a voltage. High-value feedback resistors are commonly chosen to create large gains, while feedback capacitors help stabilize the amplifier by canceling any zeros in the transfer function created by a highly capacitive sensor or cabling. A combination of low-current noise and low-voltage noise is important for these applications. Take care to calibrate out photodiode dark current if DC accuracy is important. The high bandwidth and slew rate also allows AC signal processing in certain medical photodiode sensor applications such as pulse oximetry.

MAX994



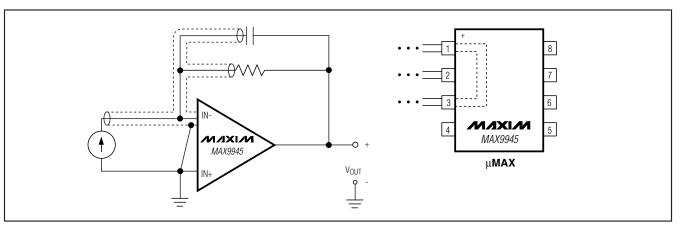


Figure 1. Shielding the Inverting Input to Reduce Leakage

For voltage-output sensors, a noninverting amplifier is typically used to buffer and/or apply a small gain to, the input voltage signal. Due to the extremely high impedance of the sensor output, a low input bias current with a small temperature variation is very important for these applications.

Power-Supply Decoupling

The MAX9945 operates from a +4.75V to +38V, V_{EE} referenced power supply. Bypass the power-supply inputs V_{CC} and V_{EE} to a quiet copper ground plane, with a 0.1μ F ceramic capacitor in parallel with a 4.7μ F electrolytic capacitor, placed close to the leads.

Layout Techniques

A good layout is critical to obtaining high performance especially when interfacing with high-impedance sensors. Use shielding techniques to guard against parasitic leakage paths. For transimpedance applications, for example, surround the inverting input, and the traces connecting to it, with a buffered version of its own voltage. A convenient source of this voltage is the noninverting input pin. Pins 1, 5, and 8 on the μ MAX package are unconnected, and can be connected to an analog common potential, or to the driven guard potential, to reduce leakage on the inverting input.

A good layout guard rail isolates sensitive nodes, such as the inverting input of the MAX9945 and the traces connecting to it (see Figure 1), from varying or large voltage differentials that otherwise occur in the rest of the circuit board. This reduces leakage and noise effects, allowing sensitive measurements to be made accurately.

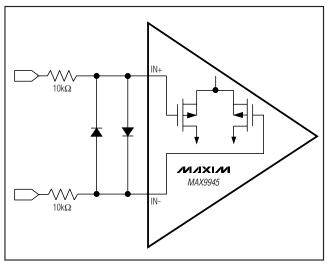


Figure 2. Input Differential Voltage Protection

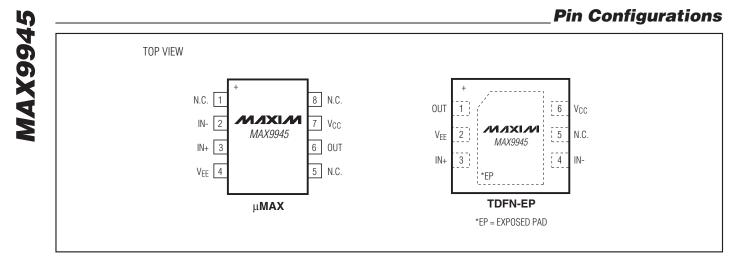
Take care to also decrease the amount of stray capacitance at the op amp's inputs to improve stability. To achieve this, minimize trace lengths and resistor leads by placing external components as close as possible to the package. If the sensor is inherently capacitive, or is connected to the amplifier through a long cable, use a low-value feedback capacitor to control high-frequency gain and peaking to stabilize the feedback loop.

Input Differential Voltage Protection

During normal op-amp operation, the inverting and noninverting inputs of the MAX9945 are at approximately the same voltage. The ±12V absolute maximum input differential voltage rating offers sufficient protection for most applications. If there is a possibility of exceeding the input differential voltage specification, in the presence of extremely fast input voltage transients or due to certain application-specific fault conditions, use external low-leakage pico-amp diodes and series resistors to protect the input stage of the amplifier (see Figure 2). The extremely low input bias current of the MAX9945 allows a wide range of input series resistors to be used. If low input voltage noise is critical to the application, size the input series resistors appropriately.

