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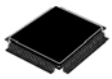
RoHS

MPC5607B


100 LQFP
14 mm x 14 mm


144 LQFP
20 mm x 20 mm


176 LQFP
24 mm x 24 mm


208 MAPBGA
17 mm x 17 mm

MPC5607B Microcontroller Data Sheet

- Single issue, 32-bit CPU core complex (e200z0h)
  - Compliant with the Power Architecture® technology embedded category
  - Enhanced instruction set allowing variable length encoding (VLE) for code size footprint reduction. With the optional encoding of mixed 16-bit and 32-bit instructions, it is possible to achieve significant code size footprint reduction.
- Up to 1.5 MB on-chip code flash memory supported with the flash memory controller
- 64 (4 × 16) KB on-chip data flash memory with ECC
- Up to 96 KB on-chip SRAM
- Memory protection unit (MPU) with 8 region descriptors and 32-byte region granularity on certain family members (Refer to Table 1 for details.)
- Interrupt controller (INTC) capable of handling 204 selectable-priority interrupt sources
- Frequency modulated phase-locked loop (FMPLL)
- Crossbar switch architecture for concurrent access to peripherals, Flash, or RAM from multiple bus masters
- 16-channel eDMA controller with multiple transfer request sources using DMA multiplexer
- Boot assist module (BAM) supports internal Flash programming via a serial link (CAN or SCI)
- Timer supports I/O channels providing a range of 16-bit input capture, output compare, and pulse width modulation functions (eMIOS)
- 2 analog-to-digital converters (ADC): one 10-bit and one 12-bit
- Cross Trigger Unit to enable synchronization of ADC conversions with a timer event from the eMIOS or PIT
- Up to 6 serial peripheral interface (DSPI) modules
- Up to 10 serial communication interface (LINFlex) modules
- Up to 6 enhanced full CAN (FlexCAN) modules with configurable buffers
- 1 inter-integrated circuit (I^2C) interface module
- Up to 149 configurable general purpose pins supporting input and output operations (package dependent)
- Real-Time Counter (RTC)
  - Clock source from internal 128 kHz or 16 MHz oscillator supporting autonomous wakeup with 1 ms resolution with maximum timeout of 2 seconds
  - Optional support for RTC with clock source from external 32 kHz crystal oscillator, supporting wakeup with 1 sec resolution and maximum timeout of 1 hour
- Up to 8 periodic interrupt timers (PIT) with 32-bit counter resolution
- Nexus development interface (NDI) per IEEE-ISTO 5001-2003 Class Two Plus
- Device/board boundary scan testing supported per Joint Test Action Group (JTAG) of IEEE (IEEE 1149.1)
- On-chip voltage regulator (VREG) for regulation of input supply for all internal levels

freescale™ semiconductor

Table of Contents

1 Introduction

1.1 Document overview

This document describes the features of the family and options available within the family members, and highlights important electrical and physical characteristics of the device.

1.2 Description

This family of 32-bit system-on-chip (SoC) microcontrollers is the latest achievement in integrated automotive application controllers. It belongs to an expanding family of automotive-focused products designed to address the next wave of body electronics applications within the vehicle.

The advanced and cost-efficient e200z0 host processor core of this automotive controller family complies with the Power Architecture technology and only implements the VLE (variable-length encoding) APU (Auxiliary Processor Unit), providing improved code density. It operates at speeds of up to 64 MHz and offers high performance processing optimized for low power consumption. It capitalizes on the available development infrastructure of current Power Architecture devices and is supported with software drivers, operating systems and configuration code to assist with users implementations.

Table 1. MPC5607B family comparison[1]

Feature	MPC5605B			MPC5606B	MPC5607B
CPU	e200z0h				
Execution speed[2]	Up to 64 MHz				
Code Flash	768 KB			1 MB	1.5 MB
Data Flash	64 (4 x 16) KB				
SRAM	64 KB			80 KB	96 KB
MPU	8-entry				
eDMA	16 ch				
10-bit ADC	Yes				
dedicated[3]	7 ch	15 ch	29 ch	15 ch	29 ch
shared with 12-bit ADC	19 ch				
12-bit ADC	Yes				
dedicated[4]	5 ch				
shared with 10-bit ADC	19 ch				
Total timer I/O[5] eMIOS	37 ch, 16-bit	64 ch, 16-bit			
Counter / OPWM / ICOC[6]	10 ch				
O(I)PWM / OPWFMB / OPWMCB / ICOC[7]	7 ch				
O(I)PWM / ICOC[8]	7 ch	14 ch			
OPWM / ICOC[9]	13 ch	33 ch			
SCI (LINFlex)	4	6	8	6	8 10
SPI (DSPI)	3	5	6	5	6

Table 1. MPC5607B family comparison[1] (continued)

Feature	MPC5605B			MPC5606B		MPC5607B	
CAN (FlexCAN)	6						
I^2C	1						
32 kHz oscillator	Yes						
GPIO[10]	77	121	149	121	149	149	
Debug	JTAG						N2+
Package	100 LQFP	144 LQFP	176 LQFP	144 LQFP	176 LQFP	176 LQFP	208 MAP BGA[11]

1 Feature set dependent on selected peripheral multiplexing; table shows example.

2 Based on 105 °C ambient operating temperature.

3 Not shared with 12-bit ADC, but possibly shared with other alternate functions.

4 Not shared with 10-bit ADC, but possibly shared with other alternate functions.

5 Refer to eMIOS section of device reference manual for information on the channel configuration and functions.

6 Each channel supports a range of modes including Modulus counters, PWM generation, Input Capture, Output Compare.

7 Each channel supports a range of modes including PWM generation with dead time, Input Capture, Output Compare.

8 Each channel supports a range of modes including PWM generation, Input Capture, Output Compare, Period and Pulse width measurement.

9 Each channel supports a range of modes including PWM generation, Input Capture, and Output Compare.

10 Maximum I/O count based on multiplexing with peripherals.

11 208 MAPBGA available only as development package for Nexus2+.

2 Block diagram

Figure 1shows a top-level block diagram of the MPC5607B.

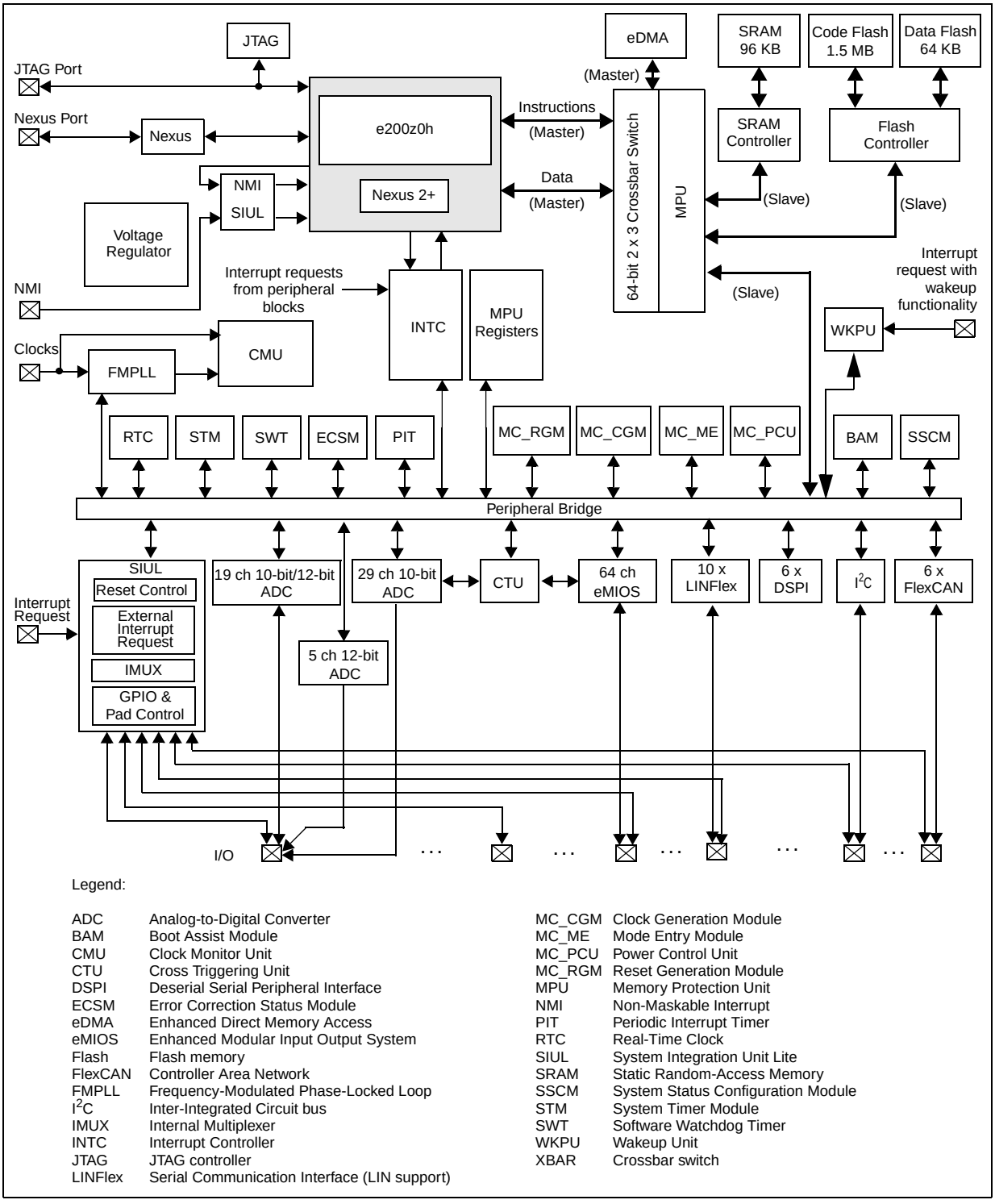


Figure 1. MPC5607B block diagram

Table 2 summarizes the functions of the blocks present on the MPC5607B.

Table 2. MPC5607B series block summary

Block	Function
Analog-to-digital converter (ADC)	Converts analog voltages to digital values
Boot assist module (BAM)	A block of read-only memory containing VLE code which is executed according to the boot mode of the device
Clock generation module (MC_CGM)	Provides logic and control required for the generation of system and peripheral clocks
Clock monitor unit (CMU)	Monitors clock source (internal and external) integrity
Cross triggering unit (CTU)	Enables synchronization of ADC conversions with a timer event from the eMIOS or from the PIT
Crossbar switch (XBAR)	Supports simultaneous connections between two master ports and three slave ports. The crossbar supports a 32-bit address bus width and a 64-bit data bus width.
Deserial serial peripheral interface (DSPI)	Provides a synchronous serial interface for communication with external devices
Enhanced Direct Memory Access (eDMA)	Performs complex data transfers with minimal intervention from a host processor via "*n*" programmable channels
Enhanced modular input output system (eMIOS)	Provides the functionality to generate or measure events
Error Correction Status Module (ECSM)	Provides a myriad of miscellaneous control functions for the device including program-visible information about configuration and revision levels, a reset status register, wakeup control for exiting sleep modes, and optional features such as information on memory errors reported by error-correcting codes
Flash memory	Provides non-volatile storage for program code, constants and variables
FlexCAN (controller area network)	Supports the standard CAN communications protocol
Frequency-modulated phase-locked loop (FMPLL)	Generates high-speed system clocks and supports programmable frequency modulation
Inter-integrated circuit (I^2C™) bus	A two wire bidirectional serial bus that provides a simple and efficient method of data exchange between devices
Internal multiplexer (IMUX) SIU subblock	Allows flexible mapping of peripheral interface on the different pins of the device
Interrupt controller (INTC)	Provides priority-based preemptive scheduling of interrupt requests
JTAG controller	Provides the means to test chip functionality and connectivity while remaining transparent to system logic when not in test mode
LINFlex controller	Manages a high number of LIN (Local Interconnect Network protocol) messages efficiently with a minimum of CPU load
Memory protection unit (MPU)	Provides hardware access control for all memory references generated in a device
Mode entry module (MC_ME)	Provides a mechanism for controlling the device operational mode and modetransition sequences in all functional states; also manages the power control unit, reset generation module and clock generation module, and holds the configuration, control and status registers accessible for applications

Table 2. MPC5607B series block summary (continued)

Block	Function
Non-Maskable Interrupt (NMI)	Handles external events that must produce an immediate response, such as power down detection
Periodic interrupt timer (PIT)	Produces periodic interrupts and triggers
Power control unit (MC_PCU)	Reduces the overall power consumption by disconnecting parts of the device from the power supply via a power switching device; device components are grouped into sections called “power domains” which are controlled by the PCU
Real-time counter (RTC)	A free running counter used for time keeping applications, the RTC can be configured to generate an interrupt at a predefined interval independent of the mode of operation (run mode or low-power mode)
Reset generation module (MC_RGM)	Centralizes reset sources and manages the device reset sequence of the device
Static random-access memory (SRAM)	Provides storage for program code, constants, and variables
System integration unit lite (SIUL)	Provides control over all the electrical pad controls and up 32 ports with 16 bits of bidirectional, general-purpose input and output signals and supports up to 32 external interrupts with trigger event configuration
System status and configuration module (SSCM)	Provides system configuration and status data (such as memory size and status, device mode and security status), device identification data, debug status port enable and selection, and bus and peripheral abort enable/disable
System timer module (STM)	Provides a set of output compare events to support AUTOSAR (AUTomotive Open System ARchitecture) and operating system tasks
System watchdog timer (SWT)	Provides protection from runaway code
WKPU (wakeup unit)	The wakeup unit supports up to 18 external sources that can generate interrupts or wakeup events, of which 1 can cause non-maskable interrupt requests or wakeup events.

3 Package pinouts and signal descriptions

3.1 Package pinouts

The available LQFP pinouts and the ballmap are provided in the following figures. For pin signal descriptions, please see Table 3.

Figure 2 shows the MPC5607B in the 176 LQFP package.

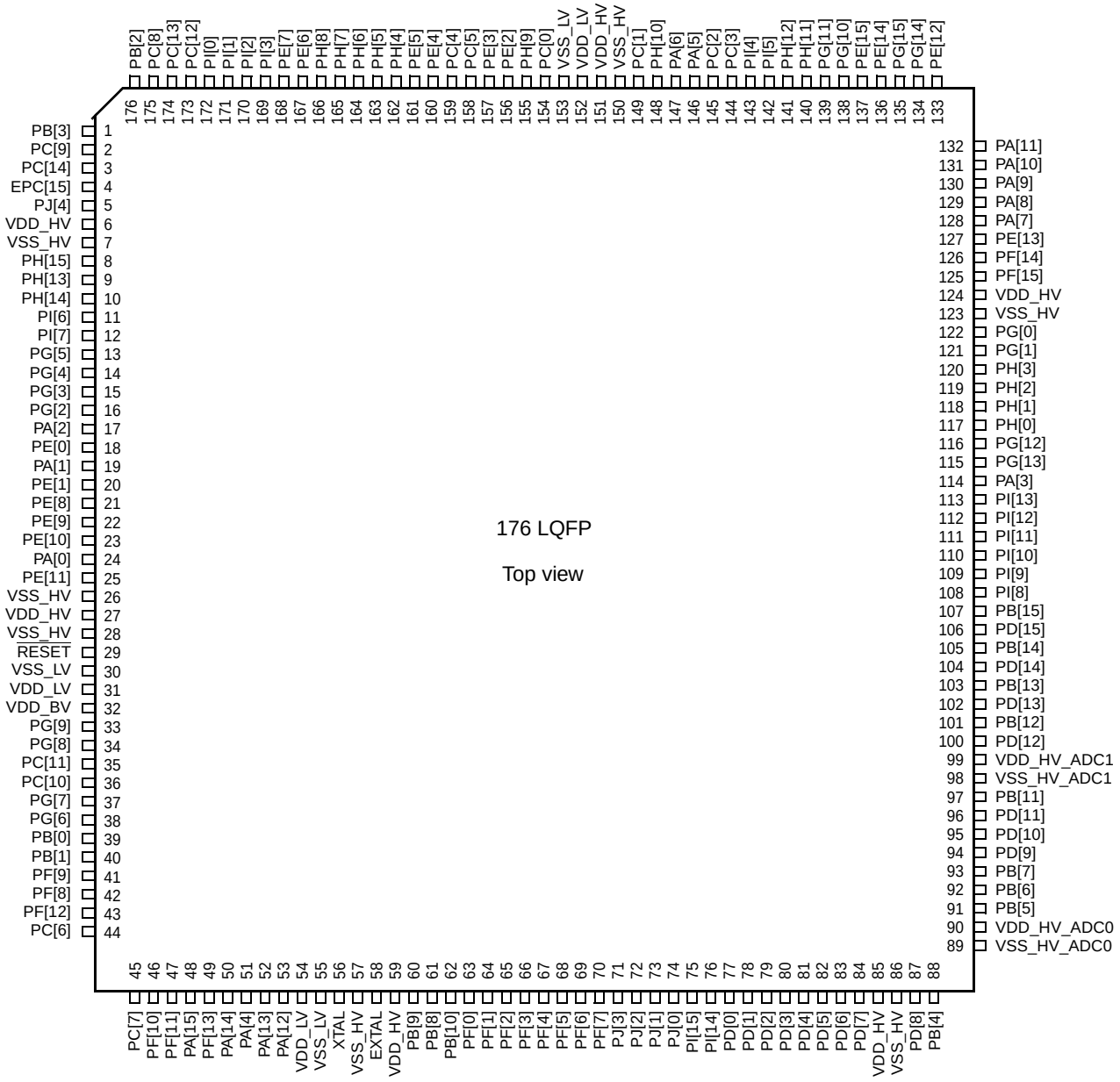


Figure 2. 176 LQFP pin configuration (top view)

Figure 3 shows the MPC5607B in the 144 LQFP package.

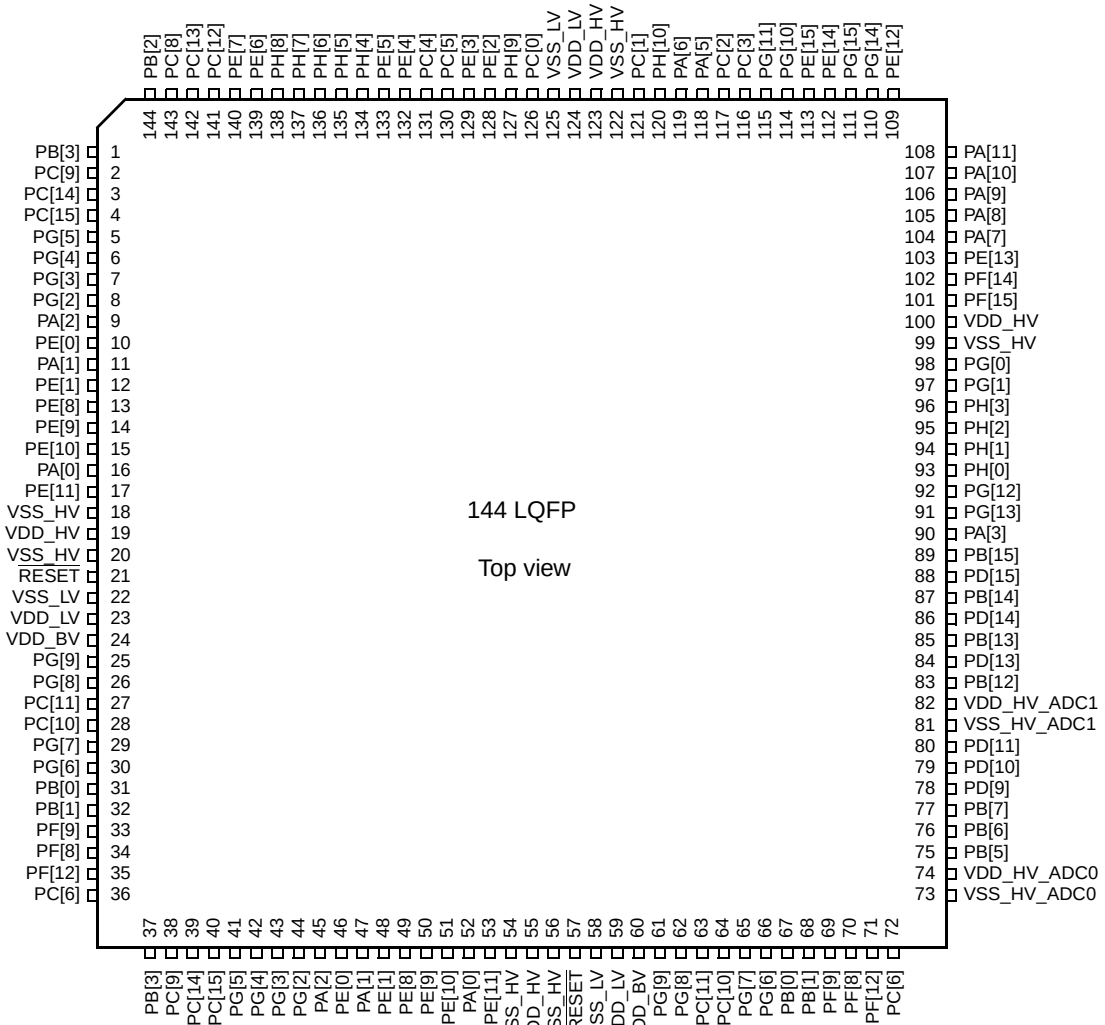


Figure 3. 144 LQFP pin configuration (top view)

Figure 4 shows the MPC5607B in the 100 LQFP package.

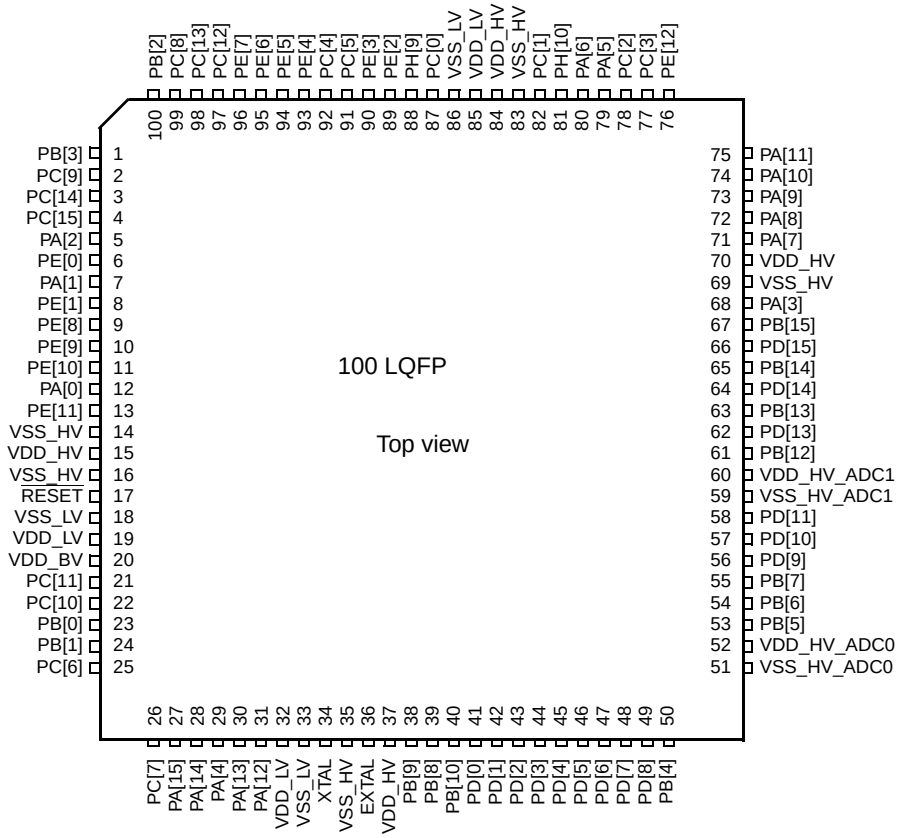


Figure 4. 100 LQFP pin configuration (top view)

Figure 5 shows the MPC5607B in the 208 MAPBGA package.

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	
A	PC[8]	PC[13]	PH[15]	PJ[4]	PH[8]	PH[4]	PC[5]	PC[0]	PI[0]	PI[1]	PC[2]	PI[4]	PE[15]	PH[11]	NC	NC	A
B	PC[9]	PB[2]	PH[13]	PC[12]	PE[6]	PH[5]	PC[4]	PH[9]	PH[10]	PI[2]	PC[3]	PG[11]	PG[15]	PG[14]	PA[11]	PA[10]	B
C	PC[14]	VDD_HV	PB[3]	PE[7]	PH[7]	PE[5]	PE[3]	VSS_LV	PC[1]	PI[3]	PA[5]	PI[5]	PE[14]	PE[12]	PA[9]	PA[8]	C
D	PH[14]	PI[6]	PC[15]	PI[7]	PH[6]	PE[4]	PE[2]	VDD_LV	VDD_HV	NC	PA[6]	PH[12]	PG[10]	PF[14]	PE[13]	PA[7]	D
E	PG[4]	PG[5]	PG[3]	PG[2]									PG[1]	PG[0]	PF[15]	VDD_HV	E
F	PE[0]	PA[2]	PA[1]	PE[1]									PH[0]	PH[1]	PH[3]	PH[2]	F
G	PE[9]	PE[8]	PE[10]	PA[0]			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV	PI[12]	PI[13]	MSEO	G
H	VSS_HV	PE[11]	VDD_HV	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			MDO3	MDO2	MDO0	MDO1	H
J	RESET	VSS_LV	NC	NC			VSS_HV	VSS_HV	VSS_HV	VSS_HV			PI[8]	PI[9]	PI[10]	PI[11]	J
K	EVTI	NC	VDD_BV	VDD_LV			VSS_HV	VSS_HV	VSS_HV	VSS_HV			VDD_HV_ADC1	PG[12]	PA[3]	PG[13]	K
L	PG[9]	PG[8]	NC	EVTO									PB[15]	PD[15]	PD[14]	PB[14]	L
M	PG[7]	PG[6]	PC[10]	PC[11]									PB[13]	PD[13]	PD[12]	PB[12]	M
N	PB[1]	PF[9]	PB[0]	VDD_HV	PJ[0]	PA[4]	VSS_LV	EXTAL	VDD_HV	PF[0]	PF[4]	VSS_HV_ADC1	PB[11]	PD[10]	PD[9]	PD[11]	N
P	PF[8]	PJ[3]	PC[7]	PJ[2]	PJ[1]	PA[14]	VDD_LV	XTAL	PB[10]	PF[1]	PF[5]	PD[0]	PD[3]	VDD_HV_ADC0	PB[6]	PB[7]	P
R	PF[12]	PC[6]	PF[10]	PF[11]	VDD_HV	PA[15]	PA[13]	PI[14]	XTAL32	PF[3]	PF[7]	PD[2]	PD[4]	PD[7]	VSS_HV_ADC0	PB[5]	R
T	NC	NC	NC	MCKO	NC	PF[13]	PA[12]	PI[15]	EXTAL32	PF[2]	PF[6]	PD[1]	PD[5]	PD[6]	PD[8]	PB[4]	T
	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	

NOTE: The 208 MAPBGA is available only as development package for Nexus 2+.

NC = Not connected

Figure 5. 208 MAPBGA configuration

3.2 Pin muxing

Table 3 defines the pin list and muxing for this device.

Each entry of Table 3 shows all the possible configurations for each pin, via the alternate functions. The default function assigned to each pin after reset is indicated by AF0.

Table 3. Functional port pin descriptions

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
Port A											
PA[0]	PCR[0]	AF0 AF1 AF2 AF3 —	GPIO[0] E0UC[0] CLKOUT E0UC[13] WKUP[19][4]	SIUL eMIOS_0 MC_CGM eMIOS_0 WKUP	I/O I/O O I/O I	M	Tristate	12	16	24	G4
PA[1]	PCR[1]	AF0 AF1 AF2 AF3 —	GPIO[1] E0UC[1] NMI[5] — WKUP[2][4]	SIUL eMIOS_0 WKUP — WKUP	I/O I/O I — I	S	Tristate	7	11	19	F3
PA[2]	PCR[2]	AF0 AF1 AF2 AF3 —	GPIO[2] E0UC[2] — MA[2] WKUP[3][4]	SIUL eMIOS_0 — ADC_0 WKUP	I/O I/O — O I	S	Tristate	5	9	17	F2
PA[3]	PCR[3]	AF0 AF1 AF2 AF3 — —	GPIO[3] E0UC[3] LIN5TX CS4_1 EIRQ[0] ADC1_S[0]	SIUL eMIOS_0 LINFlex_5 DSPI_1 SIUL ADC_1	I/O I/O O O I I	J	Tristate	68	90	114	K15
PA[4]	PCR[4]	AF0 AF1 AF2 AF3 — —	GPIO[4] E0UC[4] — CS0_1 LIN5RX WKUP[9][4]	SIUL eMIOS_0 — DSPI_1 LINFlex_5 WKUP	I/O I/O — I/O I I	S	Tristate	29	43	51	N6
PA[5]	PCR[5]	AF0 AF1 AF2 AF3	GPIO[5] E0UC[5] LIN4TX —	SIUL eMIOS_0 LINFlex_4 —	I/O I/O O —	M	Tristate	79	118	146	C11
PA[6]	PCR[6]	AF0 AF1 AF2 AF3 — —	GPIO[6] E0UC[6] — CS1_1 EIRQ[1] LIN4RX	SIUL eMIOS_0 — DSPI_1 SIUL LINFlex_4	I/O I/O — O I I	S	Tristate	80	119	147	D11

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PA[7]	PCR[7]	AF0 AF1 AF2 AF3 — —	GPIO[7] E0UC[7] LIN3TX — EIRQ[2] ADC1_S[1]	SIUL eMIOS_0 LINFlex_3 — SIUL ADC_1	I/O I/O O — I I	J	Tristate	71	104	128	D16
PA[8]	PCR[8]	AF0 AF1 AF2 AF3 — N/A[6] —	GPIO[8] E0UC[8] E0UC[14] — EIRQ[3] ABS[0] LIN3RX	SIUL eMIOS_0 eMIOS_0 — SIUL BAM LINFlex_3	I/O I/O I/O — I I I	S	Input, weak pull-up	72	105	129	C16
PA[9]	PCR[9]	AF0 AF1 AF2 AF3 N/A[6]	GPIO[9] E0UC[9] — CS2_1 FAB	SIUL eMIOS_0 — DSPI_1 BAM	I/O I/O — O I	S	Pull-down	73	106	130	C15
PA[10]	PCR[10]	AF0 AF1 AF2 AF3 —	GPIO[10] E0UC[10] SDA LIN2TX ADC1_S[2]	SIUL eMIOS_0 I^2C_0 LINFlex_2 ADC_1	I/O I/O I/O O I	J	Tristate	74	107	131	B16
PA[11]	PCR[11]	AF0 AF1 AF2 AF3 — — —	GPIO[11] E0UC[11] SCL — EIRQ[16] LIN2RX ADC1_S[3]	SIUL eMIOS_0 I^2C_0 — SIUL LINFlex_2 ADC_1	I/O I/O — — I I I	J	Tristate	75	108	132	B15
PA[12]	PCR[12]	AF0 AF1 AF2 AF3 — —	GPIO[12] — E0UC[28] CS3_1 EIRQ[17] SIN_0	SIUL — eMIOS_0 DSPI_1 SIUL DSPI_0	I/O — I/O O I I	S	Tristate	31	45	53	T7
PA[13]	PCR[13]	AF0 AF1 AF2 AF3	GPIO[13] SOUT_0 E0UC[29] —	SIUL DSPI_0 eMIOS_0 —	I/O O I/O —	M	Tristate	30	44	52	R7
PA[14]	PCR[14]	AF0 AF1 AF2 AF3 —	GPIO[14] SCK_0 CS0_0 E0UC[0] EIRQ[4]	SIUL DSPI_0 DSPI_0 eMIOS_0 SIUL	I/O I/O I/O I/O I	M	Tristate	28	42	50	P6

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PA[15]	PCR[15]	AF0 AF1 AF2 AF3 —	GPIO[15] CS0_0 SCK_0 E0UC[1] WKUP[10][4]	SIUL DSPI_0 DSPI_0 eMIOS_0 WKUP	I/O I/O I/O I/O I	M	Tristate	27	40	48	R6
Port B											
PB[0]	PCR[16]	AF0 AF1 AF2 AF3	GPIO[16] CAN0TX E0UC[30] LIN0TX	SIUL FlexCAN_0 eMIOS_0 LINFlex_0	I/O O I/O O	M	Tristate	23	31	39	N3
PB[1]	PCR[17]	AF0 AF1 AF2 AF3 — — —	GPIO[17] — E0UC[31] — WKUP[4][4] CAN0RX LIN0RX	SIUL — eMIOS_0 — WKUP FlexCAN_0 LINFlex_0	I/O — I/O — I I I	S	Tristate	24	32	40	N1
PB[2]	PCR[18]	AF0 AF1 AF2 AF3	GPIO[18] LIN0TX SDA E0UC[30]	SIUL LINFlex_0 I^2C_0 eMIOS_0	I/O O I/O I/O	M	Tristate	100	144	176	B2
PB[3]	PCR[19]	AF0 AF1 AF2 AF3 — —	GPIO[19] E0UC[31] SCL — WKUP[11][4] LIN0RX	SIUL eMIOS_0 I^2C_0 — WKUP LINFlex_0	I/O I/O I/O — I I	S	Tristate	1	1	1	C3
PB[4]	PCR[20]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[0] ADC1_P[0] GPIO[20]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	50	72	88	T16
PB[5]	PCR[21]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[1] ADC1_P[1] GPIO[21]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	53	75	91	R16

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PB[6]	PCR[22]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[2] ADC1_P[2] GPIO[22]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	54	76	92	P15
PB[7]	PCR[23]	AF0 AF1 AF2 AF3 — — —	— — — — ADC0_P[3] ADC1_P[3] GPIO[23]	— — — — ADC_0 ADC_1 SIUL	— — — — I I I	I	Tristate	55	77	93	P16
PB[8]	PCR[24]	AF0 AF1 AF2 AF3 — — — —	GPIO[24] — — — OSC32K_XTAL[7] WKUP[25] ADC0_S[0] ADC1_S[4]	SIUL — — — OSC32K WKUP ADC_0 ADC_1	I — — — — I I I	I	—	39	53	61	R9
PB[9]	PCR[25]	AF0 AF1 AF2 AF3 — — — —	GPIO[25] — — — OSC32K_EXTAL[7] WKUP[26] ADC0_S[1] ADC1_S[5]	SIUL — — — OSC32K WKUP ADC_0 ADC_1	I — — — — I I I	I	—	38	52	60	T9
PB[10]	PCR[26]	AF0 AF1 AF2 AF3 — — —	GPIO[26] — — — WKUP[8][4] ADC0_S[2] ADC1_S[6]	SIUL — — — WKUP ADC_0 ADC_1	I/O — — — I I I	J	Tristate	40	54	62	P9
PB[11]	PCR[27]	AF0 AF1 AF2 AF3 —	GPIO[27] E0UC[3] — CS0_0 ADC0_S[3]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — I/O I	J	Tristate	—	—	97	N13

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PB[12]	PCR[28]	AF0 AF1 AF2 AF3 —	GPIO[28] E0UC[4] — CS1_0 ADC0_X[0]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	61	83	101	M16
PB[13]	PCR[29]	AF0 AF1 AF2 AF3 —	GPIO[29] E0UC[5] — CS2_0 ADC0_X[1]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	63	85	103	M13
PB[14]	PCR[30]	AF0 AF1 AF2 AF3 —	GPIO[30] E0UC[6] — CS3_0 ADC0_X[2]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	65	87	105	L16
PB[15]	PCR[31]	AF0 AF1 AF2 AF3 —	GPIO[31] E0UC[7] — CS4_0 ADC0_X[3]	SIUL eMIOS_0 — DSPI_0 ADC_0	I/O I/O — O I	J	Tristate	67	89	107	L13
Port C											
PC[0][8]	PCR[32]	AF0 AF1 AF2 AF3	GPIO[32] — TDI —	SIUL — JTAGC —	I/O — I —	M	Input, weak pull-up	87	126	154	A8
PC[1][8]	PCR[33]	AF0 AF1 AF2 AF3	GPIO[33] — TDO —	SIUL — JTAGC —	I/O — O —	F[9]	Tristate	82	121	149	C9
PC[2]	PCR[34]	AF0 AF1 AF2 AF3 —	GPIO[34] SCK_1 CAN4TX DEBUG[0] EIRQ[5]	SIUL DSPI_1 FlexCAN_4 SSCM SIUL	I/O I/O O O I	M	Tristate	78	117	145	A11
PC[3]	PCR[35]	AF0 AF1 AF2 AF3 — — —	GPIO[35] CS0_1 MA[0] DEBUG[1] EIRQ[6] CAN1RX CAN4RX	SIUL DSPI_1 ADC_0 SSCM SIUL FlexCAN_1 FlexCAN_4	I/O I/O O O I I I	S	Tristate	77	116	144	B11

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PC[4]	PCR[36]	AF0 AF1 AF2 AF3 — — —	GPIO[36] E1UC[31] — DEBUG[2] EIRQ[18] SIN_1 CAN3RX	SIUL eMIOS_1 — SSCM SIUL DSPI_1 FlexCAN_3	I/O I/O — O I I I	M	Tristate	92	131	159	B7
PC[5]	PCR[37]	AF0 AF1 AF2 AF3 —	GPIO[37] SOUT_1 CAN3TX DEBUG[3] EIRQ[7]	SIUL DSPI_1 FlexCAN_3 SSCM SIUL	I/O O O O I	M	Tristate	91	130	158	A7
PC[6]	PCR[38]	AF0 AF1 AF2 AF3	GPIO[38] LIN1TX E1UC[28] DEBUG[4]	SIUL LINFlex_1 eMIOS_1 SSCM	I/O O I/O O	S	Tristate	25	36	44	R2
PC[7]	PCR[39]	AF0 AF1 AF2 AF3 — —	GPIO[39] — E1UC[29] DEBUG[5] LIN1RX WKUP[12][4]	SIUL — eMIOS_1 SSCM LINFlex_1 WKUP	I/O — I/O O I I	S	Tristate	26	37	45	P3
PC[8]	PCR[40]	AF0 AF1 AF2 AF3	GPIO[40] LIN2TX E0UC[3] DEBUG[6]	SIUL LINFlex_2 eMIOS_0 SSCM	I/O O I/O O	S	Tristate	99	143	175	A1
PC[9]	PCR[41]	AF0 AF1 AF2 AF3 — —	GPIO[41] — E0UC[7] DEBUG[7] WKUP[13][4] LIN2RX	SIUL — eMIOS_0 SSCM WKUP LINFlex_2	I/O — I/O O I I	S	Tristate	2	2	2	B1
PC[10]	PCR[42]	AF0 AF1 AF2 AF3	GPIO[42] CAN1TX CAN4TX MA[1]	SIUL FlexCAN_1 FlexCAN_4 ADC_0	I/O O O O	M	Tristate	22	28	36	M3
PC[11]	PCR[43]	AF0 AF1 AF2 AF3 — — —	GPIO[43] — — MA[2] WKUP[5][4] CAN1RX CAN4RX	SIUL — — ADC_0 WKUP FlexCAN_1 FlexCAN_4	I/O — — O I I I	S	Tristate	21	27	35	M4

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PC[12]	PCR[44]	AF0 AF1 AF2 AF3 — —	GPIO[44] E0UC[12] — — EIRQ[19] SIN_2	SIUL eMIOS_0 — — SIUL DSPI_2	I/O I/O — — I I	M	Tristate	97	141	173	B4
PC[13]	PCR[45]	AF0 AF1 AF2 AF3	GPIO[45] E0UC[13] SOUT_2 —	SIUL eMIOS_0 DSPI_2 —	I/O I/O O —	S	Tristate	98	142	174	A2
PC[14]	PCR[46]	AF0 AF1 AF2 AF3 —	GPIO[46] E0UC[14] SCK_2 — EIRQ[8]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	S	Tristate	3	3	3	C1
PC[15]	PCR[47]	AF0 AF1 AF2 AF3 —	GPIO[47] E0UC[15] CS0_2 — EIRQ[20]	SIUL eMIOS_0 DSPI_2 — SIUL	I/O I/O I/O — I	M	Tristate	4	4	4	D3
Port D											
PD[0]	PCR[48]	AF0 AF1 AF2 AF3 — — —	GPIO[48] — — — WKUP[27] ADC0_P[4] ADC1_P[4]	SIUL — — — WKUP ADC_0 ADC_1	I — — — I I I	I	Tristate	41	63	77	P12
PD[1]	PCR[49]	AF0 AF1 AF2 AF3 — — —	GPIO[49] — — — WKUP[28] ADC0_P[5] ADC1_P[5]	SIUL — — — WKUP ADC_0 ADC_1	I — — — I I I	I	Tristate	42	64	78	T12
PD[2]	PCR[50]	AF0 AF1 AF2 AF3 — —	GPIO[50] — — — ADC0_P[6] ADC1_P[6]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	43	65	79	R12

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PD[3]	PCR[51]	AF0 AF1 AF2 AF3 — —	GPIO[51] — — — ADC0_P[7] ADC1_P[7]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	44	66	80	P13
PD[4]	PCR[52]	AF0 AF1 AF2 AF3 — —	GPIO[52] — — — ADC0_P[8] ADC1_P[8]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	45	67	81	R13
PD[5]	PCR[53]	AF0 AF1 AF2 AF3 — —	GPIO[53] — — — ADC0_P[9] ADC1_P[9]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	46	68	82	T13
PD[6]	PCR[54]	AF0 AF1 AF2 AF3 — —	GPIO[54] — — — ADC0_P[10] ADC1_P[10]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	47	69	83	T14
PD[7]	PCR[55]	AF0 AF1 AF2 AF3 — —	GPIO[55] — — — ADC0_P[11] ADC1_P[11]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	48	70	84	R14
PD[8]	PCR[56]	AF0 AF1 AF2 AF3 — —	GPIO[56] — — — ADC0_P[12] ADC1_P[12]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	49	71	87	T15
PD[9]	PCR[57]	AF0 AF1 AF2 AF3 — —	GPIO[57] — — — ADC0_P[13] ADC1_P[13]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	56	78	94	N15

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PD[10]	PCR[58]	AF0 AF1 AF2 AF3 — —	GPIO[58] — — — ADC0_P[14] ADC1_P[14]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	57	79	95	N14
PD[11]	PCR[59]	AF0 AF1 AF2 AF3 — —	GPIO[59] — — — ADC0_P[15] ADC1_P[15]	SIUL — — — ADC_0 ADC_1	I — — — I I	I	Tristate	58	80	96	N16
PD[12]	PCR[60]	AF0 AF1 AF2 AF3 —	GPIO[60] CS5_0 E0UC[24] — ADC0_S[4]	SIUL DSPI_0 eMIOS_0 — ADC_0	I/O O I/O — I	J	Tristate	—	—	100	M15
PD[13]	PCR[61]	AF0 AF1 AF2 AF3 —	GPIO[61] CS0_1 E0UC[25] — ADC0_S[5]	SIUL DSPI_1 eMIOS_0 — ADC_0	I/O I/O — — I	J	Tristate	62	84	102	M14
PD[14]	PCR[62]	AF0 AF1 AF2 AF3 —	GPIO[62] CS1_1 E0UC[26] — ADC0_S[6]	SIUL DSPI_1 eMIOS_0 — ADC_0	I/O O I/O — I	J	Tristate	64	86	104	L15
PD[15]	PCR[63]	AF0 AF1 AF2 AF3 —	GPIO[63] CS2_1 E0UC[27] — ADC0_S[7]	SIUL DSPI_1 eMIOS_0 — ADC_0	I/O O I/O — I	J	Tristate	66	88	106	L14
Port E											
PE[0]	PCR[64]	AF0 AF1 AF2 AF3 — —	GPIO[64] E0UC[16] — — WKUP[6][4] CAN5RX	SIUL eMIOS_0 — — WKUP FlexCAN_5	I/O I/O — — I I	S	Tristate	6	10	18	F1
PE[1]	PCR[65]	AF0 AF1 AF2 AF3	GPIO[65] E0UC[17] CAN5TX —	SIUL eMIOS_0 FlexCAN_5 —	I/O I/O O —	M	Tristate	8	12	20	F4

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PE[2]	PCR[66]	AF0 AF1 AF2 AF3 — —	GPIO[66] E0UC[18] — — EIRQ[21] SIN_1	SIUL eMIOS_0 — — SIUL DSPI_1	I/O I/O — — I I	M	Tristate	89	128	156	D7
PE[3]	PCR[67]	AF0 AF1 AF2 AF3	GPIO[67] E0UC[19] SOUT_1 —	SIUL eMIOS_0 DSPI_1 —	I/O I/O O —	M	Tristate	90	129	157	C7
PE[4]	PCR[68]	AF0 AF1 AF2 AF3 —	GPIO[68] E0UC[20] SCK_1 — EIRQ[9]	SIUL eMIOS_0 DSPI_1 — SIUL	I/O I/O I/O — I	M	Tristate	93	132	160	D6
PE[5]	PCR[69]	AF0 AF1 AF2 AF3	GPIO[69] E0UC[21] CS0_1 MA[2]	SIUL eMIOS_0 DSPI_1 ADC_0	I/O I/O I/O O	M	Tristate	94	133	161	C6
PE[6]	PCR[70]	AF0 AF1 AF2 AF3 —	GPIO[70] E0UC[22] CS3_0 MA[1] EIRQ[22]	SIUL eMIOS_0 DSPI_0 ADC_0 SIUL	I/O I/O O O I	M	Tristate	95	139	167	B5
PE[7]	PCR[71]	AF0 AF1 AF2 AF3 —	GPIO[71] E0UC[23] CS2_0 MA[0] EIRQ[23]	SIUL eMIOS_0 DSPI_0 ADC_0 SIUL	I/O I/O O O I	M	Tristate	96	140	168	C4
PE[8]	PCR[72]	AF0 AF1 AF2 AF3	GPIO[72] CAN2TX E0UC[22] CAN3TX	SIUL FlexCAN_2 eMIOS_0 FlexCAN_3	I/O O I/O O	M	Tristate	9	13	21	G2
PE[9]	PCR[73]	AF0 AF1 AF2 AF3 — — —	GPIO[73] — E0UC[23] — WKUP[7][4] CAN2RX CAN3RX	SIUL — eMIOS_0 — WKUP FlexCAN_2 FlexCAN_3	I/O — I/O — I I I	S	Tristate	10	14	22	G1
PE[10]	PCR[74]	AF0 AF1 AF2 AF3 —	GPIO[74] LIN3TX CS3_1 E1UC[30] EIRQ[10]	SIUL LINFlex_3 DSPI_1 eMIOS_1 SIUL	I/O O O I/O I	S	Tristate	11	15	23	G3

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PE[11]	PCR[75]	AF0 AF1 AF2 AF3 — —	GPIO[75] E0UC[24] CS4_1 — LIN3RX WKUP[14][4]	SIUL eMIOS_0 DSPI_1 — LINFlex_3 WKUP	I/O I/O O — I I	S	Tristate	13	17	25	H2
PE[12]	PCR[76]	AF0 AF1 AF2 AF3 — — —	GPIO[76] — E1UC[19][10] — EIRQ[11] SIN_2 ADC1_S[7]	SIUL — eMIOS_1 — SIUL DSPI_2 ADC_1	I/O — I/O — I I I	J	Tristate	76	109	133	C14
PE[13]	PCR[77]	AF0 AF1 AF2 AF3	GPIO[77] SOUT_2 E1UC[20] —	SIUL DSPI_2 eMIOS_1 —	I/O O I/O —	S	Tristate	—	103	127	D15
PE[14]	PCR[78]	AF0 AF1 AF2 AF3 —	GPIO[78] SCK_2 E1UC[21] — EIRQ[12]	SIUL DSPI_2 eMIOS_1 — SIUL	I/O I/O I/O — I	S	Tristate	—	112	136	C13
PE[15]	PCR[79]	AF0 AF1 AF2 AF3	GPIO[79] CS0_2 E1UC[22] —	SIUL DSPI_2 eMIOS_1 —	I/O I/O I/O —	M	Tristate	—	113	137	A13
Port F											
PF[0]	PCR[80]	AF0 AF1 AF2 AF3 —	GPIO[80] E0UC[10] CS3_1 — ADC0_S[8]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	55	63	N10
PF[1]	PCR[81]	AF0 AF1 AF2 AF3 —	GPIO[81] E0UC[11] CS4_1 — ADC0_S[9]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	56	64	P10
PF[2]	PCR[82]	AF0 AF1 AF2 AF3 —	GPIO[82] E0UC[12] CS0_2 — ADC0_S[10]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	57	65	T10

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PF[3]	PCR[83]	AF0 AF1 AF2 AF3 —	GPIO[83] E0UC[13] CS1_2 — ADC0_S[11]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	58	66	R10
PF[4]	PCR[84]	AF0 AF1 AF2 AF3 —	GPIO[84] E0UC[14] CS2_2 — ADC0_S[12]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	59	67	N11
PF[5]	PCR[85]	AF0 AF1 AF2 AF3 —	GPIO[85] E0UC[22] CS3_2 — ADC0_S[13]	SIUL eMIOS_0 DSPI_2 — ADC_0	I/O I/O O — I	J	Tristate	—	60	68	P11
PF[6]	PCR[86]	AF0 AF1 AF2 AF3 —	GPIO[86] E0UC[23] CS1_1 — ADC0_S[14]	SIUL eMIOS_0 DSPI_1 — ADC_0	I/O I/O O — I	J	Tristate	—	61	69	T11
PF[7]	PCR[87]	AF0 AF1 AF2 AF3 —	GPIO[87] — CS2_1 — ADC0_S[15]	SIUL — DSPI_1 — ADC_0	I/O — O — I	J	Tristate	—	62	70	R11
PF[8]	PCR[88]	AF0 AF1 AF2 AF3	GPIO[88] CAN3TX CS4_0 CAN2TX	SIUL FlexCAN_3 DSPI_0 FlexCAN_2	I/O O O O	M	Tristate	—	34	42	P1
PF[9]	PCR[89]	AF0 AF1 AF2 AF3 — — —	GPIO[89] E1UC[1] CS5_0 — WKUP[22][4] CAN2RX CAN3RX	SIUL eMIOS_1 DSPI_0 — WKUP FlexCAN_2 FlexCAN_3	I/O I/O O — I I I	S	Tristate	—	33	41	N2
PF[10]	PCR[90]	AF0 AF1 AF2 AF3	GPIO[90] CS1_0 LIN4TX E1UC[2]	SIUL DSPI_0 LINFlex_4 eMIOS_1	I/O O O I/O	M	Tristate	—	38	46	R3

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PF[11]	PCR[91]	AF0 AF1 AF2 AF3 — —	GPIO[91] CS2_0 E1UC[3] — WKUP[15][4] LIN4RX	SIUL DSPI_0 eMIOS_1 — WKUP LINFlex_4	I/O O I/O — I I	S	Tristate	—	39	47	R4
PF[12]	PCR[92]	AF0 AF1 AF2 AF3	GPIO[92] E1UC[25] LIN5TX —	SIUL eMIOS_1 LINFlex_5 —	I/O I/O O —	M	Tristate	—	35	43	R1
PF[13]	PCR[93]	AF0 AF1 AF2 AF3 — —	GPIO[93] E1UC[26] — — WKUP[16][4] LIN5RX	SIUL eMIOS_1 — — WKUP LINFlex_5	I/O I/O — — I I	S	Tristate	—	41	49	T6
PF[14]	PCR[94]	AF0 AF1 AF2 AF3	GPIO[94] CAN4TX E1UC[27] CAN1TX	SIUL FlexCAN_4 eMIOS_1 FlexCAN_1	I/O O I/O O	M	Tristate	—	102	126	D14
PF[15]	PCR[95]	AF0 AF1 AF2 AF3 — — —	GPIO[95] E1UC[4] — — EIRQ[13] CAN1RX CAN4RX	SIUL eMIOS_1 — — SIUL FlexCAN_1 FlexCAN_4	I/O I/O — — I I I	S	Tristate	—	101	125	E15
Port G											
PG[0]	PCR[96]	AF0 AF1 AF2 AF3	GPIO[96] CAN5TX E1UC[23] —	SIUL FlexCAN_5 eMIOS_1 —	I/O O I/O —	M	Tristate	—	98	122	E14
PG[1]	PCR[97]	AF0 AF1 AF2 AF3 — —	GPIO[97] — E1UC[24] — EIRQ[14] CAN5RX	SIUL — eMIOS_1 — SIUL FlexCAN_5	I/O — I/O — I I	S	Tristate	—	97	121	E13
PG[2]	PCR[98]	AF0 AF1 AF2 AF3	GPIO[98] E1UC[11] SOUT_3 —	SIUL eMIOS_1 DSPI_3 —	I/O I/O O —	M	Tristate	—	8	16	E4

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PG[3]	PCR[99]	AF0 AF1 AF2 AF3 —	GPIO[99] E1UC[12] CS0_3 — WKUP[17][4]	SIUL eMIOS_1 DSPI_3 — WKUP	I/O I/O O — I	S	Tristate	—	7	15	E3
PG[4]	PCR[100]	AF0 AF1 AF2 AF3	GPIO[100] E1UC[13] SCK_3 —	SIUL eMIOS_1 DSPI_3 —	I/O I/O I/O —	M	Tristate	—	6	14	E1
PG[5]	PCR[101]	AF0 AF1 AF2 AF3 — —	GPIO[101] E1UC[14] — — WKUP[18][4] SIN_3	SIUL eMIOS_1 — — WKUP DSPI_3	I/O I/O — — I I	S	Tristate	—	5	13	E2
PG[6]	PCR[102]	AF0 AF1 AF2 AF3	GPIO[102] E1UC[15] LIN6TX —	SIUL eMIOS_1 LINFlex_6 —	I/O I/O O —	M	Tristate	—	30	38	M2
PG[7]	PCR[103]	AF0 AF1 AF2 AF3 — —	GPIO[103] E1UC[16] E1UC[30] — WKUP[20][4] LIN6RX	SIUL eMIOS_1 eMIOS_1 — WKUP LINFlex_6	I/O I/O I/O — I I	S	Tristate	—	29	37	M1
PG[8]	PCR[104]	AF0 AF1 AF2 AF3 —	GPIO[104] E1UC[17] LIN7TX CS0_2 EIRQ[15]	SIUL eMIOS_1 LINFlex_7 DSPI_2 SIUL	I/O I/O O I/O I	S	Tristate	—	26	34	L2
PG[9]	PCR[105]	AF0 AF1 AF2 AF3 — —	GPIO[105] E1UC[18] — SCK_2 WKUP[21][4] LIN7RX	SIUL eMIOS_1 — DSPI_2 WKUP LINFlex_7	I/O I/O — I/O I I	S	Tristate	—	25	33	L1
PG[10]	PCR[106]	AF0 AF1 AF2 AF3 —	GPIO[106] E0UC[24] E1UC[31] — SIN_4	SIUL eMIOS_0 eMIOS_1 — DSPI_4	I/O I/O I/O — I	S	Tristate	—	114	138	D13
PG[11]	PCR[107]	AF0 AF1 AF2 AF3	GPIO[107] E0UC[25] CS0_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O O —	M	Tristate	—	115	139	B12

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PG[12]	PCR[108]	AF0 AF1 AF2 AF3	GPIO[108] E0UC[26] SOUT_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O O —	M	Tristate	—	92	116	K14
PG[13]	PCR[109]	AF0 AF1 AF2 AF3	GPIO[109] E0UC[27] SCK_4 —	SIUL eMIOS_0 DSPI_4 —	I/O I/O I/O —	M	Tristate	—	91	115	K16
PG[14]	PCR[110]	AF0 AF1 AF2 AF3	GPIO[110] E1UC[0] LIN8TX —	SIUL eMIOS_1 LINFlex_8 —	I/O I/O O —	S	Tristate	—	110	134	B14
PG[15]	PCR[111]	AF0 AF1 AF2 AF3 —	GPIO[111] E1UC[1] — — LIN8RX	SIUL eMIOS_1 — — LINFlex_8	I/O I/O — — I	M	Tristate	—	111	135	B13
Port H											
PH[0]	PCR[112]	AF0 AF1 AF2 AF3 —	GPIO[112] E1UC[2] — — SIN_1	SIUL eMIOS_1 — — DSPI_1	I/O I/O — — I	M	Tristate	—	93	117	F13
PH[1]	PCR[113]	AF0 AF1 AF2 AF3	GPIO[113] E1UC[3] SOUT_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O O —	M	Tristate	—	94	118	F14
PH[2]	PCR[114]	AF0 AF1 AF2 AF3	GPIO[114] E1UC[4] SCK_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O I/O —	M	Tristate	—	95	119	F16
PH[3]	PCR[115]	AF0 AF1 AF2 AF3	GPIO[115] E1UC[5] CS0_1 —	SIUL eMIOS_1 DSPI_1 —	I/O I/O I/O —	M	Tristate	—	96	120	F15
PH[4]	PCR[116]	AF0 AF1 AF2 AF3	GPIO[116] E1UC[6] — —	SIUL eMIOS_1 — —	I/O I/O — —	M	Tristate	—	134	162	A6
PH[5]	PCR[117]	AF0 AF1 AF2 AF3	GPIO[117] E1UC[7] — —	SIUL eMIOS_1 — —	I/O I/O — —	S	Tristate	—	135	163	B6

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PH[6]	PCR[118]	AF0 AF1 AF2 AF3	GPIO[118] E1UC[8] — MA[2]	SIUL eMIOS_1 — ADC_0	I/O I/O — O	M	Tristate	—	136	164	D5
PH[7]	PCR[119]	AF0 AF1 AF2 AF3	GPIO[119] E1UC[9] CS3_2 MA[1]	SIUL eMIOS_1 DSPI_2 ADC_0	I/O I/O O O	M	Tristate	—	137	165	C5
PH[8]	PCR[120]	AF0 AF1 AF2 AF3	GPIO[120] E1UC[10] CS2_2 MA[0]	SIUL eMIOS_1 DSPI_2 ADC_0	I/O I/O O O	M	Tristate	—	138	166	A5
PH[9][8]	PCR[121]	AF0 AF1 AF2 AF3	GPIO[121] — TCK —	SIUL — JTAGC —	I/O — I —	S	Input, weak pull-up	88	127	155	B8
PH[10][8]	PCR[122]	AF0 AF1 AF2 AF3	GPIO[122] — TMS —	SIUL — JTAGC —	I/O — I —	M	Input, weak pull-up	81	120	148	B9
PH[11]	PCR[123]	AF0 AF1 AF2 AF3	GPIO[123] SOUT_3 CS0_4 E1UC[5]	SIUL DSPI_3 DSPI_4 eMIOS_1	I/O O I/O —	M	Tristate	—	—	140	A14
PH[12]	PCR[124]	AF0 AF1 AF2 AF3	GPIO[124] SCK_3 CS1_4 E1UC[25]	SIUL DSPI_3 DSPI_4 eMIOS_1	I/O I/O I/O —	M	Tristate	—	—	141	D12
PH[13]	PCR[125]	AF0 AF1 AF2 AF3	GPIO[125] SOUT_4 CS0_3 E1UC[26]	SIUL DSPI_4 DSPI_3 eMIOS_1	I/O O I/O —	M	Tristate	—	—	9	B3
PH[14]	PCR[126]	AF0 AF1 AF2 AF3	GPIO[126] SCK_4 CS1_3 E1UC[27]	SIUL DSPI_4 DSPI_3 eMIOS_1	I/O I/O I/O —	M	Tristate	—	—	10	D1
PH[15]	PCR[127]	AF0 AF1 AF2 AF3	GPIO[127] SOUT_5 — E1UC[17]	SIUL DSPI_5 — eMIOS_1	I/O O — —	M	Tristate	—	—	8	A3
Port I											

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PI[0]	PCR[128]	AF0 AF1 AF2 AF3	GPIO[128] E0UC[28] LIN8TX —	SIUL eMIOS_0 LINFlex_8 —	I/O I/O O —	S	Tristate	—	—	172	A9
PI[1]	PCR[129]	AF0 AF1 AF2 AF3 — —	GPIO[129] E0UC[29] — — WKUP[24][4] LIN8RX	SIUL eMIOS_0 — — WKUP LINFlex_8	I/O I/O — — I I	S	Tristate	—	—	171	A10
PI[2]	PCR[130]	AF0 AF1 AF2 AF3	GPIO[130] E0UC[30] LIN9TX —	SIUL eMIOS_0 LINFlex_9 —	I/O I/O O —	S	Tristate	—	—	170	B10
PI[3]	PCR[131]	AF0 AF1 AF2 AF3 — —	GPIO[131] E0UC[31] — — WKUP[23][4] LIN9RX	SIUL eMIOS_0 — — WKUP LINFlex_9	I/O I/O — — I I	S	Tristate	—	—	169	C10
PI[4]	PCR[132]	AF0 AF1 AF2 AF3	GPIO[132] E1UC[28] SOUT_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O O —	S	Tristate	—	—	143	A12
PI[5]	PCR[133]	AF0 AF1 AF2 AF3	GPIO[133] E1UC[29] SCK_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	—	—	142	C12
PI[6]	PCR[134]	AF0 AF1 AF2 AF3	GPIO[134] E1UC[30] CS0_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	—	—	11	D2
PI[7]	PCR[135]	AF0 AF1 AF2 AF3	GPIO[135] E1UC[31] CS1_4 —	SIUL eMIOS_1 DSPI_4 —	I/O I/O I/O —	S	Tristate	—	—	12	D3
PI[8]	PCR[136]	AF0 AF1 AF2 AF3 —	GPIO[136] — — — ADC0_S[16]	SIUL — — — ADC_0	I/O — — — I	J	Tristate	—	—	108	J13

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PI[9]	PCR[137]	AF0 AF1 AF2 AF3 —	GPIO[137] — — — ADC0_S[17]	SIUL — — — ADC_0	I/O — — — I	J	Tristate	—	—	109	J14
PI[10]	PCR[138]	AF0 AF1 AF2 AF3 —	GPIO[138] — — — ADC0_S[18]	SIUL — — — ADC_0	I/O — — — I	J	Tristate	—	—	110	J15
PI[11]	PCR[139]	AF0 AF1 AF2 AF3 — —	GPIO[139] — — — ADC0_S[19] SIN_3	SIUL — — — ADC_0 DSPI_3	I/O — — — I I	J	Tristate	—	—	111	J16
PI[12]	PCR[140]	AF0 AF1 AF2 AF3 —	GPIO[140] CS0_3 — — ADC0_S[20]	SIUL DSPI_3 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	112	G14
PI[13]	PCR[141]	AF0 AF1 AF2 AF3 —	GPIO[141] CS1_3 — — ADC0_S[21]	SIUL DSPI_3 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	113	G15
PI[14]	PCR[142]	AF0 AF1 AF2 AF3 — —	GPIO[142] — — — ADC0_S[22] SIN_4	SIUL — — — ADC_0 DSPI_4	I/O — — — I I	J	Tristate	—	—	76	R8
PI[15]	PCR[143]	AF0 AF1 AF2 AF3 —	GPIO[143] CS0_4 — — ADC0_S[23]	SIUL DSPI_4 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	75	T8
Port J											
PJ[0]	PCR[144]	AF0 AF1 AF2 AF3 —	GPIO[144] CS1_4 — — ADC0_S[24]	SIUL DSPI_4 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	74	N5

Table 3. Functional port pin descriptions (continued)

Port pin	PCR register	Alternate function[1]	Function	Peripheral	I/O direction	Pad type	RESET config.[2]	Pin No.			
								LQFP 100 LQFP	LQFP 144 LQFP	LQFP 176 LQFP	208 MAP BGA[3]
PJ[1]	PCR[145]	AF0 AF1 AF2 AF3 — —	GPIO[145] — — — ADC0_S[25] SIN_5	SIUL — — —— ADC_0 DSPI_5	I/O — — — I I	J	Tristate	—	—	73	P5
PJ[2]	PCR[146]	AF0 AF1 AF2 AF3 —	GPIO[146] CS0_5 — — ADC0_S[26]	SIUL DSPI_5 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	72	P4
PJ[3]	PCR[147]	AF0 AF1 AF2 AF3 —	GPIO[147] CS1_5 — — ADC0_S[27]	SIUL DSPI_5 — — ADC_0	I/O I/O — — I	J	Tristate	—	—	71	P2
PJ[4]	PCR[148]	AF0 AF1 AF2 AF3	GPIO[148] SCK_5 E1UC[18] —	SIUL DSPI_5 eMIOS_1 —	I/O I/O — —	M	Tristate	—	—	5	A4

1 Alternate functions are chosen by setting the values of the PCR.PA bitfields inside the SIUL module. PCR.PA = 00 → AF0; PCR.PA = 01 → AF1; PCR.PA = 10 → AF2; PCR.PA = 11→ AF3. This is intended to select the output functions; to use one of the input functions, the PCR.IBE bit must be written to '1', regardless of the values selected in the PCR.PA bitfields. For this reason, the value corresponding to an input only function is reported as "—".

2 The RESET configuration applies during and after reset.

3 208 MAPBGA available only as development package for Nexus2+

4 All WKUP pins also support external interrupt capability. See the WKPU chapter of the MPC5607B Microcontroller Reference Manual for further details.

5 NMI has higher priority than alternate function. When NMI is selected, the PCR.AF field is ignored.

6 "Not applicable" because these functions are available only while the device is booting. See the BAM chapter of the MPC5607B Microcontroller Reference Manual for details.

7 Value of PCR.IBE bit must be 0

8 Out of reset all the functional pins except PC[0:1] and PH[9:10] are available to the user as GPIO.
PC[0:1] are available as JTAG pins (TDI and TDO respectively).
PH[9:10] are available as JTAG pins (TCK and TMS respectively).
It is up to the user to configure these pins as GPIO when needed.

9 PC[1] is a fast/medium pad but is in medium configuration by default. This pad is in Alternate Function 2 mode after reset which has TDO functionality. The reset value of PCR.OBE is '1', but this setting has no impact as long as this pad stays in AF2 mode. After configuring this pad as GPIO (PCR.PA = 0), output buffer is enabled as reset value of PCR.OBE = 1.

10 Not available in 100 LQFP package

4 Electrical characteristics

This section contains electrical characteristics of the device as well as temperature and power considerations.

This product contains devices to protect the inputs against damage due to high static voltages. However, it is advisable to take precautions to avoid application of any voltage higher than the specified maximum rated voltages.

To enhance reliability, unused inputs can be driven to an appropriate logic voltage level (V_{DD} or V_{SS}). This could be done by the internal pull-up and pull-down, which is provided by the product for most general purpose pins.

The parameters listed in the following tables represent the characteristics of the device and its demands on the system.

In the tables where the device logic provides signals with their respective timing characteristics, the symbol “CC” for Controller Characteristics is included in the Symbol column.

In the tables where the external system must provide signals with their respective timing characteristics to the device, the symbol “SR” for System Requirement is included in the Symbol column.

4.1 Parameter classification

The electrical parameters shown in this supplement are guaranteed by various methods. To give the customer a better understanding, the classifications listed in Table 4 are used and the parameters are tagged accordingly in the tables where appropriate.

Table 4. Parameter classifications

Classification tag	Tag description
P	Those parameters are guaranteed during production testing on each individual device.
C	Those parameters are achieved by the design characterization by measuring a statistically relevant sample size across process variations.
T	Those parameters are achieved by design characterization on a small sample size from typical devices under typical conditions unless otherwise noted. All values shown in the typical column are within this category.
D	Those parameters are derived mainly from simulations.

NOTE

The classification is shown in the column labeled “C” in the parameter tables where appropriate.

4.2 NVUSRO register

Portions of the device configuration, such as high voltage supply, oscillator margin, and watchdog enable/disable after reset are controlled via bit values in the Non-Volatile User Options Register (NVUSRO) register.

For a detailed description of the NVUSRO register, please refer to the MPC5607B Microcontroller Reference Manual.

4.2.1 NVUSRO[PAD3V5V] field description

Table 5 shows how NVUSRO[PAD3V5V] controls the device configuration.

Table 5. PAD3V5V field description[1]

Value[2]	Description
0	High voltage supply is 5.0 V
1	High voltage supply is 3.3 V

1 See the device reference manual for more information on the NVUSRO register.

2 '1' is delivery value. It is part of shadow Flash, thus programmable by customer.

The DC electrical characteristics are dependent on the PAD3V5V bit value.

4.2.2 NVUSRO[OSCILLATOR_MARGIN] field description

Table 6 shows how NVUSRO[OSCILLATOR_MARGIN] controls the device configuration.

Table 6. OSCILLATOR_MARGIN field description[1]

Value[2]	Description
0	Low consumption configuration (4 MHz/8 MHz)
1	High margin configuration (4 MHz/16 MHz)

1 See the device reference manual for more information on the NVUSRO register.

2 '1' is delivery value. It is part of shadow Flash, thus programmable by customer.

The fast external crystal oscillator consumption is dependent on the OSCILLATOR_MARGIN bit value.

4.3 Absolute maximum ratings

Table 7. Absolute maximum ratings

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V_{DD}	SR	Voltage on VDD_HV pins with respect to ground (V_{SS})	—	–0.3	6.0	V
V_{SS_LV}	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_BV}	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V_{SS})	—	–0.3	6.0	V
			Relative to V_{DD}	–0.3	V_{DD} + 0.3	
V_{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pin with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) with respect to ground (V_{SS})	—	–0.3	6.0	V
			Relative to V_{DD}	V_{DD} – 0.3	V_{DD} + 0.3	
V_{IN}	SR	Voltage on any GPIO pin with respect to ground (V_{SS})	—	–0.3	6.0	V
			Relative to V_{DD}	—	V_{DD} + 0.3	

Table 7. Absolute maximum ratings (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
I_{INJPAD}	SR	Injected input current on any pin during overload condition	—	–10	10	mA
I_{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	–50	50	
I_{AVGSEG}	SR	Sum of all the static I/O current within a supply segment	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	70	mA
			V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	64	
$T_{STORAGE}$	SR	Storage temperature	—	–55	150	°C

NOTE

Stresses exceeding the recommended absolute maximum ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During overload conditions ($V_{IN} > V_{DD}$ or $V_{IN} < V_{SS}$), the voltage on pins with respect to ground (V_{SS}) must not exceed the recommended values.

4.4 Recommended operating conditions

Table 8. Recommended operating conditions (3.3 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V_{DD}[1]	SR	Voltage on VDD_HV pins with respect to ground (V_{SS})	—	3.0	3.6	V
V_{SS_LV}[2]	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_BV}[3]	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V_{SS})	—	3.0	3.6	V
			Relative to V_{DD}	V_{DD} – 0.1	V_{DD} + 0.1	
V_{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pin with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_ADC}[4]	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) with respect to ground (V_{SS})	—	3.0[5]	3.6	V
			Relative to V_{DD}	V_{DD} – 0.1	V_{DD} + 0.1	
V_{IN}	SR	Voltage on any GPIO pin with respect to ground (V_{SS})	—	V_{SS} – 0.1	—	V
			Relative to V_{DD}	—	V_{DD} + 0.1	

Table 8. Recommended operating conditions (3.3 V) (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
I_{INJPAD}	SR	Injected input current on any pin during overload condition	—	–5	5	mA
I_{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	–50	50	
TV_{DD}	SR	V_{DD} slope to ensure correct power up[6]	—	—	0.25	V/µs
T_{A} C-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	85	°C
T_{J} C-Grade Part	SR	Junction temperature under bias	—	–40	110	
T_{A} V-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	105	
T_{J} V-Grade Part	SR	Junction temperature under bias	—	–40	130	
T_{A} M-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	125	
T_{J} M-Grade Part	SR	Junction temperature under bias	—	–40	150	

1 100 nF capacitance needs to be provided between each V_{DD}/V_{SS} pair.

2 330 nF capacitance needs to be provided between each V_{DD_LV}/V_{SS_LV} supply pair.

3 470 nF capacitance needs to be provided between V_{DD_BV} and the nearest V_{SS_LV} (higher value may be needed depending on external regulator characteristics). Supply ramp slope on VDD_BV should always be faster or equal to slope of VDD_HV. Otherwise, device may enter regulator bypass mode if slope on VDD_BV is slower.

4 100 nF capacitance needs to be provided between V_{DD_ADC}/V_{SS_ADC} pair.

5 Full electrical specification cannot be guaranteed when voltage drops below 3.0 V. In particular, ADC electrical characteristics and I/Os DC electrical specification may not be guaranteed. When voltage drops below V_{LVDHVL}, device is reset.

6 Guaranteed by device validation

7 This frequency includes the 4% frequency modulation guardband.

Table 9. Recommended operating conditions (5.0 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{SS}	SR	Digital ground on VSS_HV pins	—	0	0	V
V_{DD}[1]	SR	Voltage on VDD_HV pins with respect to ground (V_{SS})	—	4.5	5.5	V
			Voltage drop[2]	3.0	5.5	
V_{SS_LV}[3]	SR	Voltage on VSS_LV (low voltage digital supply) pins with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V

Table 9. Recommended operating conditions (5.0 V) (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max	
V_{DD_BV}[4]	SR	Voltage on VDD_BV pin (regulator supply) with respect to ground (V_{SS})	—	4.5	5.5	V
			Voltage drop[2]	3.0	5.5	
			Relative to V_{DD}	3.0	V_{DD} + 0.1	
V_{SS_ADC}	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) pin with respect to ground (V_{SS})	—	V_{SS} – 0.1	V_{SS} + 0.1	V
V_{DD_ADC}[5]	SR	Voltage on VSS_HV_ADC0, VSS_HV_ADC1 (ADC reference) with respect to ground (V_{SS})	—	4.5	5.5	V
			Voltage drop[2]	3.0	5.5	
			Relative to V_{DD}	V_{DD} – 0.1	V_{DD} + 0.1	
V_{IN}	SR	Voltage on any GPIO pin with respect to ground (V_{SS})	—	V_{SS} – 0.1	—	V
			Relative to V_{DD}	—	V_{DD} + 0.1	
I_{INJPAD}	SR	Injected input current on any pin during overload condition	—	–5	5	mA
I_{INJSUM}	SR	Absolute sum of all injected input currents during overload condition	—	–50	50	
TV_{DD}	SR	V_{DD} slope to ensure correct power up[6]	—	—	0.25	V/µs
T_{A} C-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	85	°C
T_{J} C-Grade Part	SR	Junction temperature under bias	—	–40	110	
T_{A} V-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	105	
T_{J} V-Grade Part	SR	Junction temperature under bias	—	–40	130	
T_{A} M-Grade Part	SR	Ambient temperature under bias	f_{CPU} < 64 MHz[7]	–40	125	
T_{J} M-Grade Part	SR	Junction temperature under bias	—	–40	150	

1 100 nF capacitance needs to be provided between each V_{DD}/V_{SS} pair.

2 Full device operation is guaranteed by design when the voltage drops below 4.5V down to 3.6V. However, certain analog electrical characteristics will not be guaranteed to stay within the stated limits.

3 330 nF capacitance needs to be provided between each V_{DD_LV}/V_{SS_LV} supply pair.

4 470 nF capacitance needs to be provided between V_{DD_BV} and the nearest V_{SS_LV} (higher value may be needed depending on external regulator characteristics). While the supply voltage ramps up, the slope on V_{DD_BV} should be less than $0.9V_{DD\_HV}$ in order to ensure the device does not enter regulator bypass mode.

5 100 nF capacitance needs to be provided between V_{DD_ADC}/V_{SS_ADC} pair.

6 Guaranteed by device validation. Please refer to Section 4.5.1, "External ballast resistor recommendations for minimum V_{DD} slope to be guaranteed to ensure correct power up in case of external resistor usage.

7 This frequency includes the 4% frequency modulation guardband.

NOTE

RAM data retention is guaranteed with V_{DD_LV} not below 1.08 V.

4.5 Thermal characteristics

4.5.1 External ballast resistor recommendations

External ballast resistor on V_{DD_BV} pin helps in reducing the overall power dissipation inside the device. This resistor is required only when maximum power consumption exceeds the limit imposed by package thermal characteristics.

As stated in Table 10 LQFP thermal characteristics, considering a thermal resistance of 144 LQFP as 48.3 °C/W, at ambient temperature T_A = 125 °C, the junction temperature T_j will cross 150 °C if the total power dissipation is greater than (150 – 125)/48.3 = 517 mW. Therefore, the total device current I_{DDMAX} at 125 °C/5.5 V must not exceed 94.1 mA (i.e., PD/VDD). Assuming an average $I_{DD}(V_{DD\_HV})$ of 15–20 mA consumption typically during device RUN mode, the LV domain consumption $I_{DD}(V_{DD\_BV})$ is thus limited to $I_{DDMAX} - I_{DD}(V_{DD\_HV})$, i.e., 80 mA.

Therefore, respecting the maximum power allowed as explained in Section 4.5.2, “Package thermal characteristics, it is recommended to use this resistor only in the 125 °C/5.5 V operating corner as per the following guidelines:

- If $I_{DD}(V_{DD\_BV})$ < 80 mA, then no resistor is required.
- If 80 mA < $I_{DD}(V_{DD\_BV})$ < 90 mA, then 4 Ω resistor can be used.
- If $I_{DD}(V_{DD\_BV})$ > 90 mA, then 8 Ω resistor can be used.

Using resistance in the range of 4–8 Ω, the gain will be around 10–20% of total consumption on V_{DD_BV}. For example, if 8 Ω resistor is used, then power consumption when $I_{DD}(V_{DD\_BV})$ is 110 mA is equivalent to power consumption when $I_{DD}(V_{DD\_BV})$ is 90 mA (approximately) when resistor not used.

In order to ensure correct power up, the minimum V_{DD_BV} to be guaranteed is 30 ms/V. If the supply ramp is slower than this value, then LVDHV3B monitoring ballast supply V_{DD_BV} pin gets triggered leading to device reset. Until the supply reaches certain threshold, this low voltage monitor generates destructive reset event in the system. This threshold depends on the maximum $I_{DD}(V_{DD\_BV})$ possible across the external resistor.

4.5.2 Package thermal characteristics

Table 10. LQFP thermal characteristics[1]

Symbol		C	Parameter	Conditions[2]	Pin count	Value			Unit
						Min	Typ	Max	
$R_{\theta JA}$	CC	D	Thermal resistance, junction-to-ambient natural convection[3]	Single-layer board — 1s	100	—	—	64	°C/W
					144	—	—	64	
					176	—	—	64	
				Four-layer board — 2s2p	100	—	—	49.7	
					144	—	—	48.3	
					176	—	—	47.3	

Table 10. LQFP thermal characteristics[1] (continued)

Symbol		C	Parameter	Conditions[2]	Pin count	Value			Unit
						Min	Typ	Max	
$R_{\theta JB}$	CC		Thermal resistance, junction-to-board[4]	Single-layer board — 1s	100	—	—	36	°C/W
					144	—	—	38	
					176	—	—	38	
				Four-layer board — 2s2p	100	—	—	33.6	
					144	—	—	33.4	
					176	—	—	33.4	
$R_{\theta JC}$	CC		Thermal resistance, junction-to-case[5]	Single-layer board — 1s	100	—	—	23	°C/W
					144	—	—	23	
					176	—	—	23	
				Four-layer board — 2s2p	100	—	—	19.8	
					144	—	—	19.2	
					176	—	—	18.8	

1 Thermal characteristics are targets based on simulation.

2 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C.

3 Junction-to-ambient thermal resistance determined per JEDEC JESD51-3 and JESD51-6. Thermal test board meets JEDEC specification for this package. When Greek letters are not available, the symbols are typed as R_{thJA} and R_{thJMA}.

4 Junction-to-board thermal resistance determined per JEDEC JESD51-8. Thermal test board meets JEDEC specification for the specified package. When Greek letters are not available, the symbols are typed as R_{thJB}.

5 Junction-to-case at the top of the package determined using MIL-STD 883 Method 1012.1. The cold plate temperature is used for the case temperature. Reported value includes the thermal resistance of the interface layer. When Greek letters are not available, the symbols are typed as R_{thJC}.

4.5.3 Power considerations

The average chip-junction temperature, T_J, in degrees Celsius, may be calculated using Equation 1:

$$T_J = T_A + (P_D \times R_{\theta JA}) \qquad \textit{Eqn. 1}$$

Where:

T_A is the ambient temperature in °C.

$R_{\theta JA}$ is the package junction-to-ambient thermal resistance, in °C/W.

P_D is the sum of P_{INT} and $P_{I/O}$ ($P_D = P_{INT} + P_{I/O}$).

P_{INT} is the product of I_{DD} and V_{DD}, expressed in watts. This is the chip internal power.

$P_{I/O}$ represents the power dissipation on input and output pins; user determined.

Most of the time for the applications, $P_{I/O} < P_{INT}$ and may be neglected. On the other hand, $P_{I/O}$ may be significant, if the device is configured to continuously drive external modules and/or memories.

An approximate relationship between P_D and T_J (if $P_{I/O}$ is neglected) is given by:

$$P_D = K / (T_J + 273\ °C) \quad \textit{Eqn. 2}$$

Therefore, solving equations 1 and 2:

$$K = P_D \times (T_A + 273\ °C) + R_{\theta JA} \times P_D^2 \quad \textit{Eqn. 3}$$

Where:

K is a constant for the particular part, which may be determined from Equation 3 by measuring P_D (at equilibrium) for a known T_A. Using this value of K, the values of P_D and T_J may be obtained by solving equations 1 and 2 iteratively for any value of T_A.

4.6 I/O pad electrical characteristics

4.6.1 I/O pad types

The device provides four main I/O pad types depending on the associated alternate functions:

- Slow pads - are the most common pads, providing a good compromise between transition time and low electromagnetic emission.
- Medium pads - provide transition fast enough for the serial communication channels with controlled current to reduce electromagnetic emission.
- Fast pads - provide maximum speed. These are used for improved Nexus debugging capability.
- Input only pads - are associated with ADC channels and 32 kHz low power external crystal oscillator providing low input leakage.

Medium and Fast pads can use slow configuration to reduce electromagnetic emission, at the cost of reducing AC performance.

4.6.2 I/O input DC characteristics

Table 11 provides input DC electrical characteristics as described in Figure 6.

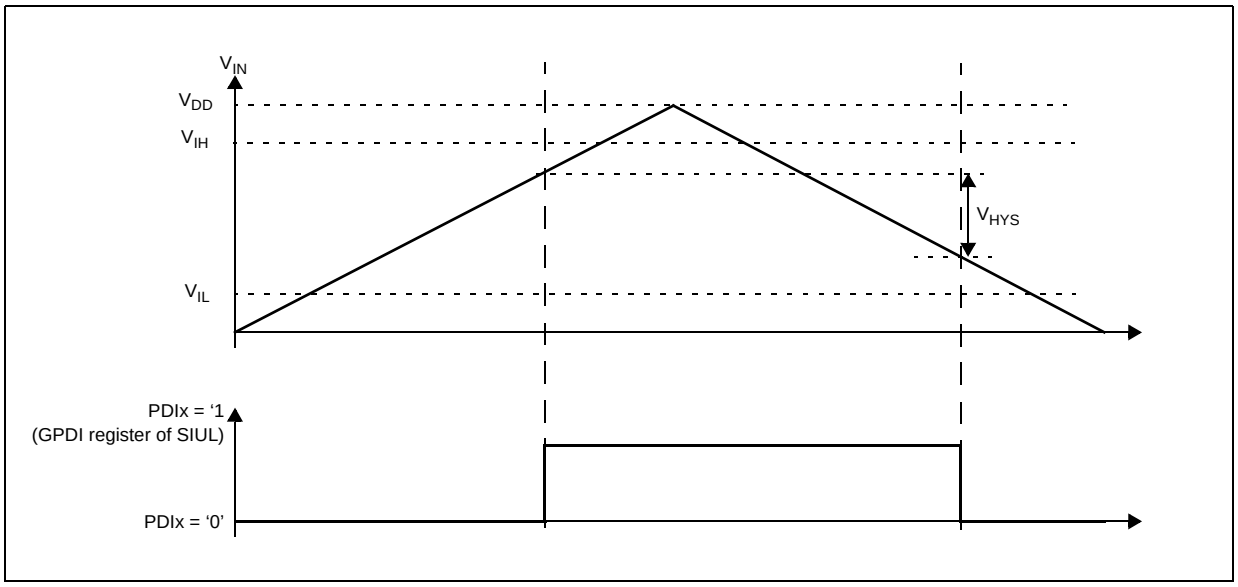


Figure 6. I/O input DC electrical characteristics definition

Table 11. I/O input DC electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{IH}	SR	P	Input high level CMOS (Schmitt Trigger)	—		$0.65V_{DD}$	—	V_{DD} + 0.4	V
V_{IL}	SR	P	Input low level CMOS (Schmitt Trigger)	—		–0.4	—	$0.35V_{DD}$	
V_{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—		$0.1V_{DD}$	—	—	
I_{LKG}	CC	P	Digital input leakage	No injection on adjacent pin	T_A = –40 °C	—	2	—	nA
		P			T_A = 25 °C	—	2	—	
		D			T_A = 105 °C	—	12	500	
		P			T_A = 125 °C	—	70	1000	
W_{FI}[2]	SR	P	Wakeup input filtered pulse	—		—	—	40	ns
W_{NFI}[2]	SR	P	Wakeup input not filtered pulse	—		1000	—	—	ns

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 In the range from 40 to 1000 ns, pulses can be filtered or not filtered, according to operating temperature and voltage.

4.6.3 I/O output DC characteristics

The following tables provide DC characteristics for bidirectional pads:

- Table 12 provides weak pull figures. Both pull-up and pull-down resistances are supported.
- Table 13 provides output driver characteristics for I/O pads when in SLOW configuration.
- Table 14 provides output driver characteristics for I/O pads when in MEDIUM configuration.
- Table 15 provides output driver characteristics for I/O pads when in FAST configuration.

Table 12. I/O pull-up/pull-down DC electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
$\lvert I_{WPU}\rvert$	CC	P	Weak pull-up current absolute value	$V_{IN} = V_{IL}$, V_{DD} = 5.0 V ± 10%	PAD3V5V = 0	10	—	150	µA
		C			PAD3V5V = 1[2]	10	—	250	
		P		$V_{IN} = V_{IL}$, V_{DD} = 3.3 V ± 10%	PAD3V5V = 1	10	—	150	
$\lvert I_{WPD}\rvert$	CC	P	Weak pull-down current absolute value	$V_{IN} = V_{IH}$, V_{DD} = 5.0 V ± 10%	PAD3V5V = 0	10	—	150	µA
		C			PAD3V5V = 1	10	—	250	
		P		$V_{IN} = V_{IH}$, V_{DD} = 3.3 V ± 10%	PAD3V5V = 1	10	—	150	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.

2 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 13. SLOW configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OH}	CC	P	Output high level SLOW configuration	Push Pull	I_{OH} = –2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	$0.8V_{DD}$	—	—	V
		C			I_{OH} = –2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	$0.8V_{DD}$	—	—	
		C			I_{OH} = –1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V_{DD} – 0.8	—	—	
V_{OL}	CC	P	Output low level SLOW configuration	Push Pull	I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
		C			I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 14. MEDIUM configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OH}	CC	C	Output high level MEDIUM configuration	Push Pull	I_{OH} = –3.8 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	$0.8V_{DD}$	—	—	V
		P			I_{OH} = –2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	$0.8V_{DD}$	—	—	
		C			I_{OH} = –1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	$0.8V_{DD}$	—	—	
		C			I_{OH} = –1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V_{DD} – 0.8	—	—	
		C			I_{OH} = –100 µA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	$0.8V_{DD}$	—	—	

Table 14. MEDIUM configuration output buffer electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OL}	CC	C	Output low level MEDIUM configuration	Push Pull	I_{OL} = 3.8 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	$0.2V_{DD}$	V
		P			I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
		C			I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	
		C			I_{OH} = 100 µA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	$0.1V_{DD}$	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

Table 15. FAST configuration output buffer electrical characteristics

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
V_{OH}	CC	P	Output high level FAST configuration	Push Pull	I_{OH} = –14 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	$0.8V_{DD}$	—	—	V
		C			I_{OH} = –7 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	$0.8V_{DD}$	—	—	
		C			I_{OH} = –11 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	V_{DD} – 0.8	—	—	
V_{OL}	CC	P	Output low level FAST configuration	Push Pull	I_{OL} = 14 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
		C			I_{OL} = 7 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
		C			I_{OL} = 11 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

1 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only a transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

4.6.4 Output pin transition times

Table 16. Output pin transition times

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
T_{tr}	CC	D	Output transition time output pin[2] SLOW configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	50	ns
		T		C_L = 50 pF		—	—	100	
		D		C_L = 100 pF		—	—	125	
		D		C_L = 25 pF	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	50	
		T		C_L = 50 pF		—	—	100	
		D		C_L = 100 pF		—	—	125	
T_{tr}	CC	D	Output transition time output pin[2] MEDIUM configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 SIUL.PCRx.SRC = 1	—	—	10	ns
		T		C_L = 50 pF		—	—	20	
		D		C_L = 100 pF		—	—	40	
		D		C_L = 25 pF	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 SIUL.PCRx.SRC = 1	—	—	12	
		T		C_L = 50 pF		—	—	25	
		D		C_L = 100 pF		—	—	40	
T_{tr}	CC	D	Output transition time output pin[2] FAST configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	4	ns
				C_L = 50 pF		—	—	6	
				C_L = 100 pF		—	—	12	
				C_L = 25 pF	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	4	
				C_L = 50 pF		—	—	7	
				C_L = 100 pF		—	—	12	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 C_L includes device and package capacitances (C_{PKG} < 5 pF).

4.6.5 I/O pad current specification

The I/O pads are distributed across the I/O supply segment. Each I/O supply segment is associated to a V_{DD}/V_{SS} supply pair as described in Table 17.

Table 18 provides I/O consumption figures.

In order to ensure device reliability, the average current of the I/O on a single segment should remain below the I_{AVGSEG} maximum value.

Table 17. I/O supply segments

Package	Supply segment							
	1	2	3	4	5	6	7	8
208 MAPBGA[1]	Equivalent to 176 LQFP segment pad distribution						MCKO	MDOn /MSEO
176 LQFP	pin7 – pin27	pin28 – pin57	pin59 – pin85	pin86 – pin123	pin124 – pin150	pin151 – pin6	—	—
144 LQFP	pin20 – pin49	pin51 – pin99	pin100 – pin122	pin 123 – pin19	—	—	—	—
100 LQFP	pin16 – pin35	pin37 – pin69	pin70 – pin83	pin84 – pin15	—	—	—	—

1 208 MAPBGA available only as development package for Nexus2+

Table 18. I/O consumption

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{SWTSLW}[2]	CC	D	Dynamic I/O current for SLOW configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	20	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	16	
I_{SWTMED}[2]	CC	D	Dynamic I/O current for MEDIUM configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	29	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	17	
I_{SWTFST}[2]	CC	D	Dynamic I/O current for FAST configuration	C_L = 25 pF	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	110	mA
					V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	50	
I_{RMSSLW}	CC	D	Root medium square I/O current for SLOW configuration	C_L = 25 pF, 2 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	2.3	mA
				C_L = 25 pF, 4 MHz		—	—	3.2	
				C_L = 100 pF, 2 MHz		—	—	6.6	
				C_L = 25 pF, 2 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	1.6	
				C_L = 25 pF, 4 MHz		—	—	2.3	
				C_L = 100 pF, 2 MHz		—	—	4.7	
I_{RMSMED}	CC	D	Root medium square I/O current for MEDIUM configuration	C_L = 25 pF, 13 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	6.6	mA
				C_L = 25 pF, 40 MHz		—	—	13.4	
				C_L = 100 pF, 13 MHz		—	—	18.3	
				C_L = 25 pF, 13 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	5	
				C_L = 25 pF, 40 MHz		—	—	8.5	
				C_L = 100 pF, 13 MHz		—	—	11	

Table 18. I/O consumption (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{RMSFST}	CC	D	Root medium square I/O current for FAST configuration	C_L = 25 pF, 40 MHz	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	22	mA
				C_L = 25 pF, 64 MHz		—	—	33	
				C_L = 100 pF, 40 MHz		—	—	56	
				C_L = 25 pF, 40 MHz	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	14	
				C_L = 25 pF, 64 MHz		—	—	20	
				C_L = 100 pF, 40 MHz		—	—	35	
I_{AVGSEG}	SR	D	Sum of all the static I/O current within a supply segment	V_{DD} = 5.0 V ± 10%, PAD3V5V = 0		—	—	70	mA
				V_{DD} = 3.3 V ± 10%, PAD3V5V = 1		—	—	65	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to125 °C, unless otherwise specified

2 Stated maximum values represent peak consumption that lasts only a few ns during I/O transition.

Table 19 provides the weight of concurrent switching I/Os.

In order to ensure device functionality, the sum of the weight of concurrent switching I/Os on a single segment should remain below the 100%.

Table 19. I/O weight[1]

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PB[3]	5%	—	6%	—	13%	—	15%	—
PC[9]	4%	—	5%	—	13%	—	15%	—
PC[14]	4%	—	4%	—	13%	—	15%	—
PC[15]	3%	4%	4%	4%	12%	18%	15%	16%
PJ[4]	3%	4%	3%	3%	12%	18%	15%	16%
PH[15]	2%	3%	3%	3%	11%	16%	13%	14%
PH[13]	3%	4%	3%	4%	11%	15%	13%	14%
PH[14]	3%	4%	4%	4%	11%	15%	13%	13%
PI[6]	4%	—	4%	—	10%	—	12%	—
PI[7]	4%	—	4%	—	10%	—	12%	—
PG[5]	4%	—	5%	—	10%	—	12%	—
PG[4]	4%	6%	5%	5%	9%	13%	11%	12%
PG[3]	4%	—	5%	—	9%	—	11%	—
PG[2]	4%	6%	5%	5%	9%	12%	10%	11%
PA[2]	4%	—	5%	—	8%	—	10%	—
PE[0]	4%	—	5%	—	8%	—	9%	—

Table 19. I/O weight[1] (continued)

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PA[1]	4%	—	5%	—	8%	—	9%	—
PE[1]	4%	6%	5%	6%	7%	10%	9%	9%
PE[8]	4%	6%	5%	6%	7%	10%	8%	9%
PE[9]	4%	—	5%	—	6%	—	8%	—
PE[10]	4%	—	5%	—	6%	—	7%	—
PA[0]	4%	6%	5%	5%	6%	8%	7%	7%
PE[11]	4%	—	5%	—	5%	—	6%	—
PG[9]	9%	—	10%	—	9%	—	10%	—
PG[8]	9%	—	11%	—	9%	—	11%	—
PC[11]	9%	—	11%	—	9%	—	11%	—
PC[10]	9%	13%	11%	12%	9%	13%	11%	12%
PG[7]	9%	—	11%	—	9%	—	11%	—
PG[6]	10%	14%	11%	12%	10%	14%	11%	12%
PB[0]	10%	14%	12%	12%	10%	14%	12%	12%
PB[1]	10%	—	12%	—	10%	—	12%	—
PF[9]	10%	—	12%	—	10%	—	12%	—
PF[8]	10%	14%	12%	13%	10%	14%	12%	13%
PF[12]	10%	15%	12%	13%	10%	15%	12%	13%
PC[6]	10%	—	12%	—	10%	—	12%	—
PC[7]	10%	—	12%	—	10%	—	12%	—
PF[10]	10%	14%	11%	12%	10%	14%	11%	12%
PF[11]	9%	—	11%	—	9%	—	11%	—
PA[15]	8%	12%	10%	10%	8%	12%	10%	10%
PF[13]	8%	—	10%	—	8%	—	10%	—
PA[14]	8%	11%	9%	10%	8%	11%	9%	10%
PA[4]	7%	—	9%	—	7%	—	9%	—
PA[13]	7%	10%	8%	9%	7%	10%	8%	9%
PA[12]	7%	—	8%	—	7%	—	8%	—
PB[9]	1%	—	1%	—	1%	—	1%	—
PB[8]	1%	—	1%	—	1%	—	1%	—
PB[10]	5%	—	6%	—	6%	—	7%	—
PF[0]	5%	—	6%	—	6%	—	8%	—
PF[1]	5%	—	6%	—	7%	—	8%	—

Table 19. I/O weight[1] (continued)

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PF[2]	6%	—	7%	—	7%	—	9%	—
PF[3]	6%	—	7%	—	8%	—	9%	—
PF[4]	6%	—	7%	—	8%	—	10%	—
PF[5]	6%	—	7%	—	9%	—	10%	—
PF[6]	6%	—	7%	—	9%	—	11%	—
PF[7]	6%	—	7%	—	9%	—	11%	—
PJ[3]	6%	—	7%	—	10%	—	12%	—
PJ[2]	6%	—	7%	—	10%	—	12%	—
PJ[1]	6%	—	7%	—	11%	—	13%	—
PJ[0]	6%	—	7%	—	11%	—	13%	—
PI[15]	6%	—	7%	—	12%	—	14%	—
PI[14]	6%	—	7%	—	12%	—	14%	—
PD[0]	1%	—	1%	—	1%	—	1%	—
PD[1]	1%	—	1%	—	1%	—	1%	—
PD[2]	1%	—	1%	—	1%	—	1%	—
PD[3]	1%	—	1%	—	1%	—	1%	—
PD[4]	1%	—	1%	—	1%	—	1%	—
PD[5]	1%	—	1%	—	1%	—	1%	—
PD[6]	1%	—	1%	—	1%	—	2%	—
PD[7]	1%	—	1%	—	1%	—	2%	—
PD[8]	1%	—	1%	—	1%	—	2%	—
PB[4]	1%	—	1%	—	1%	—	2%	—
PB[5]	1%	—	1%	—	1%	—	2%	—
PB[6]	1%	—	1%	—	1%	—	2%	—
PB[7]	1%	—	1%	—	1%	—	2%	—
PD[9]	1%	—	1%	—	1%	—	2%	—
PD[10]	1%	—	1%	—	1%	—	2%	—
PD[11]	1%	—	1%	—	1%	—	2%	—
PB[11]	1%	—	1%	—	1%	—	2%	—
PD[12]	11%	—	13%	—	15%	—	18%	—
PB[12]	11%	—	13%	—	15%	—	17%	—
PD[13]	11%	—	13%	—	14%	—	17%	—
PB[13]	11%	—	13%	—	14%	—	17%	—

Table 19. I/O weight[1] (continued)

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PD[14]	11%	—	13%	—	14%	—	17%	—
PB[14]	11%	—	13%	—	14%	—	16%	—
PD[15]	11%	—	13%	—	13%	—	16%	—
PB[15]	11%	—	13%	—	13%	—	15%	—
PI[8]	10%	—	12%	—	12%	—	15%	—
PI[9]	10%	—	12%	—	12%	—	15%	—
PI[10]	10%	—	12%	—	12%	—	14%	—
PI[11]	10%	—	12%	—	12%	—	14%	—
PI[12]	10%	—	12%	—	11%	—	13%	—
PI[13]	10%	—	11%	—	11%	—	13%	—
PA[3]	9%	—	11%	—	11%	—	13%	—
PG[13]	9%	13%	11%	11%	10%	14%	12%	13%
PG[12]	9%	13%	10%	11%	10%	14%	12%	12%
PH[0]	6%	8%	7%	7%	6%	9%	7%	8%
PH[1]	6%	8%	7%	7%	6%	8%	7%	7%
PH[2]	5%	7%	6%	6%	5%	7%	6%	7%
PH[3]	5%	7%	5%	6%	5%	7%	6%	6%
PG[1]	4%	—	5%	—	4%	—	5%	—
PG[0]	4%	5%	4%	5%	4%	5%	4%	5%
PF[15]	4%	—	4%	—	4%	—	4%	—
PF[14]	4%	6%	5%	5%	4%	6%	5%	5%
PE[13]	4%	—	5%	—	4%	—	5%	—
PA[7]	5%	—	6%	—	5%	—	6%	—
PA[8]	5%	—	6%	—	5%	—	6%	—
PA[9]	6%	—	7%	—	6%	—	7%	—
PA[10]	6%	—	8%	—	6%	—	8%	—
PA[11]	8%	—	9%	—	8%	—	9%	—
PE[12]	8%	—	9%	—	8%	—	9%	—
PG[14]	8%	—	9%	—	8%	—	9%	—
PG[15]	8%	11%	9%	10%	8%	11%	9%	10%
PE[14]	8%	—	9%	—	8%	—	9%	—
PE[15]	8%	11%	9%	10%	8%	11%	9%	10%
PG[10]	8%	—	9%	—	8%	—	9%	—

Table 19. I/O weight[1] (continued)

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PG[11]	7%	11%	9%	9%	7%	11%	9%	9%
PH[11]	7%	10%	9%	9%	7%	10%	9%	9%
PH[12]	7%	10%	8%	9%	7%	10%	8%	9%
PI[5]	7%	—	8%	—	7%	—	8%	—
PI[4]	7%	—	8%	—	7%	—	8%	—
PC[3]	6%	—	8%	—	6%	—	8%	—
PC[2]	6%	8%	7%	7%	6%	8%	7%	7%
PA[5]	6%	8%	7%	7%	6%	8%	7%	7%
PA[6]	5%	—	6%	—	5%	—	6%	—
PH[10]	5%	7%	6%	6%	5%	7%	6%	6%
PC[1]	5%	19%	5%	13%	5%	19%	5%	13%
PC[0]	6%	9%	7%	8%	7%	10%	8%	8%
PH[9]	7%	—	8%	—	7%	—	9%	—
PE[2]	7%	10%	8%	9%	8%	11%	9%	10%
PE[3]	7%	10%	9%	9%	8%	12%	10%	10%
PC[5]	7%	11%	9%	9%	8%	12%	10%	11%
PC[4]	8%	11%	9%	10%	9%	13%	10%	11%
PE[4]	8%	11%	9%	10%	9%	13%	11%	12%
PE[5]	8%	11%	10%	10%	9%	14%	11%	12%
PH[4]	8%	12%	10%	10%	10%	14%	12%	12%
PH[5]	8%	—	10%	—	10%	—	12%	—
PH[6]	8%	12%	10%	11%	10%	15%	12%	13%
PH[7]	9%	12%	10%	11%	11%	15%	13%	13%
PH[8]	9%	12%	10%	11%	11%	16%	13%	14%
PE[6]	9%	12%	10%	11%	11%	16%	13%	14%
PE[7]	9%	12%	10%	11%	11%	16%	14%	14%
PI[3]	9%	—	10%	—	12%	—	14%	—
PI[2]	9%	—	10%	—	12%	—	14%	—
PI[1]	9%	—	10%	—	12%	—	15%	—
PI[0]	9%	—	10%	—	12%	—	15%	—

Table 19. I/O weight[1] (continued)

PAD	176 LQFP				144/100 LQFP			
	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1	Weight 5V SRE = 0	Weight 5V SRE = 1	Weight 3.3V SRE = 0	Weight 3.3V SRE = 1
PC[12]	8%	12%	10%	11%	12%	18%	15%	16%
PC[13]	8%	—	10%	—	13%	—	15%	—
PC[8]	8%	—	10%	—	13%	—	15%	—
PB[2]	8%	11%	9%	10%	13%	18%	15%	16%

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

4.7 $\overline{\text{RESET}}$ electrical characteristics

The device implements a dedicated bidirectional $\overline{\text{RESET}}$ pin.

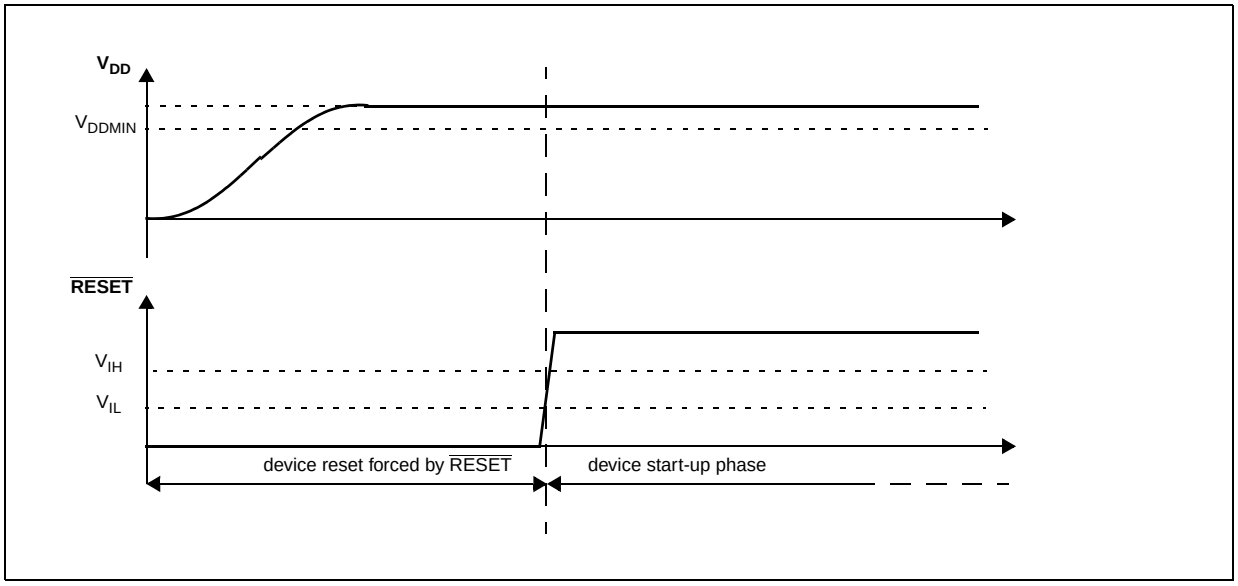


Figure 7. Start-up reset requirements

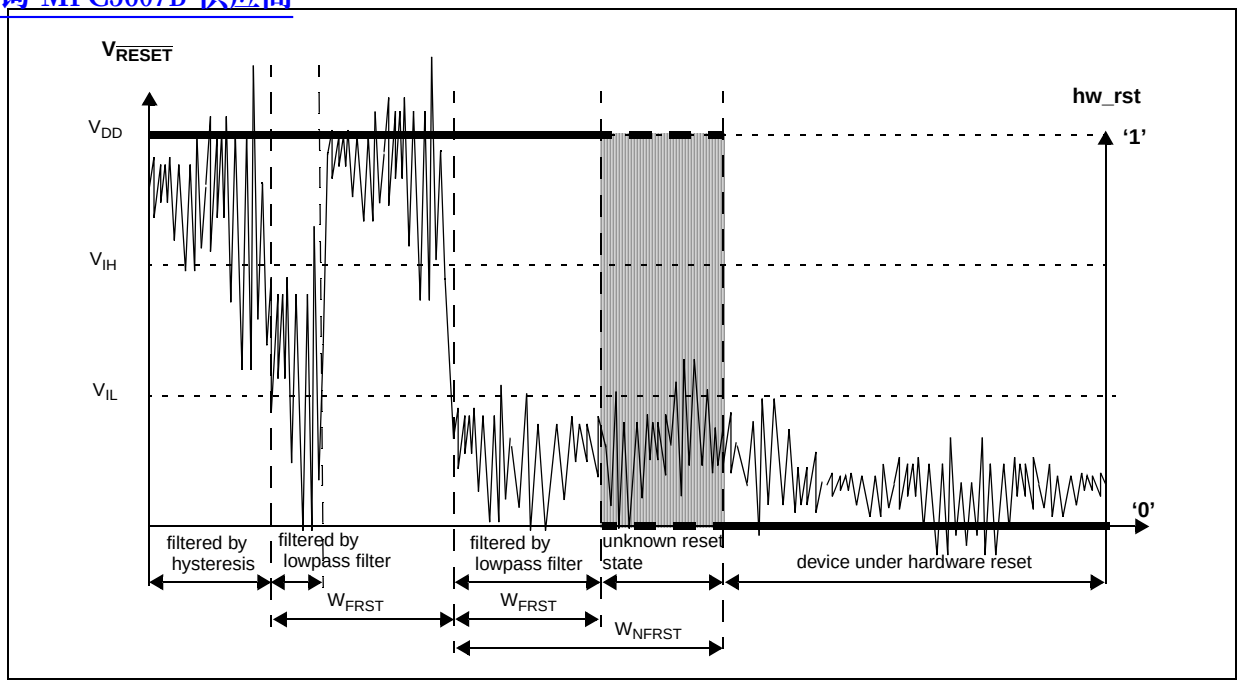


Figure 8. Noise filtering on reset signal

Table 20. Reset electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{IH}	SR	P	Input High Level CMOS (Schmitt Trigger)	—	$0.65V_{DD}$	—	V_{DD} + 0.4	V
V_{IL}	SR	P	Input low Level CMOS (Schmitt Trigger)	—	–0.4	—	$0.35V_{DD}$	V
V_{HYS}	CC	C	Input hysteresis CMOS (Schmitt Trigger)	—	$0.1V_{DD}$	—	—	V
V_{OL}	CC	P	Output low level	Push Pull, I_{OL} = 2 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 (recommended)	—	—	$0.1V_{DD}$	V
				Push Pull, I_{OL} = 1 mA, V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[2]	—	—	$0.1V_{DD}$	
				Push Pull, I_{OL} = 1 mA, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 (recommended)	—	—	0.5	

Table 20. Reset electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
T_{tr}	CC	D	Output transition time output pin[3] MEDIUM configuration	C_L = 25 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	10	ns
				C_L = 50 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	20	
				C_L = 100 pF, V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	—	—	40	
				C_L = 25 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	12	
				C_L = 50 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	25	
				C_L = 100 pF, V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	—	—	40	
W_{FRST}	SR	P	$\overline{\text{RESET}}$ input filtered pulse	—	—	—	40	ns
W_{NFRST}	SR	P	$\overline{\text{RESET}}$ input not filtered pulse	—	1000	—	—	ns
$\lvert I_{WPU} \rvert$	CC	P	Weak pull-up current absolute value	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1	10	—	150	µA
				V_{DD} = 5.0 V ± 10%, PAD3V5V = 0	10	—	150	
				V_{DD} = 5.0 V ± 10%, PAD3V5V = 1[4]	10	—	250	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This is a transient configuration during power-up, up to the end of reset PHASE2 (refer to RGM module section of the device reference manual).

