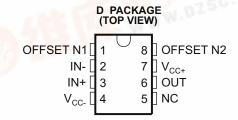
# TLE2027-EP Excalibur™ LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL AMPLIFIER

SLOS511-JUNE 2007

#### **FEATURES**

- Controlled Baseline
  - One Assembly/Test Site, One Fabrication Site
- Extended Temperature Performance of -55°C to 125°C
- Enhanced Diminishing Manufacturing Sources (DMS) Support
- Enhanced Product-Change Notification
- Qualification Pedigree<sup>(1)</sup>
- Outstanding Combination of DC Precision and AC Performance:
  - Unity-Gain Bandwidth . . . 13 MHz Typ
  - $V_n \dots 3.3 \text{ nV/}\sqrt{\text{Hz}}$  at f = 10 Hz Typ, 2.5 nV/ $\sqrt{\text{Hz}}$  at f = 1 kHz Typ
- (1) Component qualification in accordance with JEDEC and industry standards to ensure reliable operation over an extended temperature range. This includes, but is not limited to, Highly Accelerated Stress Test (HAST) or biased 85/85, temperature cycle, autoclave or unbiased HAST, electromigration, bond intermetallic life, and mold compound life. Such qualification testing should not be viewed as justifying use of this component beyond specified performance and environmental limits.

- V<sub>IO</sub> . . . 100 μV Max
- $A_{VD}$  . . . 45 V/μV Typ With  $R_L$  = 2 kΩ, 19 V/μV Typ With  $R_L$  = 600 Ω
- Available in Standard-Pinout Small-Outline Package
- Output Features Saturation Recovery Circuitry
- Macromodels and Statistical information



### **DESCRIPTION**

The TLE2027 contains innovative circuit design expertise and high-quality process control techniques to produce a level of ac performance and dc precision previously unavailable in single operational amplifiers. Manufactured using Tl's state-of-the-art Excalibur process, these devices allow upgrades to systems that use lower-precision devices.

In the area of dc precision, the TLE2027 offers maximum offset voltages of 100  $\mu$ V, common-mode rejection ratio of 131 dB (typ), supply voltage rejection ratio of 144 dB (typ), and dc gain of 45 V/ $\mu$ V (typ).

The ac performance of the TLE2027 is highlighted by a typical unity-gain bandwidth specification of 15 MHz, 55° of phase margin, and noise voltage specifications of 3.3 nV/ $\sqrt{\text{Hz}}$  and 2.5 nV/ $\sqrt{\text{Hz}}$  at frequencies of 10 Hz and 1 kHz, respectively.

The TLE2027 is available in a wide variety of packages, including the industry-standard 8-pin small-outline version for high-density system applications. The device is characterized for operation over the full military temperature range of –55°C to 125°C.

#### ORDERING INFORMATION(1)

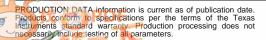
001/6 - 1	V <sub>IO</sub> max AT	PACKAGED DEVICES		
¹A	25°C	SMALL OUTLINE (2) (D)		
−55°C to 125°C	100 μV	TLE2027MDREP		

<sup>(1)</sup> For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.

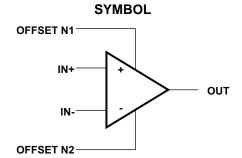
(2) The D package is available taped and reeled with 2500 units/reel.

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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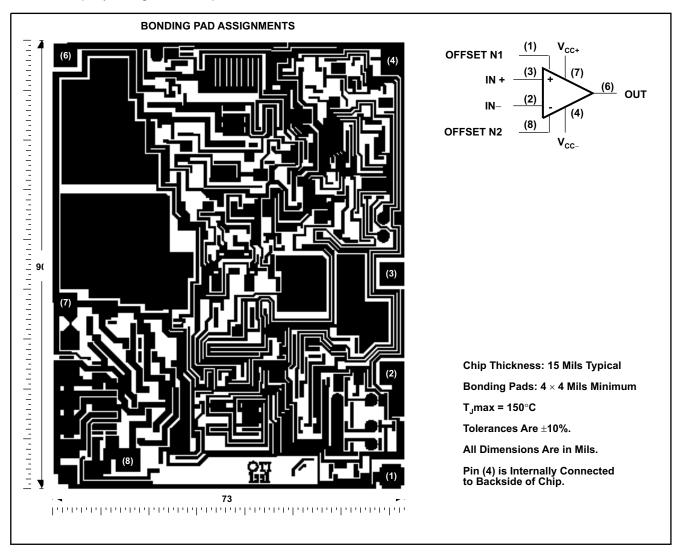






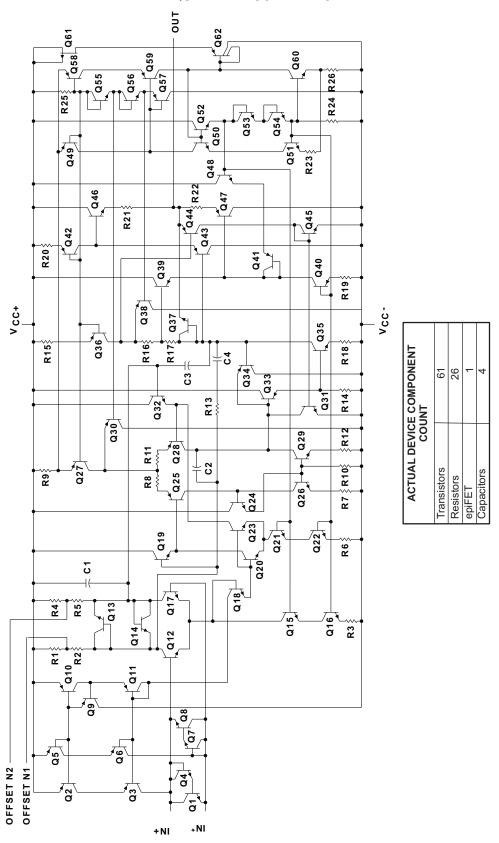
# **TLE202XY CHIP INFORMATION**

This chip, when properly assembled, displays characteristics similar to the TLE202xC. Thermal compression or ultrasonic bonding may be used on the doped-aluminum bonding pads. The chip may be mounted with conductive epoxy or a gold-silicon preform.