_Chip Information

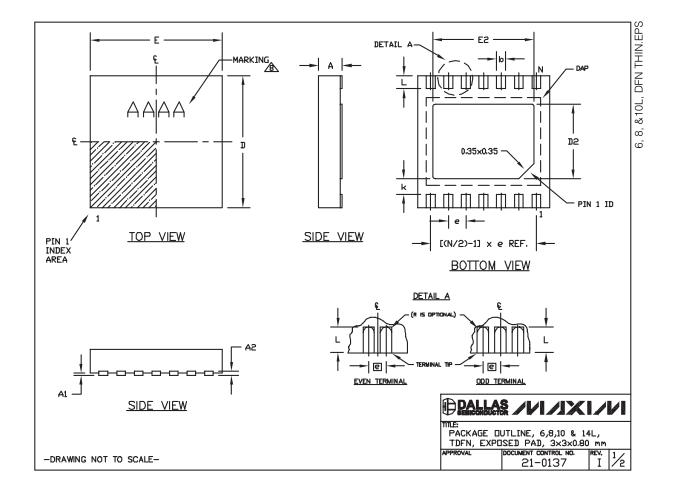
PROCESS: BICMOS



_Package Information

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.		
6 TDFN-EP	T633-2	<u>21-0137</u>		
8 µMAX	U8-1	<u>21-0036</u>		



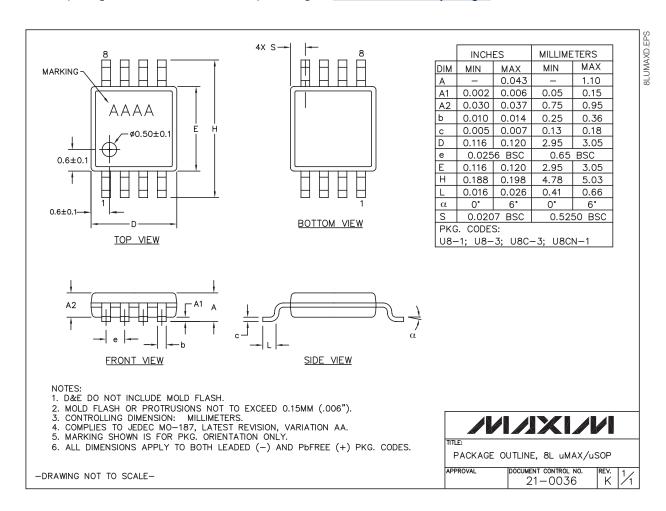
Package Information (continued)

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages.

COMMON	DIMENS	SIONS	PACKAGE VA	RIATI	ONS					
SYMBOL	MIN.	MAX.	PKG. CODE	N	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e
А	0.70	0.80	T633-2	6	1.50±0.10	2.30±0.10	0.95 BSC	MO229 / WEEA	0.40±0.05	1.90 REF
D	2.90	3.10	T833-2	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
E	2.90	3.10	T833-3	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF
A1	0.00	0.05	T1033-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229/WEED-3	0.25±0.05	2.00 REF
L	0.20	0.40	T1033-2	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229/WEED-3	0.25±0.05	2.00 REF
k	0.25	MIN.	T1433-1	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
A2	0.20	REF.	T1433-2	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF
2. COPL 3. WARP 4. PACK 5. DRAW 6. "N" 1	ANARITY AGE SH AGE LEI ING CO S THE	ONS ARE IN mm ' SHALL NOT EX IALL NOT EXCEE NGTH/PACKAGE NFORMS TO JEL TOTAL NUMBER LEADS SHOWN	CEED 0.08 m D 0.10 mm. WIDTH ARE CO EC MO229, E OF LEADS.	m. DNSID XCEP	ERED AS S T DIMENSIO			C(S). ND T1433-1 & T	1433–2.	
	ING IS	FOR PACKAGE	DRIENTATION F	EFER	ENCE ONLY					

Package Information (continued)

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MAX994: