

#### LMV981 Single / LMV982 Dual 1.8V, RRIO Operational Amplifiers with Shutdown **General Description** Features

LMV981/LMV982 are low voltage, low power operational amplifiers. LMV981/LMV982 are guaranteed to operate from +1.8V to +5.0V supply voltages and have rail-to-rail input and output. LMV981/LMV982 input common mode voltage extends 200mV beyond the supplies which enables user enhanced functionality beyond the supply voltage range. The output can swing rail-to-rail unloaded and within 105mV from the rail with  $600\Omega$  load at 1.8V supply. LMV981/ LMV982 are optimized to work at 1.8V which make them ideal for portable two-cell battery powered systems and single cell Li-Ion systems.

LMV981/LMV982 offer a shutdown pin that can be used to disable the device and reduce the supply current. The device is in shutdown when the  $\overline{SHDN}$ -pin = low.

LMV981/LMV982 exhibit excellent speed-power ratio, achieving 1.4MHz gain bandwidth product at 1.8V supply voltage with very low supply current. LMV981/LMV982 are capable of driving a  $600\Omega$  load and up to 1000pF capacitive load with minimal ringing. LMV981/LMV982 have a high DC gain of 101dB, making them suitable for low frequency applications.

LMV981 is offered in space saving 6-Bump micro SMD, SC70-6 and SOT23-6 packages. The 6-Bump micro SMD package has only a 1.006mm x 1.514mm x 0.945mm footprint. LMV982 is offered in space saving MSOP-10 package. These small packages are ideal solutions for area constrained PC boards and portable electronics such as cellular phones and PDAs.

(Typical 1.8V Supply Values; Unless Otherwise Noted)

Guaranteed 1.8V, 2.7V and 5V specifications

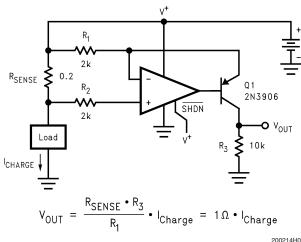
<ul> <li>Output swing</li> </ul>	
— w/600Ω load	80mV from rail
— w/2kΩ load	30mV from rail
■ V <sub>CM</sub>	200mV beyond rails
<ul> <li>Supply current (per channel)</li> </ul>	100µA
Gain bandwidth product	1.4MHz
Maximum V <sub>OS</sub>	4.0mV
■ Gain w/600Ω load	101dB
<ul> <li>Ultra tiny package micro SMD</li> </ul>	1.0mm x 1.5mm
Turn-on time from shutdown	19µs
Temperature range	–40°C to 125°C

#### Applications

Industrial and automotive

- Consumer communication
- Consumer computing
- PDAs
- Portable audio
- Portable/battery-powered electronic equipment
- Supply current monitoring
- Battery monitoring

# **Typical Application**



#### Absolute Maximum Ratings (Note 1) 词"山孙孙岛小孙e"供应商specified devices are required,

ut Military Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2)	
Machine Model	200V
Human Body Model	2000V
Differential Input Voltage	± Supply Voltage
Supply Voltage (V <sup>+</sup> –V <sup>-</sup> )	5.5V
Output Short Circuit to V <sup>+</sup> (Note 3)	
Output Short Circuit to V <sup>-</sup> (Note 3)	
Storage Temperature Range	–65°C to 150°C
Junction Temperature (Note 4)	150°C

Mounting Temp.

235°C

#### **Operating Ratings** (Note 1)

Infrared or Convection (20 sec)

Supply Voltage Range	1.8V to 5.0V
Temperature Range	–40°C to 125°C
Thermal Resistance $(\theta_{JA})$	
6-Bump micro SMD	286°C/W
SC70-6	414°C/W
SOT23-6	265°C/W
MSOP-10	235°C/W

#### **1.8V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ . V<sup>+</sup> = 1.8V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sup>+</sup>/2, V<sub>O</sub> = V<sup>+</sup>/2, R<sub>L</sub> > 1 M $\Omega$  and SHDN tied to V<sup>+</sup>. **Boldface** limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Co	ndition	Min	Тур	Max	Units
				(Note 6)	(Note 5)	(Note 6)	
V <sub>os</sub>	Input Offset Voltage	LMV981 (Single	e)		1	4	
						6	mV
		LMV982 (Dual)			1	5.5	
						7.5	
TCV <sub>OS</sub>	Input Offset Voltage Average Drift				5.5		µV/°C
I <sub>B</sub>	Input Bias Current				15	35	nA
						50	
l <sub>os</sub>	Input Offset Current				13	25	nA
						40	
I <sub>S</sub> Supply Current (per cha	Supply Current (per channel)				103	185	
						205	
		In Shutdown	LMV981 (Single)		0.156	1	μA
						2	μΑ
			LMV982 (Dual)		0.178	3.5	
						5	
CMRR	Common Mode Rejection	LMV981, $0 \le V$	• · · ·	60	78		
	Ratio	$1.4V \le V_{CM} \le 1$	.8V	55			
		(Note 8)					
		LMV982, $0 \le V_{i}$	• · · ·	55	76		dB
		$1.4V \le V_{CM} \le 1$		50			_
		$-0.2V \le V_{CM} \le$		50	72		
		$1.8V \le V_{CM} \le 2$					
PSRR	Power Supply Rejection	$1.8V \le V^+ \le 5V$		75	100		dB
	Ratio			70			
CMVR	Input Common-Mode Voltage	For CMRR	T <sub>A</sub> = 25°C	V <sup>-</sup> -0.2	-0.2 to 2.1	V <sup>+</sup> +0.2	
	Range	Range $\geq$ 50dB	$T_A = -40^{\circ}C$ to	V-		V <sup>+</sup>	v
			85°C				ľ
			T <sub>A</sub> = 125°C	V <sup>-</sup> +0.2		V <sup>+</sup> –0.2	

Symbol	Parameter	Condition	Min (Noto 6)	Typ (Note 5)	Max (Note 6)	Units
٨			(Note 6)	, ,	(NOLE 0)	
Av	Large Signal Voltage Gain	$R_{L} = 600\Omega$ to 0.9V,	77	101		
	LMV981 (Single)	$V_{\rm O} = 0.2V$ to 1.6V, $V_{\rm CM} = 0.5V$	73	105		dB
		$R_L = 2k\Omega$ to 0.9V,	80	105		
		$V_{\rm O} = 0.2V$ to 1.6V, $V_{\rm CM} = 0.5V$	75			
	Large Signal Voltage Gain	$R_L = 600\Omega$ to 0.9V,	75	90		
	LMV982 (Dual)	$V_{\rm O} = 0.2V$ to 1.6V, $V_{\rm CM} = 0.5V$	72			dB
		$R_{L} = 2k\Omega$ to 0.9V,	78	100		uD
		$V_{O} = 0.2V$ to 1.6V, $V_{CM} = 0.5V$	75			
Vo	Output Swing	$R_L = 600\Omega$ to 0.9V	1.65	1.72		
		$V_{IN} = \pm 100 \text{mV}$	1.63			
				0.077	0.105	1
					0.120	
		$R_{\rm I} = 2k\Omega$ to 0.9V	1.75	1.77		V
		$V_{\rm IN} = \pm 100 {\rm mV}$	1.74			
				0.024	0.035	1
					0.04	
lo	Output Short Circuit Current	Sourcing, V <sub>O</sub> = 0V	4	8		
-		$V_{IN} = 100 \text{mV}$	3.3			
		Sinking, V <sub>O</sub> = 1.8V	7	9		mA
		$V_{IN} = -100 \text{mV}$	5			
Ton	Turn-on Time from Shutdown			19		μs
V <sub>SHDN</sub>	Turn-on Voltage to enable			1.0		
	part					V
	part					V 1

## **1.8V AC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ . V<sup>+</sup> = 1.8V, V<sup>-</sup> = 0V,  $V_{CM} = V^+/2$ ,  $V_O = V^+/2$ ,  $R_L > 1 M\Omega$  and  $\overline{SHDN}$  tied to V<sup>+</sup>. **Boldface** limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Conditions	Min	Тур	Max	Units
			(Note 6)	(Note 5)	(Note 6)	
SR	Slew Rate	(Note 7)		0.35		V/µs
GBW	Gain-Bandwidth Product			1.4		MHz
$\Phi_{m}$	Phase Margin			67		deg
G <sub>m</sub>	Gain Margin			7		dB
e <sub>n</sub>	Input-Referred Voltage Noise	$f = 1 \text{kHz}, V_{CM} = 0.5 \text{V}$		60		nV √Hz
i <sub>n</sub>	Input-Referred Current Noise	f = 1kHz		0.06		<u>pA</u> √Hz
THD	Total Harmonic Distortion	$f = 1 \text{kHz}, \text{ A}_{V} = +1$ $\text{R}_{L} = 600\Omega, \text{ V}_{\text{IN}} = 1 \text{ V}_{\text{PP}}$		0.023		%
	Amp-to-Amp Isolation	(Note 9)		123		dB

LMV981/LMV982

# **1.8V DC Electrical Characteristics** (Continued)

112 1	M $\Omega$ and SHDN tied to V <sup>+</sup> . Boldf			ktremes. See	(Note 10).		
Symbol	Parameter	Co	ndition	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Un
V <sub>OS</sub>	Input Offset Voltage	LMV981 (Single	e)		1	4 6	m
		LMV982 (Dual)			1	6 <b>7.5</b>	m
TCV <sub>OS</sub>	Input Offset Voltage Average Drift				5.5		μV
Ι <sub>Β</sub>	Input Bias Current				15	35 <b>50</b>	n
l <sub>os</sub>	Input Offset Current				8	25 <b>40</b>	n
I <sub>S</sub>	Supply Current (per channel)				105	190 <b>210</b>	
		In Shutdown	LMV981 (Single)		0.061	1 2	μΑ
			LMV982 (Dual)		0.101	3.5 <b>5</b>	
CMRR	Ratio 2	LMV981, $0 \le V$ 2.3V $\le V_{CM} \le 2$		60 <b>55</b>	81		
		LMV982, $0 \le V$ , 2.3V $\le V_{CM} \le 2$		55 <b>50</b>	80		d
		$-0.2V \le V_{CM} \le 2.7V \le V_{CM} \le 2$		50	74		
PSRR	Power Supply Rejection Ratio	$1.8V \le V^+ \le 5V$ $V_{CM} = 0.5V$		75 <b>70</b>	100		d
CMVR	Input Common-Mode Voltage	For CMRR	$T_A = 25^{\circ}C$	V <sup>-</sup> -0.2	-0.2 to 3.0	V ++0.2	-
	Range	Range ≥ 50dB	85°C	V-		V <sup>+</sup>	\ \
A <sub>V</sub>	Large Signal Voltage Gain	$R_{L} = 600\Omega$ to 1	$T_{A} = 125^{\circ}C$	V <sup>-</sup> +0.2 87	104	V +-0.2	
70	LMV981(Single)	$V_{\rm O} = 0.2V$ to 2.		86	104		.
		$R_L = 2k\Omega$ to 1.3		92	110		- d
	Large Cignel Veltage Cain	$V_{\rm O} = 0.2V$ to 2.		91	00		
	Large Signal Voltage Gain LMV982 (Dual)	$R_{L} = 600\Omega$ to 1 $V_{O} = 0.2V$ to 2.		78 75	90		
		$R_{\rm L} = 2k\Omega$ to 1.3		81	100		
		$V_{\rm O} = 0.2V$ to 2.	5V	78			
V <sub>O</sub>	Output Swing	$R_L = 600\Omega$ to 1 $V_{IN} = \pm 100 \text{mV}$	.35V	2.55 <b>2.53</b>	2.62		
					0.083	0.110 <b>0.130</b>	
		$R_L = 2k\Omega$ to 1.3 $V_{IN} = \pm 100 \text{mV}$	35V	2.65 <b>2.64</b>	2.675		
					0.025	0.04 <b>0.045</b>	
lo	Output Short Circuit Current	Sourcing, $V_O = V_{IN} = 100 \text{mV}$	0V	20 <b>15</b>	30		
		Sinking, $V_O = 0$ $V_{IN} = -100 \text{mV}$	V	18 <b>12</b>	25		- n
Ton	Turn-on Time from Shutdown				12.5		L L

Unies	<b>2.7V DC Electrical Characteristics</b> (Continued) Unespot Machine guaranteed for $T_J = 25^{\circ}$ C. V <sup>+</sup> = 2.7V, V <sup>-</sup> = 0V, V <sub>CM</sub> = V <sup>+</sup> /2, V <sub>O</sub> = V <sup>+</sup> /2, R <sub>L</sub> > 1 M\Omega and SHDN tied to V <sup>+</sup> . Boldface limits apply at the temperature extremes. See (Note 10).										
Symbol	Parameter	Condition	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units					
V <sub>SHDN</sub>	Turn-on Voltage to enable part		(1000 0)	1.9	(	v					
	Turn-off Voltage			0.8							

### 2.7V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ . V<sup>+</sup> = 2.7V, V<sup>-</sup> = 0V, V<sub>CM</sub> = 1.0V, V<sub>O</sub> = 1.35V, R<sub>L</sub> > 1 M $\Omega$  and SHDN tied to V<sup>+</sup>. **Boldface** limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
SR	Slew Rate	(Note 7)		0.4		V/µs
GBW	Gain-Bandwidth Product			1.4		MHz
$\Phi_{m}$	Phase Margin			70		deg
G <sub>m</sub>	Gain Margin			7.5		dB
e <sub>n</sub>	Input-Referred Voltage Noise	$f = 1 kHz, V_{CM} = 0.5V$		57		nV √Hz
i <sub>n</sub>	Input-Referred Current Noise	f = 1kHz		0.082		<u>pA</u> √Hz
THD	Total Harmonic Distortion	$f = 1 \text{kHz}, \text{ A}_{V} = +1$ $\text{R}_{L} = 600 \text{k}\Omega, \text{ V}_{\text{IN}} = 1 \text{ V}_{\text{PP}}$		0.022		%
	Amp-to-Amp Isolation	(Note 9)		123		dB

### **5V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ . V<sup>+</sup> = 5V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sup>+</sup>/2, V<sub>O</sub> = V<sup>+</sup>/2, R<sub>L</sub> > 1 M $\Omega$  and SHDN tied to V<sup>+</sup>. **Boldface** limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Co	ondition	Min	Тур	Max	Units
				(Note 6)	(Note 5)	(Note 6)	
Vos	Input Offset Voltage	LMV981 (Singl	e)		1	4	
						6	mV
		LMV982 (Dual)	)		1	5.5	
						7.5	
TCV <sub>os</sub>	Input Offset Voltage Average				5.5		μV/°C
	Drift						
I <sub>B</sub>	Input Bias Current				14	35	nA
						50	
I <sub>OS</sub> Input Offset Current				9	25	nA	
						40	
I <sub>s</sub>	Supply Current (per Channel)				116	210	
						230	μA
		In Shutdown	LMV981 (Single)		0.201	1	
						2	
			LMV982 (Dual)		0.302	3.5	μA
						5	
CMRR	Common Mode Rejection	$0 \le V_{CM} \le 3.8$	/	60	86		
	Ratio	$4.6V \le V_{CM} \le 8$	5.0V (Note 8)	55			- dB
		$-0.2V \le V_{CM} \le$	0V	50	78		
		$5.0V \le V_{CM} \le 3$	5.2V				
PSRR	Power Supply Rejection	$1.8V \le V^+ \le 5V$	/	75	100		dB
	Ratio	$V_{CM} = 0.5V$		70			

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LMV981/LMV982

LMV981/HMV982

**5V DC Electrical Characteristics** (Continued) UNASSET WHAT SHOW AND AND THE SUBJECT OF T<sub>J</sub> = 25°C. V<sup>+</sup> = 5V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sup>+</sup>/2, V<sub>O</sub> = V<sup>+</sup>/2, R<sub>L</sub> > 1 MΩ and SHDN tied to V<sup>+</sup>. Boldface limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Condition		Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
CMVR	Input Common-Mode Voltage	For CMRR	T <sub>A</sub> = 25°C	V <sup>-</sup> -0.2	-0.2 to 5.3	V <sup>+</sup> +0.2	
	Range	Range ≥ 50dB	$T_A = -40^{\circ}C$ to 85°C	V-		V+	v
			T <sub>A</sub> = 125°C	V <sup>-</sup> +0.3	1	V <sup>+</sup> -0.3	1
A <sub>V</sub>	Large Signal Voltage Gain	$R_L = 600\Omega$ to 2	.5V,	88	102		
	(LMV981 Single)	$V_{\rm O} = 0.2V$ to 4.	8V	87			dB
		$R_L = 2k\Omega$ to 2.5	δV,	94	113		
		$V_{\rm O} = 0.2V$ to 4.	8V	93			
	Large Signal Voltage Gain	$R_L = 600\Omega$ to 2	.5V,	81	90		
	LMV982 (Dual)	$V_{\rm O} = 0.2V$ to 4.	8V	78			- dB
		$R_{L} = 2k\Omega$ to 2.5V,		85	100		UD
		$V_{\rm O} = 0.2V$ to 4.	8V	82			
Vo	Output Swing	$R_L = 600\Omega$ to 2		4.855	4.890		
		$V_{IN} = \pm 100 mV$	(Note 8)	4.835			
					0.120	0.160	
						0.180	- v
		$R_L = 2k\Omega$ to 2.5	δV	4.945	4.967		
		$V_{IN} = \pm 100 mV$		4.935			
					0.037	0.065 <b>0.075</b>	
I <sub>o</sub>	Output Short Circuit Current	LMV981, Sourc	ing, $V_O = 0V$	80	100		
		$V_{IN} = 100 mV$		68			mA
		Sinking, V <sub>O</sub> = 5	V	58	65		
		$V_{IN} = -100 mV$		45			
Ton	Turn-on Time from Shutdown				8.4		μs
V <sub>SHDN</sub>	Turn-on Voltage to enable				4.2		
	part						V
	Turn-off Voltage				0.8		1

### **5V AC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25$  °C. V<sup>+</sup> = 5V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sup>+</sup>/2, V<sub>O</sub> = 2.5V, R<sub>L</sub> > 1 M $\Omega$  and SHDN tied to V<sup>+</sup>.**Boldface** limits apply at the temperature extremes. See (Note 10).

Symbol	Parameter	Conditions	Min (Note 6)	Typ (Note 5)	Max (Note 6)	Units
SR	Slew Rate	(Note 7)		0.42		V/µs
GBW	Gain-Bandwidth Product			1.5		MHz
$\Phi_{m}$	Phase Margin			71		deg
G <sub>m</sub>	Gain Margin			8		dB
e <sub>n</sub>	Input-Referred Voltage Noise	$f = 1$ kHz, $V_{CM} = 1$ V		50		nV √Hz
i <sub>n</sub>	Input-Referred Current Noise	f = 1kHz		0.07		<u>pA</u> √Hz
THD	Total Harmonic Distortion	f = 1kHz, A <sub>V</sub> = +1 R <sub>L</sub> = 600Ω, V <sub>O</sub> = 1V <sub>PP</sub>		0.022		%
	Amp-to-Amp Isolation	(Note 9)		123		dB

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 3: Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Output currents in excess of 45mA over long term may adversely affect reliability.

Note 4: The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

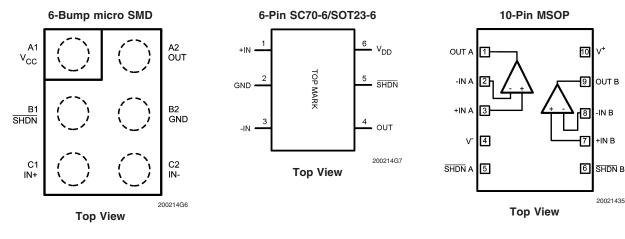
Note 7: V<sup>+</sup> = 5V. Connected as voltage follower with 5V step input. Number specified is the slower of the positive and negative slew rates.

Note 8: For guaranteed temperature ranges, see Input Common-Mode Voltage Range specifications.

Note 9: Input referred, V<sup>+</sup> = 5V and R<sub>L</sub> =  $100k\Omega$  connected to 2.5V. Each amp excited in turn with 1kHz to produce V<sub>O</sub> =  $3V_{PP}$ .

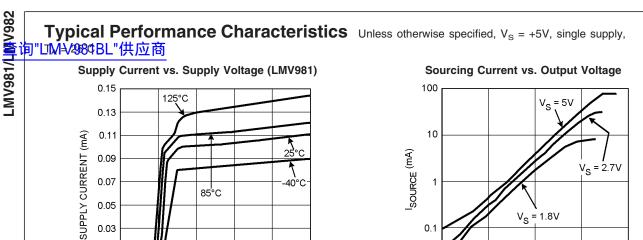
**Note 10:** Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self heating where  $T_J > T_A$ . See Applications section for information on temperature derating of this device. Absolute Maximum Ratings indicated junction temperature limits beyond which the device may be permanently degraded, either mechanically or electrically.

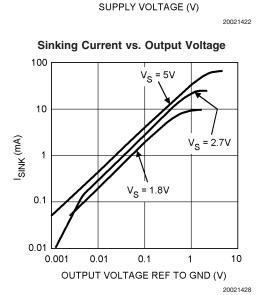
## **Connection Diagrams**



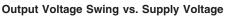
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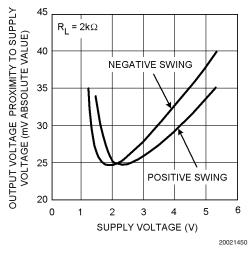
Package	Part Number	Packaging Marking	Transport Media	NSC Drawing
6-Bump micro SMD	LMV981BL	A	250 Units Tape and Reel	BLA006AAB
	LMV981BLX		3k Units Tape and Reel	
6-Pin SC70	LMV981MG	A77	1k Units Tape and Reel	MAA06A
	LMV981MGX		3k Units Tape and Reel	
6-Pin SOT23	LMV981MF	A78A	1k Units Tape and Reel	MF06A
	LMV981MFX		3.5k Units Tape and Reel	- IVIFU6A
10-Pin MSOP	LMV982MM	A87A	1k Unit Tape and Reel	MUB10A
	LMV982MMX	7	3.5k Unit Tape and Reel	IVIOBIUA

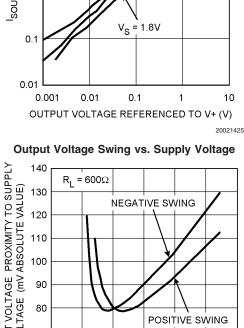


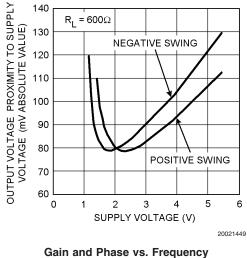


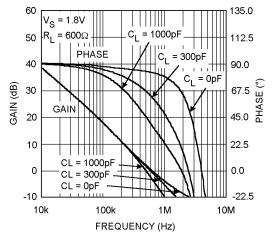
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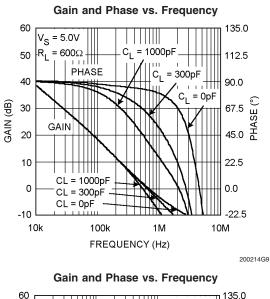


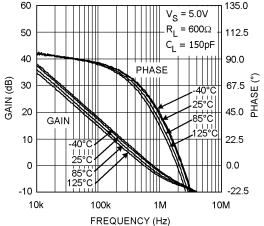




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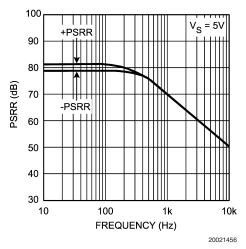
# Typical Performance Characteristics Unless otherwise specified, V<sub>s</sub> = +5V, single supply, T<sub>A</sub>查询几仰姆哈里"供应商

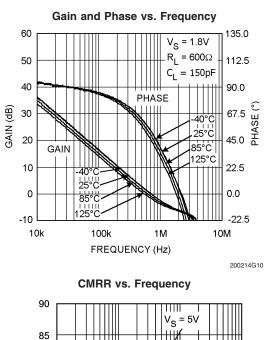


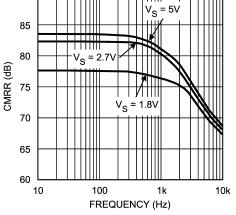




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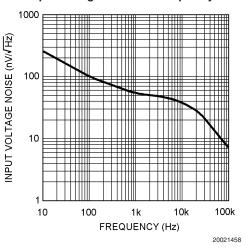


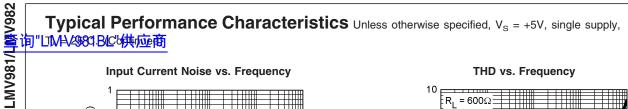


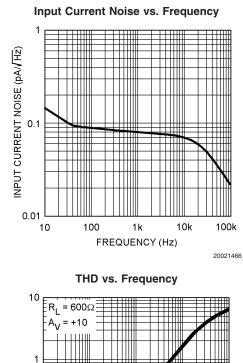


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Input Voltage Noise vs. Frequency









1k

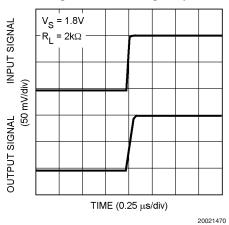
FREQUENCY (Hz)

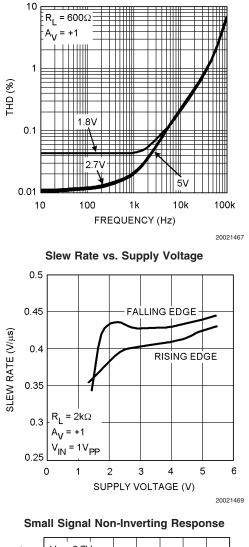
10k

100k

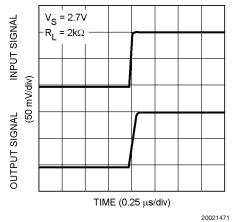
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1.8\





THD vs. Frequency



THD (%)

0.1

0.01

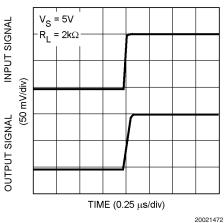
10

2.7V

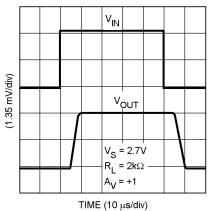
100

# Typical Performance Characteristics Unless otherwise specified, V<sub>s</sub> = +5V, single supply, T<sub>A</sub> 查询CLND/98+BL"供应商

**Small Signal Non-Inverting Response** 

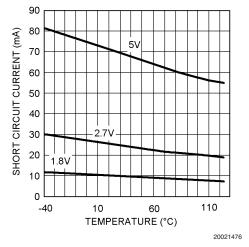


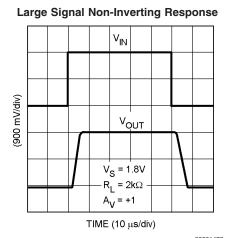
Large Signal Non-Inverting Response



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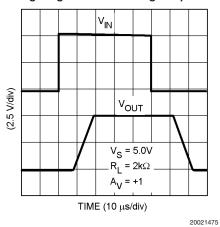
Short Circuit Current vs. Temperature (Sinking)



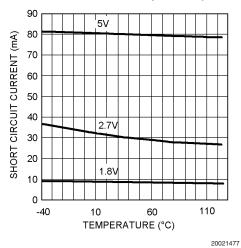


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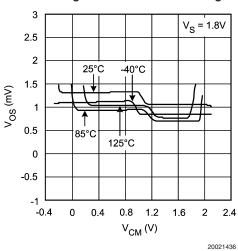
Short Circuit Current vs. Temperature (Sourcing)



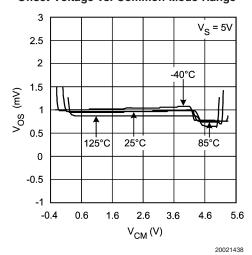
#### Typical Performance Characteristics Unless otherwise specified, V<sub>s</sub> = +5V, single supply, 间"LTMA终约.BC均换应商

Offset Voltage vs. Common Mode Range

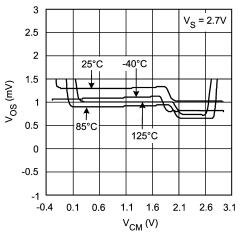
LMV981/HMV982











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#### Application Note 查询"LMV981BL"供应商 1.0 INPUT AND OUTPUT STAGE

The rail-to-rail input stage of this family provides more flexibility for the designer. The LMV981/LMV982 use a complimentary PNP and NPN input stage in which the PNP stage senses common mode voltage near V<sup>-</sup> and the NPN stage senses common mode voltage near V<sup>+</sup>. The transition from the PNP stage to NPN stage occurs 1V below V<sup>+</sup>. Since both input stages have their own offset voltage, the offset of the amplifier becomes a function of the input common mode voltage and has a crossover point at 1V below V<sup>+</sup>.

This V<sub>OS</sub> crossover point can create problems for both DC and AC coupled signals if proper care is not taken. Large input signals that include the VOS crossover point will cause distortion in the output signal. One way to avoid such distortion is to keep the signal away from the crossover. For example, in a unity gain buffer configuration and with  $V_{S}$  = 5V, a 5V peak-to-peak signal will contain input-crossover distortion while a 3V peak-to-peak signal centered at 1.5V will not contain input-crossover distortion as it avoids the crossover point. Another way to avoid large signal distortion is to use a gain of -1 circuit which avoids any voltage excursions at the input terminals of the amplifier. In that circuit, the common mode DC voltage can be set at a level away from the  $V_{\rm OS}$  cross-over point. For small signals, this transition in  $V_{\rm OS}$  shows up as a  $V_{\rm CM}$  dependent spurious signal in series with the input signal and can effectively degrade small signal parameters such as gain and common mode rejection ratio. To resolve this problem, the small signal should be placed such that it avoids the  $V_{OS}$  crossover point. In addition to the rail-to-rail performance, the output stage can provide enough output current to drive  $600\Omega$  loads. Because of the high current capability, care should be taken not to exceed the 150°C maximum junction temperature specification.

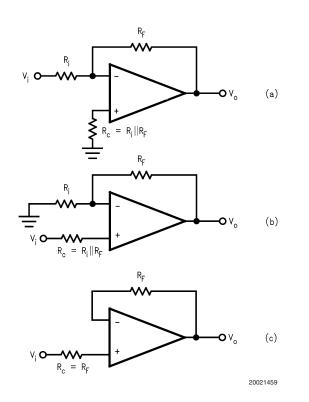
#### 2.0 SHUTDOWN MODE

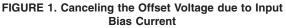
ing between the two modes.

The LMV981/LMV982 have a shutdown pin. To conserve battery life in portable applications, the LMV981/LMV982 can be disabled when the shutdown pin voltage is pulled low. The shutdown pin can't be left unconnected. In case shutdown operation is not needed, the shutdown pin should be connected to V<sup>+</sup> when the LMV981/LMV982 are used. Leaving the shutdown pin floating will result in an undefined operation mode, either shutdown or active, or even oscillat-

#### 3.0 INPUT BIAS CURRENT CONSIDERATION

The LMV981/LMV982 family has a complementary bipolar input stage. The typical input bias current (I<sub>B</sub>) is 15nA. The input bias current can develop a significant offset voltage. This offset is primarily due to I<sub>B</sub> flowing through the negative feedback resistor, R<sub>F</sub>. For example, if I<sub>B</sub> is 50nA and R<sub>F</sub> is 100k $\Omega$ , then an offset voltage of 5mV will develop (V<sub>OS</sub> = I<sub>B</sub> x R<sub>F</sub>). Using a compensation resistor (R<sub>C</sub>), as shown in *Figure 1*, cancels this effect. But the input offset current (I<sub>OS</sub>) will still contribute to an offset voltage in the same manner.

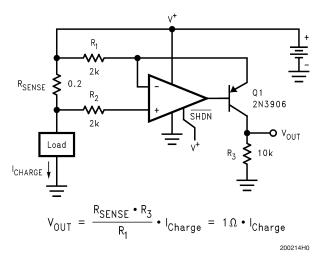


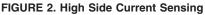


# **Typical Applications**

#### 4.0 HIGH SIDE CURRENT SENSING

The high side current sensing circuit (*Figure 2*) is commonly used in a battery charger to monitor charging current to prevent over charging. A sense resistor  $R_{SENSE}$  is connected to the battery directly. This system requires an op amp with rail-to-rail input. The LMV981/LMV982 are ideal for this application because the common mode input range goes up to the rail.





#### Typical Applications (Continued) LMV9818L "供应商 OUTPUT SWING

Since the LMV981/LMV982 input common mode range includes both positive and negative supply rails and the output can also swing to either supply, achieving half-wave rectifier functions in either direction is an easy task. All that is needed are two external resistors; there is no need for diodes or matched resistors. The half wave rectifier can have either positive or negative going outputs, depending on the way the circuit is arranged. In *Figure 3* the circuit is referenced to ground, while in *Figure 4* the circuit is biased to the positive supply. These configurations implement the half wave rectifier since the LMV981/LMV982 can not respond to one-half of the incoming waveform. It can not respond to one-half of the incoming because the amplifier can not swing the output beyond either rail therefore the output disengages during this half cycle. During the other half cycle, however, the amplifier achieves a half wave that can have a peak equal to the total supply voltage.  $R_1$  should be large enough not to load the LMV981/LMV982.

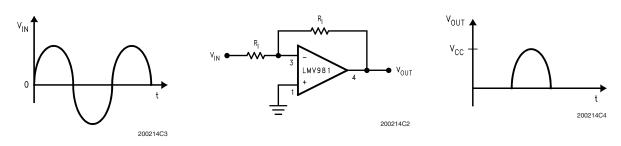


FIGURE 3. Half-Wave Rectifier with Rail-To-Ground Output Swing Referenced to Ground

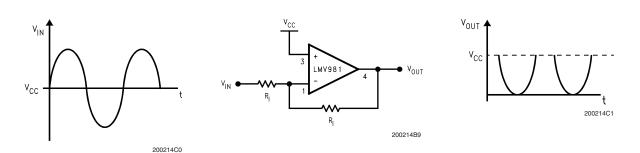


FIGURE 4. Half-Wave Rectifier with Negative-Going Output Referenced to V<sub>CC</sub>

# 6.0 INSTRUMENTATION AMPLIFIER WITH RAIL-TO-RAIL INPUT AND OUTPUT

Some manufactures make a non-"rail-to-rail"-op amp rail-torail by using a resistive divider on the inputs. The resistors divide the input voltage to get a rail-to-rail input range. The problem with this method is that it also divides the signal, so in order to get the obtained gain, the amplifier must have a higher closed loop gain. This raises the noise and drift by the internal gain factor and lowers the input impedance. Any mismatch in these precision resistors reduces the CMRR as well. The LMV981/LMV982 is rail-to-rail and therefore doesn't have these disadvantages.

Using three of the LMV981/LMV982 amplifiers, an instrumentation amplifier with rail-to-rail inputs and outputs can be made as shown in *Figure 5*.

In this example, amplifiers on the left side act as buffers to the differential stage. These buffers assure that the input impedance is very high and require no precision matched resistors in the input stage. They also assure that the difference amp is driven from a voltage source. This is necessary to maintain the CMRR set by the matching R<sub>1</sub>-R<sub>2</sub> with R<sub>3</sub>-R<sub>4</sub>. The gain is set by the ratio of R<sub>2</sub>/R<sub>1</sub> and R<sub>3</sub> should equal R<sub>1</sub> and R<sub>4</sub> equal R<sub>2</sub>. With both rail-to-rail input and output ranges, the input and output are only limited by the supply

voltages. Remember that even with rail-to-rail outputs, the output can not swing past the supplies so the combined common mode voltages plus the signal should not be greater that the supplies or limiting will occur. For additional applications, see National Semiconductor application notes AN–29, AN–31, AN–71, and AN–127.