# **EQUIVALENT SCHEMATIC**



# Absolute Maximum Ratings(1)

over operating free-air temperature range (unless otherwise noted)

			N	IIN MAX	UNIT
V <sub>CC+</sub>	Supply voltage <sup>(2)</sup>			19	V
V <sub>CC</sub> -	Supply voltage		-19	V	
$V_{ID}$	Differential input voltage <sup>(3)</sup>			±1.2	V
VI	Input voltage range (any input)			$V_{CC\pm}$	
II	Input current (each input)			±1	mA
Io	Output current		±50	mA	
	Total current into V <sub>CC+</sub>			50	mA
	Total current out of V <sub>CC</sub> -			50	mA
	Duration of short-circuit current at (or below) 25°C (4)		l	Jnlimited	
	Continuous total power dissipation			Dissipation ating Table	
T <sub>A</sub>	Operating free-air temperature range		-	-55 125	°C
T <sub>stg</sub>	Storage temperature range <sup>(5)</sup>		-	-65 150	°C
	Lead temperature 1,6 mm (1/16 in) from case for 10 s	D package		260	°C

<sup>(1)</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values, except differential voltages, are with respect to the midpoint between V<sub>CC+</sub> and V<sub>CC-</sub>

# **Dissipation Rating Table**

PACKAGE	KAGE $T_A \le 25^{\circ}C$ DERATING FACTOR ABOVE $T_A = 25^{\circ}C$		T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 105°C POWER RATING	T <sub>A</sub> = 125°C POWER RATING	
D	725 mW	5.8 mW/°C	464 mW	261 mW	145 mW	

# **Recommended Operating Conditions**

			MIN	MAX	UNIT
$V_{CC\pm}$	Supply voltage		±4	±19	V
V <sub>IC</sub> Common-mode input voltage		T <sub>A</sub> = 25°C	-11	11	\/
		T <sub>A</sub> = Full range <sup>(1)</sup>	-10.3	10.3	V
T <sub>A</sub>	Operating free-air temperature		-55	125	°C

(1) Full range is -55°C to 125°C.

<sup>(3)</sup> Differential voltages are at IN+ with respect to IN-. Excessive current flows if a differential input voltage in excess of approximately ±1.2 V is applied between the inputs, unless some limiting resistance is used.

<sup>(4)</sup> The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

<sup>(5)</sup> Long-term high-temperature storage and/or extended use at maximum recommended operating conditions may result in a reduction of overall device life. See <a href="http://www.ti.com/ep\_quality">http://www.ti.com/ep\_quality</a> for additional information on enhanced product packaging.

# **TLE2027-EP** Excalibur™ LOW-NOISE HIGH-SPEED PRECISION OPERATIONAL AMPLIFIER

stas緬"-TUNE 20077-EP"供应商



# **Electrical Characteristics**

at specified free-air temperature,  $V_{CC\pm}$  =  $\pm 15$  V (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	T <sub>A</sub> <sup>(1)</sup>	MIN	TYP	MAX	UNIT
	land effect william	V 0. D 50.0	25°C		20	100	
$V_{IO}$	Input offset voltage	$V_{IC} = 0$ , $R_S = 50 \Omega$	Full range			200	μV
$\alpha_{VIO}$	Temperature coefficient of input offset voltage	$V_{IC} = 0$ , $R_S = 50 \Omega$	Full range		0.4		μV/°C
	Input offset voltage long-term drift <sup>(2)</sup>	$V_{IC} = 0, R_S = 50 \Omega$	25°C		0.006		μV/mo
	lanut effect current	V 0.B 50.0	25°C		6	90	nA
I <sub>IO</sub>	Input offset current	$V_{IC} = 0$ , $R_S = 50 \Omega$	Full range			150	ΠA
	lanut biog gurrant	$V_{IC} = 0, R_S = 50 \Omega$	25°C		15	90	- Λ
I <sub>IB</sub>	Input bias current	$V_{IC} = 0, R_S = 50.22$	Full range			150	nA
V <sub>ICR</sub>	Common-mode input voltage range	$R_S = 50 \Omega$	25°C	-11 to 11	–13 to 13		V
VICR	Johnnon-mode input voitage range	1/8 - 30 22	Full range	-10.3 to 10.3			V
	Maximum positive peak output voltage swing	D 600 O	25°C	10.5	12.9		
V		$R_L = 600 \Omega$	Full range	10			V
V <sub>OM+</sub>		$R_L = 2 k\Omega$	25°C	12	13.2		
			Full range	11			
	Maximum negative peak output voltage swing	$R_{L} = 600 \Omega$ $R_{L} = 2 k\Omega$	25°C	-10.5	-13		V
V <sub>OM</sub> _			Full range	-10			
v OM-			25°C	-12	-13.5		
		IV[ = 2 K22	Full range	-11			
		$V_O = \pm 11 \text{ V}, R_L = 2 \text{ k}\Omega$	25°C	5	45		
		$V_O = \pm 10 \text{ V}, R_L = 2 \text{ k}\Omega$	Full range	2.5			
$A_{VD}$	Large-signal differential voltage amplification	$V_O = \pm 10 \text{ V}, R_L = 1 \text{ k}\Omega$	25°C	3.5	38		V/µV
		V <sub>0</sub> = ±10 V, K <sub>L</sub> = 1 K22	Full range	1.8			
		$V_O = \pm 10 \text{ V}, R_L = 600 \Omega$	25°C	2	19		
$C_{i}$	Input capacitance		25°C		8		pF
z <sub>o</sub>	Open-loop output impedance	I <sub>O</sub> = 0	25°C		50		Ω
CMRR	Common-mode rejection ratio	$V_{IC} = V_{ICR}min,$	25°C	100	131		dB
CIVIKK	Common-mode rejection ratio	$R_S = 50 \Omega$	Full range	96			uБ
k <sub>SVR</sub>	Supply-voltage rejection ratio ( $\Delta V_{CC\pm}/\Delta V_{IO}$ )	$V_{CC\pm} = \pm 4 \text{ V to } \pm 18 \text{ V},$ $R_S = 50 \Omega$	25°C	94	144		dB
··ovk		$V_{CC\pm} = \pm 4 \text{ V to } \pm 18 \text{ V},$ $R_S = 50 \Omega$	Full range	90			32
I <sub>CC</sub>	Supply current	V <sub>O</sub> = 0, No load	25°C		3.8	5.3	mA
-00	очерну очноги	10 = 0, 110 load	Full range			5.6	1117 (

 <sup>(1)</sup> Full range is -55°C to 125°C.
 (2) Typical values are based on the input offset voltage shift observed through 168 hours of operating life test at T<sub>A</sub> = 150°C extrapolated to T<sub>A</sub> = 25°C using the Arrhenius equation and assuming an activation energy of 0.96 eV.

# **Operating Characteristics**

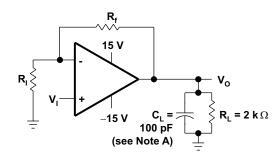
at specified free-air temperature,  $V_{CC\pm} = \pm 15$  V,  $T_A = 25^{\circ}C$  (unless otherwise noted)

	PARAMETER		TEST CONDITIONS		TYP	MAX	UNIT
		$R_L = 2 k\Omega$ , $C_L = 100 pF$ , See Figure 1		1.7	2.8		
SR Slew rate at unity gain		$R_L = 2 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$ , $T_A = -55^{\circ}\text{C}$ to 125°C, See Figure 1		1			V/µs
V	Equivalent input poice valtage (see Figure 2)	D 20.0	f = 10 Hz		3.3		nV/√ <del>Hz</del>
V <sub>n</sub>	Equivalent input noise voltage (see Figure 2)	$R_S = 20 \Omega$	f = 1 kHz		2.5	IIV/VII.	IIV/√⊓Z
V <sub>N(PP)</sub>	Peak-to-peak equivalent input noise voltage	f = 0.1 Hz to	10 Hz		50		nV
	Cavityalant input paiga augrent	f = 10 Hz			1.5		5 A /s/II=
In	In Equivalent input noise current		f = 1 kHz		0.4		pA/√ <del>Hz</del>
THD	Total harmonic distortion	$V_O = 10 \text{ V}, A_{VD} = 1^{(1)}$			<0.002%		
B <sub>1</sub>	Unity-gain bandwidth (see Figure 3)	$R_L = 2 \text{ k}\Omega$ , $C_L = 100 \text{ pF}$			13		MHz
B <sub>OM</sub>	Maximum output-swing bandwidth	$R_L = 2 k\Omega$			30		kHz
$\phi_{\text{m}}$	Phase margin at unity gain (see Figure 3)	$R_L = 2 k\Omega$ , 0	C <sub>L</sub> = 100 pF		55°		

<sup>(1)</sup> Measured distortion of the source used in the analysis was 0.002%.

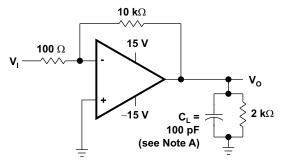


# PARAMETER MEASUREMENT INFORMATION



NOTE A: C<sub>1</sub> includes fixture capacitance.

Figure 1. Slew-Rate Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 3. Unity-Gain Bandwidth and Phase-Margin Test Circuit

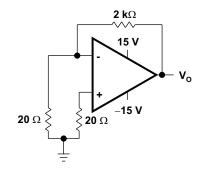
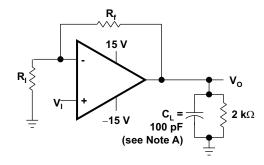


Figure 2. Noise-Voltage Test Circuit



NOTE A:  $C_L$  includes fixture capacitance.

Figure 4. Small-Signal Pulse-Response Test Circuit



# Typical Values

Typical values presented in this data sheet represent the median (50% point) of device parametric performance.

#### **Initial Estimates of Parameter Distributions**

In the ongoing program of improving data sheets and supplying more information to our customers, Texas Instruments has added an estimate of not only the typical values but also the spread around these values. These are in the form of distribution bars that show the 95% (upper) points and the 5% (lower) points from the characterization of the initial wafer lots of this new device type (see Figure 5). The distribution bars are shown at the points where data was actually collected. The 95% and 5% points are used instead of  $\pm 3$  sigma since some of the distributions are not true Gaussian distributions.

The number of units tested and the number of different wafer lots used are on all of the graphs where distribution bars are shown. As noted in Figure 5, there were a total of 835 units from two wafer lots. In this case, there is a good estimate for the within-lot variability and a possibly poor estimate of the lot-to-lot variability. This is always the case on newly released products since there can only be data available from a few wafer lots.

The distribution bars are not intended to replace the minimum and maximum limits in the electrical tables. Each distribution bar represents 90% of the total units tested at a specific temperature. While 10% of the units tested fell outside any given distribution bar, this should not be interpreted to mean that the same individual devices fell outside every distribution bar.

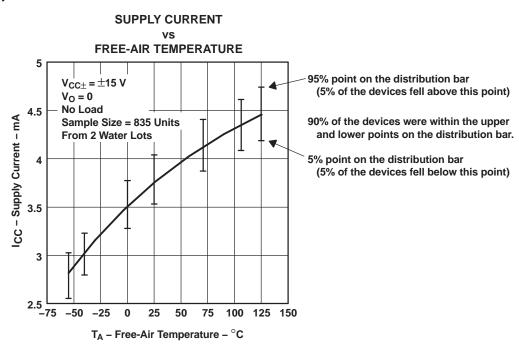


Figure 5. Sample Graph With Distribution Bars



# **TYPICAL CHARACTERISTICS**

# **Table of Graphs**

			FIGURE
V <sub>IO</sub>	Input offset voltage	Distribution	6,
$\Delta V_{IO}$	Input offset voltage change	vs Time after power on	7, 8
I <sub>IO</sub>	Input offset current	vs Free-air temperature	9
	Input higo ourrent	vs Free-air temperature	10
I <sub>IB</sub>	Input bias current	vs Common-mode input voltage	11
I <sub>I</sub>	Input current	vs Differential input voltage	12
V <sub>O(PP)</sub>	Maximum peak-to-peak output voltage	vs Frequency	13, 14
	Maximum (positive/secative) poek autout voltage	vs Load resistance	15, 16
$V_{OM}$	Maximum (positive/negative) peak output voltage	vs Free-air temperature	17, 18
		vs Supply voltage	19
$A_{VD}$	Lorge cional differential valte as complification	vs Load resistance	20
	Large-signal differential voltage amplification	vs Frequency	21, 22
		vs Free-air temperature	23
z <sub>o</sub>	Output impedance	vs Frequency	24
CMRR	Common-mode rejection ratio	vs Frequency	25
k <sub>SVR</sub>	Supply-voltage rejection ratio	vs Frequency	26
		vs Supply voltage	27, 28
los	Short-circuit output current	vs Elapsed time	29, 30
		vs Free-air temperature	31, 32
1	Complex account	vs Supply voltage	33
lcc	Supply current	vs Free-air temperature	34
	Valtaria fallacciar neclas reconers	Small signal	35
	Voltage-follower pulse response	Large signal	36
V <sub>n</sub>	Equivalent input noise voltage	vs Frequency	37
	Noise voltage (referred to input)	Over 10-s interval	38
<b>D</b>	Heito ania langakoidak	vs Supply voltage	39
B <sub>1</sub>	Unity-gain bandwidth	vs Load capacitance	40
SR	Slew rate	vs Free-air temperature	41
		vs Supply voltage	42
Φm	Phase margin	vs Loadcapacitance	43
		vs Free-air temperature	44



#### **TYPICAL CHARACTERISTICS**

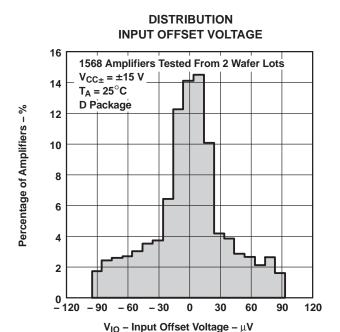


Figure 6.

# INPUT OFFSET VOLTAGE CHANGE vs TIME AFTER POWER ON

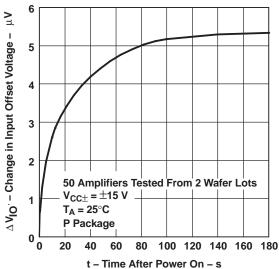


Figure 8.

# INPUT OFFSET VOLTAGE CHANGE vs TIME AFTER POWER ON

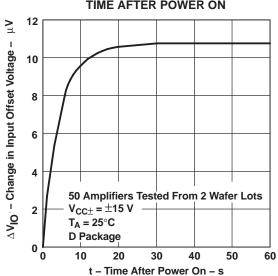
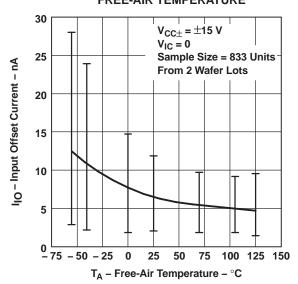


Figure 7.

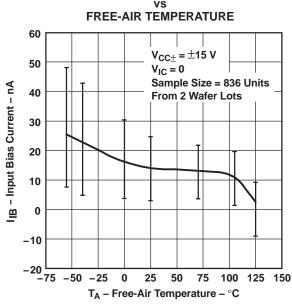
### INPUT OFFSET CURRENT vs FREE-AIR TEMPERATURE



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 9.

# INPUT BIAS CURRENT



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

#### Figure 10.

# INPUT CURRENT vs DIFFERENTIAL INPUT VOLTAGE

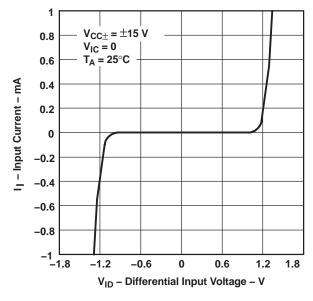


Figure 12.

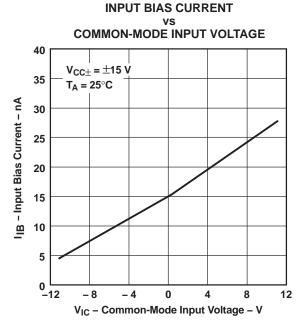
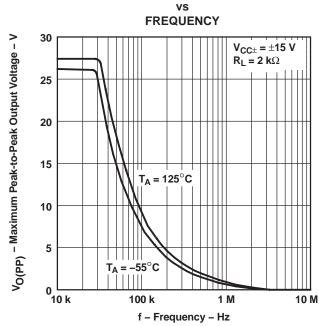


Figure 11.

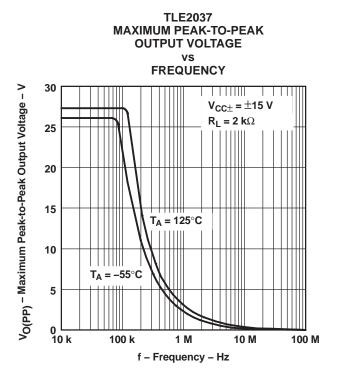
# MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 13.





NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 14.

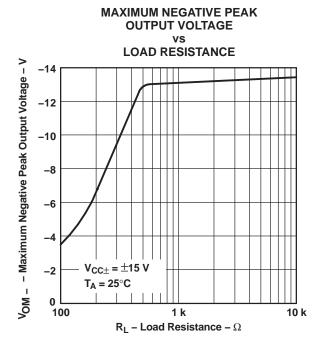


Figure 16.

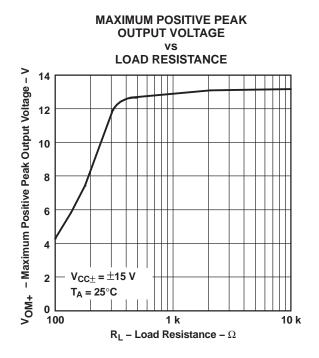
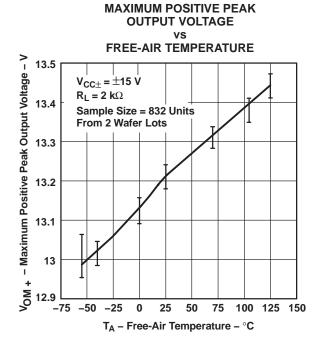


Figure 15.



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 17.



# **MAXIMUM NEGATIVE PEAK OUTPUT VOLTAGE** FREE-AIR TEMPERATURE - Maximum Negative Peak Output Voltage - V -13 $V_{CC\pm} = \pm 15 \text{ V}$ $R_L = 2 k\Omega$ Sample Size = 831 Units -13.2 From 2 Wafer Lots -13.4 -13.6-13.8 VOM--75 -50 25 50 75 100 T<sub>A</sub> - Free-Air Temperature - °C

NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

# Figure 18. **LARGE-SIGNAL DIFFERENTIAL**

**VOLTAGE AMPLIFICATION** 

# vs **LOAD RESISTANCE** $V_{CC\pm} = \pm 15 \text{ V}$ $T_{\Delta} = 25^{\circ}C$ A<sub>VD</sub> – Large-Signal differential Voltage Amplification − V/µV 40 30 20 10

 $R_L$  – Load Resistance –  $\Omega$ Figure 20.

1 k

2 k

4 k

# LARGE-SIGNAL DIFFERENTIAL **VOLTAGE AMPLIFICATION** SUPPLY VOLTAGE $R_L = 2 k\Omega$

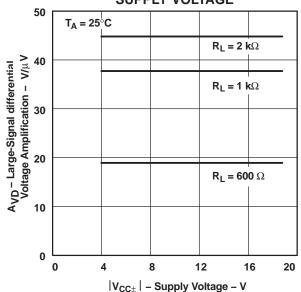


Figure 19.

# LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT

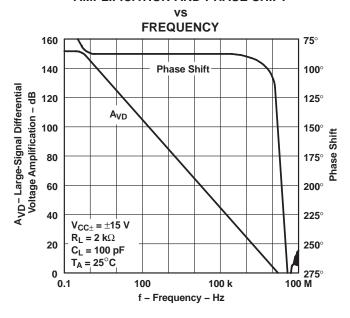


Figure 21.

100

200

400

10 k

30

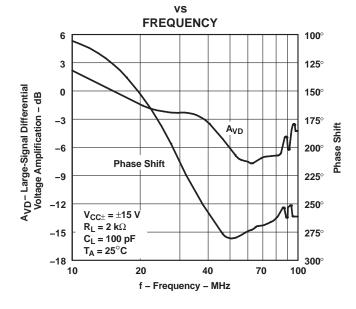
**-75 -50** 

-25 0

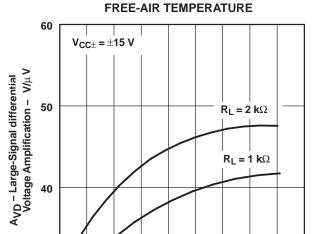


# TYPICAL CHARACTERISTICS (continued)

# LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT



# LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

#### Figure 22.

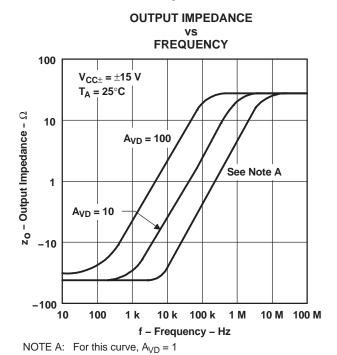


Figure 24.

## Figure 23.

25

T<sub>A</sub> - Free-Air Temperature - °C

50

75 100

125 150

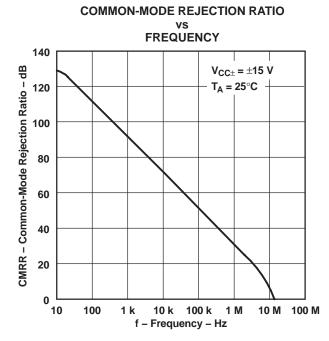


Figure 25.



# SUPPLY-VOLTAGE REJECTION RATIO

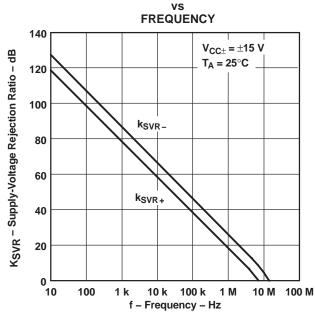


Figure 26.

# SHORT-CIRCUIT OUTPUT CURRENT vs

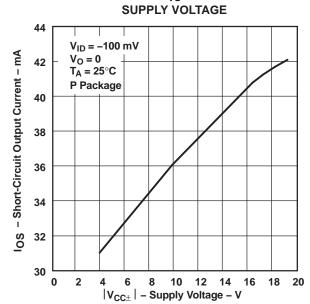


Figure 28.

# SHORT-CIRCUIT OUTPUT CURRENT

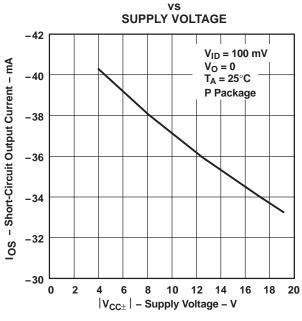


Figure 27.

# SHORT-CIRCUIT OUTPUT CURRENT vs

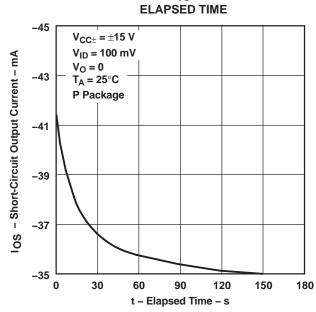
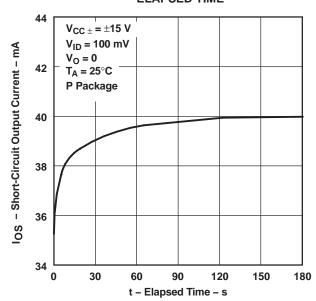


Figure 29.

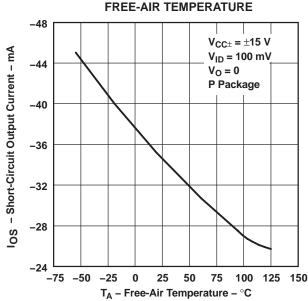


# SHORT-CIRCUIT OUTPUT CURRENT

#### vs ELAPSED TIME



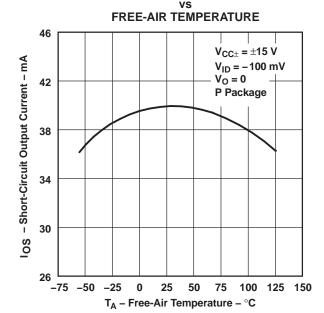
## SHORT-CIRCUIT OUTPUT CURRENT vs FREE-AIR TEMPERATURE



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

#### Figure 30.

# SHORT-CIRCUIT OUTPUT CURRENT

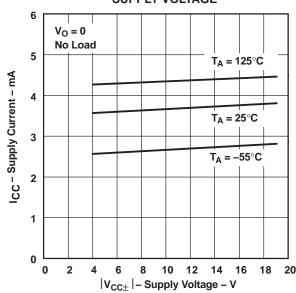


NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 32.

#### Figure 31.

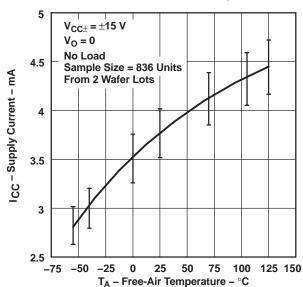
#### SUPPLY CURRENT vs SUPPLY VOLTAGE



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 33.

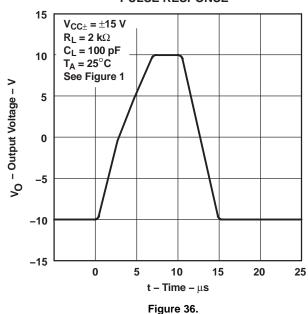
#### SUPPLY CURRENT vs FREE-AIR TEMPERATURE



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 34.

# VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE



VOLTAGE-FOLLOWER
SMALL-SIGNAL
PULSE RESPONSE

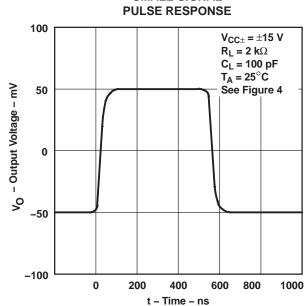


Figure 35.

# EQUIVALENT INPUT NOISE VOLTAGE vs

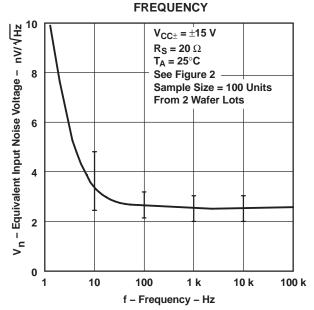
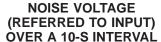


Figure 37.





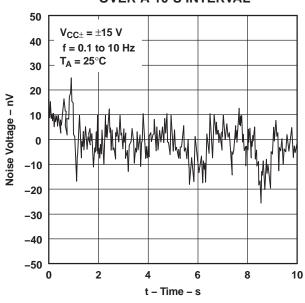


Figure 38.

## **UNITY-GAIN BANDWIDTH**

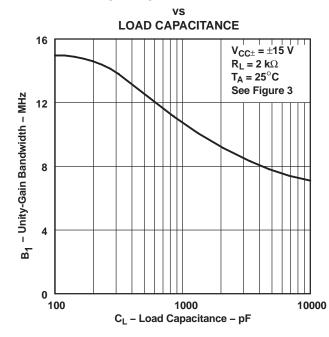


Figure 40.

# UNITY-GAIN BANDWIDTH vs SUPPLY VOLTAGE

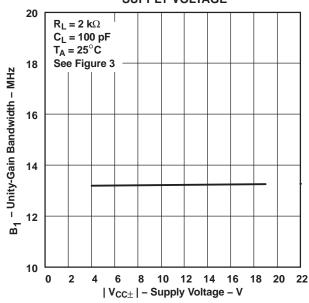
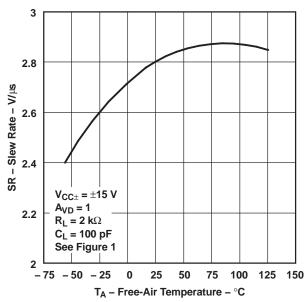


Figure 39.

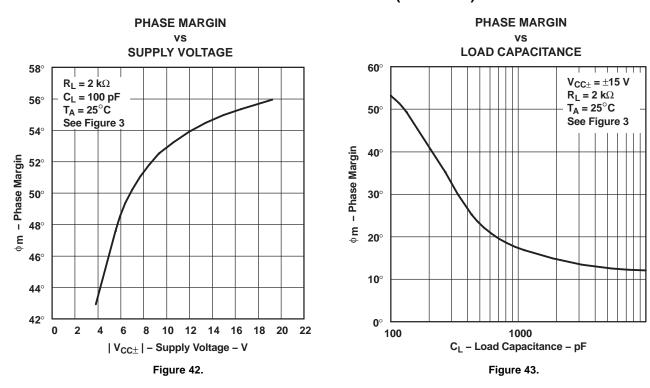
# SLEW RATE vs FREE-AIR TEMPERATURE



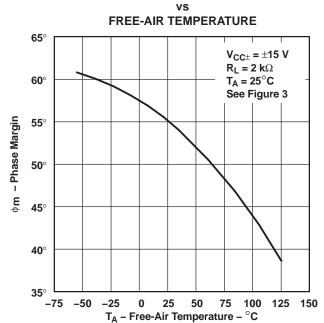
NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 41.





# PHASE MARGIN



NOTE A: Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

Figure 44.

#### **APPLICATION INFORMATION**

## **Input Offset Voltage Nulling**

The TLE2027 series offers external null pins that can be used to further reduce the input offset voltage. The circuits of Figure 45 can be connected as shown if the feature is desired. If external nulling is not needed, the null pins may be left disconnected.

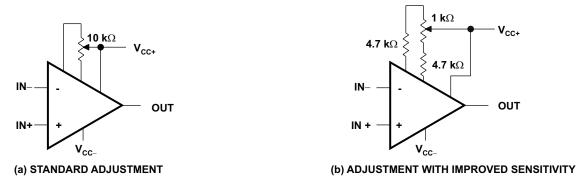


Figure 45. Input Offset Voltage Nulling Circuits

#### **Voltage-Follower Applications**

The TLE2027 circuitry includes input-protection diodes to limit the voltage across the input transistors; however, no provision is made in the circuit to limit the current if these diodes are forward biased. This condition can occur when the device is operated in the voltage-follower configuration and driven with a fast, large-signal pulse. It is recommended that a feedback resistor be used to limit the current to a maximum of 1 mA to prevent degradation of the device. Also, this feedback resistor forms a pole with the input capacitance of the device. For feedback resistor values greater than 10 k $\Omega$ , this pole degrades the amplifier phase margin. This problem can be alleviated by adding a capacitor (20 pF to 50 pF) in parallel with the feedback resistor (see Figure 46).