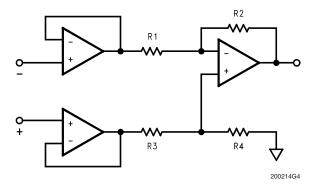
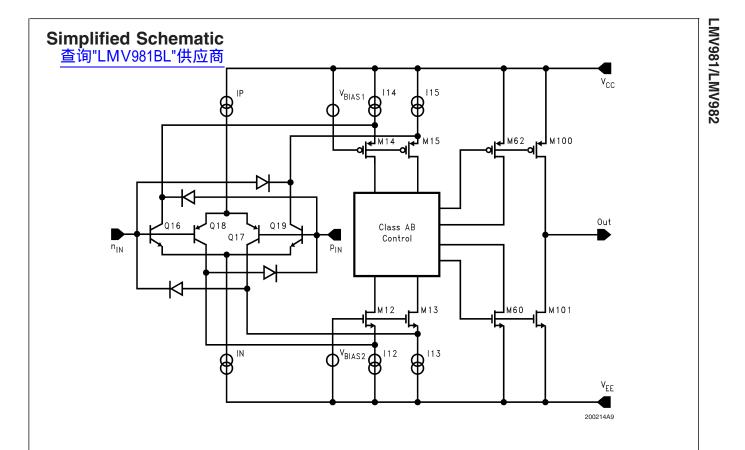
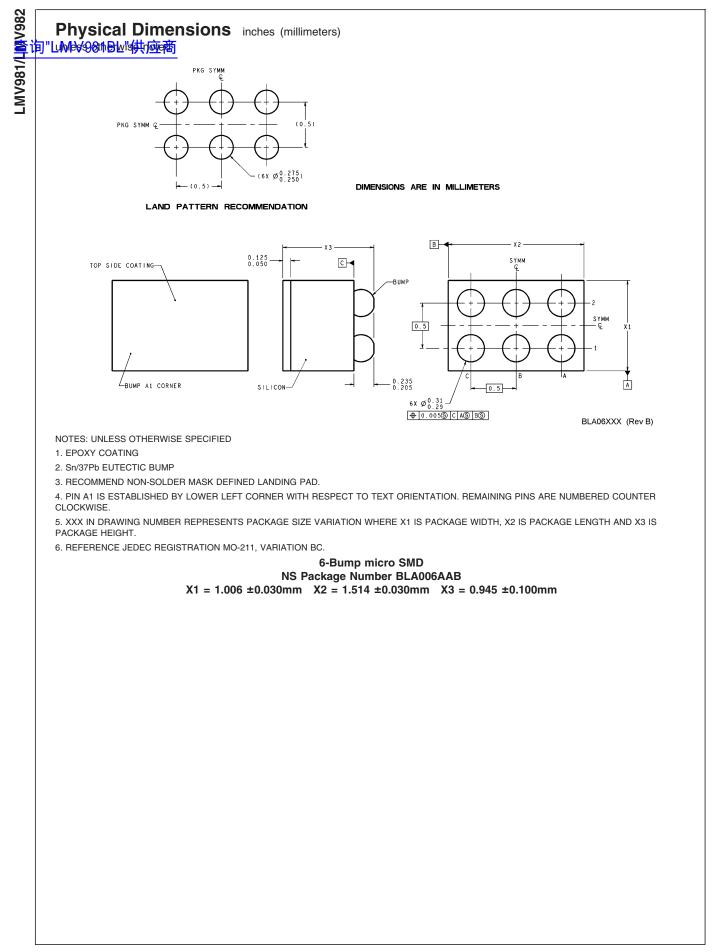
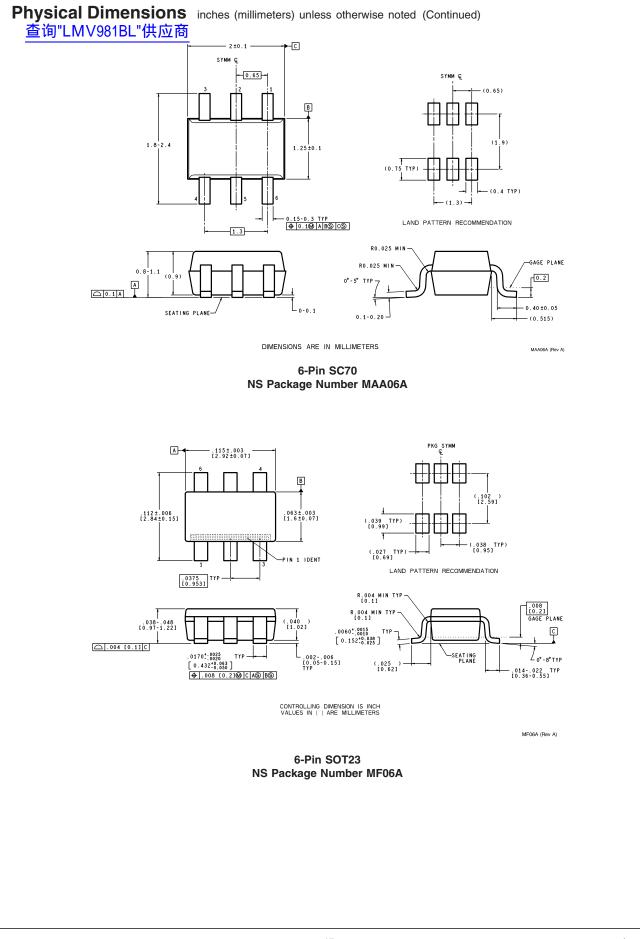
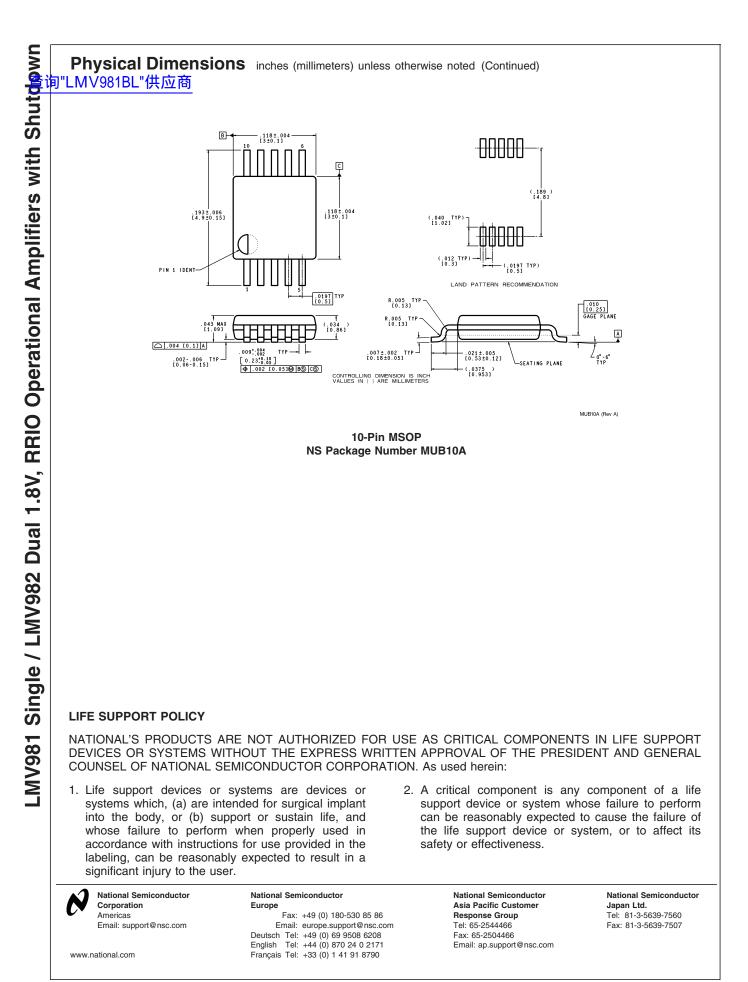


FIGURE 5. Rail-to-rail instrumentation amplifier









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