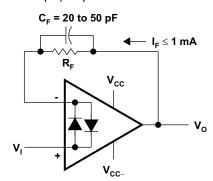


Figure 46. Voltage Follower

## **APPLICATION INFORMATION (continued)**

#### **Macromodel Information**

Macromodel information provided was derived using Microsim Parts<sup>™</sup>, the model generation software used with Microsim PSpice<sup>™</sup>. The Boyle macromodel (see Note and Figure 47) and subcircuit (see Figure 48) were generated using the TLE202x7 typical electrical and operating characteristics at 25°C. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- · Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification

- · Gain-bandwidth product
- Common-mode rejection ratio
- · Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

#### NOTE:

G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Integrated Circuit Operational Amplifiers", IEEE Journal of Solid-State Circuits, SC-9, 353 (1974).

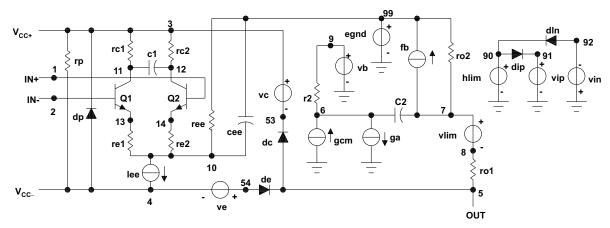


Figure 47. Boyle Macromodel

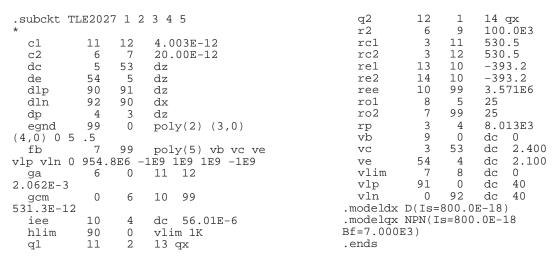


Figure 48. TLE2027 Macromodel Subcircuit



18-Sep-2008

#### PACKAGING INFORMATION

0	Orderable Device	Status (1)	Package Type	Package Drawing	Pins P	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
Т	TLE2027MDREP	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
\	V62/06674-01XE	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND**: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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#### OTHER QUALIFIED VERSIONS OF TLE2027-EP:

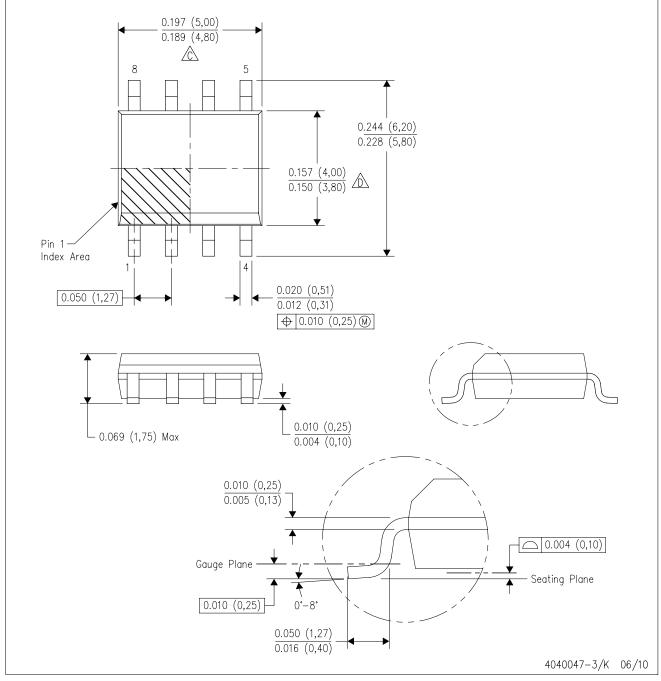
Catalog: TLE2027Military: TLE2027M

NOTE: Qualified Version Definitions:

- Catalog TI's standard catalog product
- Military QML certified for Military and Defense Applications

# D (R-PDSO-G8)

# PLASTIC SMALL-OUTLINE PACKAGE



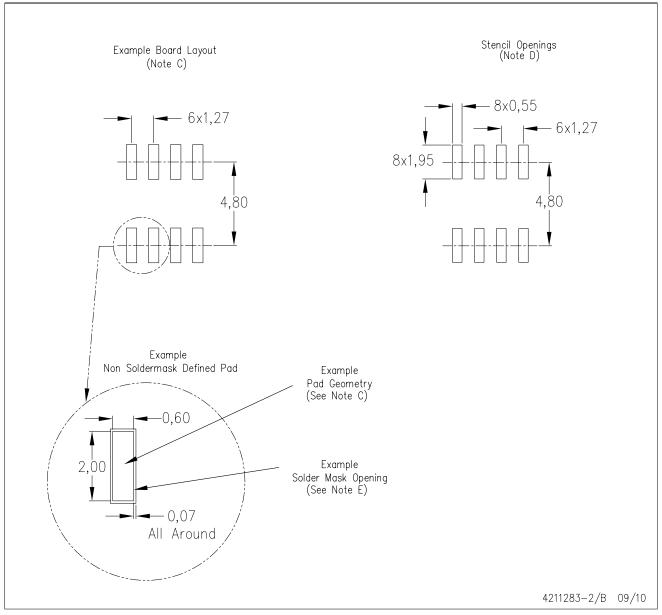
NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
- Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
- E. Reference JEDEC MS-012 variation AA.



# D (R-PDSO-G8)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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