3 C_L includes device and package capacitance (C_{PKG} < 5 pF).

4 The configuration PAD3V5 = 1 when V_{DD} = 5 V is only transient configuration during power-up. All pads but RESET and Nexus output (MDOx, EVTO, MCKO) are configured in input or in high impedance state.

4.8 Power management electrical characteristics

4.8.1 Voltage regulator electrical characteristics

The device implements an internal voltage regulator to generate the low voltage core supply V_{DD_LV} from the high voltage ballast supply V_{DD_BV}. The regulator itself is supplied by the common I/O supply V_{DD}. The following supplies are involved:

- HV: High voltage external power supply for voltage regulator module. This must be provided externally through V_{DD} power pin.
- BV: High voltage external power supply for internal ballast module. This must be provided externally through V_{DD_BV} power pin. Voltage values should be aligned with V_{DD}.
- LV: Low voltage internal power supply for core, FMPLL and Flash digital logic. This is generated by the internal voltage regulator but provided outside to connect stability capacitor. It is further split into four main domains to ensure noise isolation between critical LV modules within the device:
  - LV_COR: Low voltage supply for the core. It is also used to provide supply for FMPLL through double bonding.
  - LV_CFLA: Low voltage supply for code Flash module. It is supplied with dedicated ballast and shorted to LV_COR through double bonding.

— LV_DFLA: Low voltage supply for data Flash module. It is supplied with dedicated ballast and shorted to LV_COR through double bonding.

— LV_PLL: Low voltage supply for FMPLL. It is shorted to LV_COR through double bonding.

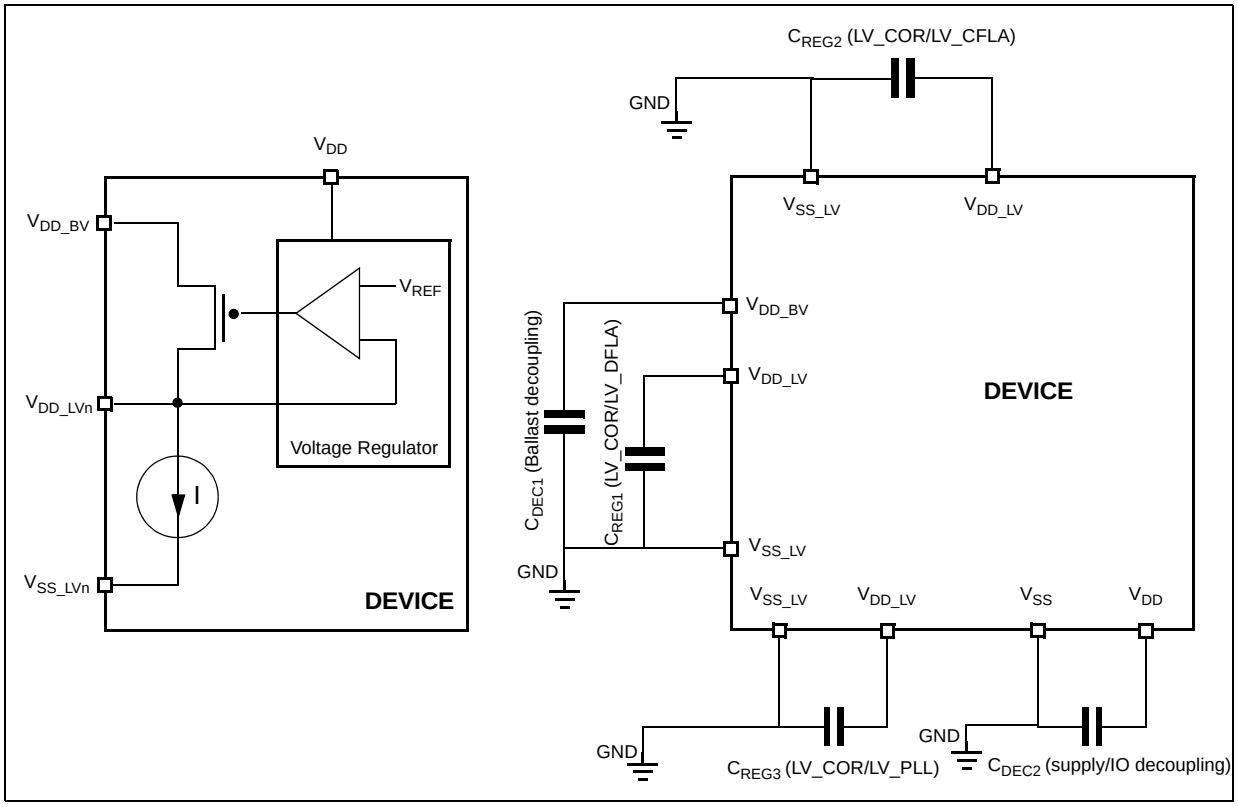


Figure 9. Voltage regulator capacitance connection

The internal voltage regulator requires external capacitance (C_{REGn}) to be connected to the device in order to provide a stable low voltage digital supply to the device. Capacitances should be placed on the board as near as possible to the associated pins. Care should also be taken to limit the serial inductance of the board to less than 5 nH.

Each decoupling capacitor must be placed between each of the three V_{DD_LV}/V_{SS_LV} supply pairs to ensure stable voltage (see Section 4.4, "Recommended operating conditions).

Table 21. Voltage regulator electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
C_{REGn}	SR	—	Internal voltage regulator external capacitance	—	200	—	500	nF
R_{REG}	SR	—	Stability capacitor equivalent serial resistance	—	—	—	0.2	Ω
C_{DEC1}	SR	—	Decoupling capacitance[2] ballast	V_{DD_BV}/V_{SS_LV} pair: V_{DD_BV} = 4.5 V to 5.5 V	100[3]	470[4]	—	nF
				V_{DD_BV}/V_{SS_LV} pair: V_{DD_BV} = 3 V to 3.6 V	400		—	

Table 21. Voltage regulator electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
C_{DEC2}	SR	—	Decoupling capacitance regulator supply	V_{DD}/V_{SS} pair	10	100	—	nF
V_{MREG}	CC	P	Main regulator output voltage	Before exting from reset	—	1.32	—	V
				After trimming	1.15	1.28	1.32	
I_{MREG}	SR	—	Main regulator current provided to V_{DD_LV} domain	—	—	—	150	mA
$I_{MREGINT}$	CC	D	Main regulator module current consumption	I_{MREG} = 200 mA	—	—	2	mA
				I_{MREG} = 0 mA	—	—	1	
V_{LPREG}	CC	P	Low power regulator output voltage	After trimming	1.15	1.23	1.32	V
I_{LPREG}	SR	—	Low power regulator current provided to V_{DD_LV} domain	—	—	—	15	mA
$I_{LPREGINT}$	CC	D	Low power regulator module current consumption	I_{LPREG} = 15 mA; T_A = 55 °C	—	—	600	µA
		—		I_{LPREG} = 0 mA; T_A = 55 °C	—	5	—	
V_{ULPREG}	CC	P	Ultra low power regulator output voltage	After trimming	1.15	1.23	1.32	V
I_{ULPREG}	SR	—	Ultra low power regulator current provided to V_{DD_LV} domain	—	—	—	5	mA
$I_{ULPREGINT}$	CC	D	Ultra low power regulator module current consumption	I_{ULPREG} = 5 mA; T_A = 55 °C	—	—	100	µA
				I_{ULPREG} = 0 mA; T_A = 55 °C	—	2	—	
I_{DD_BV}	CC	D	In-rush current on V_{DD_BV} during power-up[5]	—	—	—	400[6]	mA

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This capacitance value is driven by the constraints of the external voltage regulator supplying the V_{DD_BV} voltage. A typical value is in the range of 470 nF.

3 This value is acceptable to guarantee operation from 4.5 V to 5.5 V

4 External regulator and capacitance circuitry must be capable of providing I_{DD_BV} while maintaining supply V_{DD_BV} in operating range.

5 In-rush current is seen only for short time during power-up and on standby exit (max 20µs, depending on external capacitances to be load).

6 The duration of the in-rush current depends on the capacitance placed on LV pins. BV decaps must be sized accordingly. Refer to IMREG value for minimum amount of current to be provided in cc.

4.8.2 Voltage monitor electrical characteristics

The device implements a Power-on Reset module to ensure correct power-up initialization, as well as four low voltage detectors to monitor the V_{DD} and the V_{DD_LV} voltage while device is supplied:

- POR monitors V_{DD} during the power-up phase to ensure device is maintained in a safe reset state
- LVDHV3 monitors V_{DD} to ensure device reset below minimum functional supply
- LVDHV3B monitors VDD_BV to ensure device reset below minimum functional supply
- LVDHV5 monitors V_{DD} when application uses device in the 5.0 V ± 10% range
- LVDLVCOR monitors power domain No. 1
- LVDLVBKP monitors power domain No. 0

NOTE

When enabled, power domain No. 2 is monitored through LVDLVBKP.

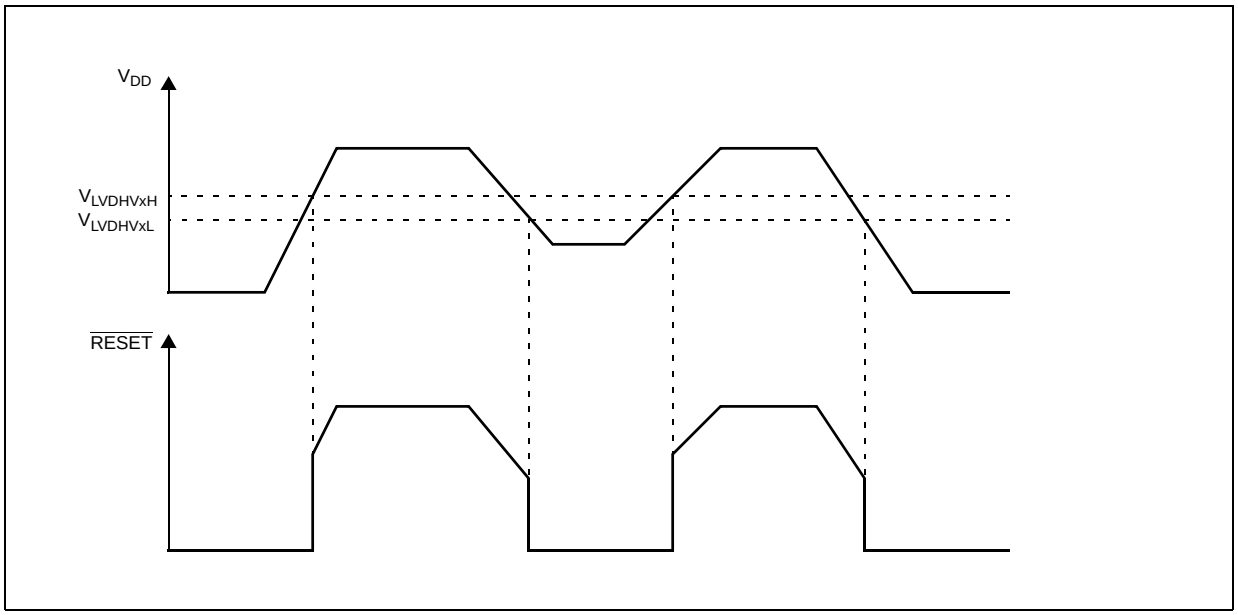


Figure 10. Low voltage monitor vs reset

Table 22. Low voltage monitor electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{PORUP}	SR	P	Supply for functional POR module	T_A = 25 °C, after trimming	1.0	—	5.5	V
V_{PORH}	CC	P	Power-on reset threshold		1.5	—	2.6	
$V_{LVDHV3H}$	CC	T	LVDHV3 low voltage detector high threshold		—	—	2.95	
$V_{LVDHV3L}$	CC	P	LVDHV3 low voltage detector low threshold		2.7	—	2.9	
$V_{LVDHV3BH}$	CC	P	LVDHV3B low voltage detector high threshold		—	—	2.95	
$V_{LVDHV3BL}$	CC	P	LVDHV3BL low voltage detector low threshold		2.7	—	2.9	
$V_{LVDHV5H}$	CC	T	LVDHV5 low voltage detector high threshold		—	—	4.5	
$V_{LVDHV5L}$	CC	P	LVDHV5 low voltage detector low threshold		3.8	—	4.4	
$V_{LVDLVCORL}$	CC	P	LVDLVCOR low voltage detector low threshold		1.08	—	—	
$V_{LVDLVBKPL}$	CC	P	LVDLVBKP low voltage detector low threshold		1.08	—	1.14	

[1] V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

4.9 Low voltage domain power consumption

Table 23 provides DC electrical characteristics for significant application modes. These values are indicative values; actual consumption depends on the application.

Table 23. Low voltage power domain electrical characteristics[1]

Symbol		C	Parameter	Conditions[2]		Value			Unit
						Min	Typ	Max	
I_{DDMAX}[3]	CC	D	RUN mode maximum average current	—		—	115	140[4]	mA
I_{DDRUN}[5]	CC	T	RUN mode typical average current[6]	f_{CPU} = 8 MHz		—	12	—	mA
		T		f_{CPU} = 16 MHz		—	27	—	
		T		f_{CPU} = 32 MHz		—	43	—	
		P		f_{CPU} = 48 MHz		—	56	—	
		P		f_{CPU} = 64 MHz		—	70	—	
I_{DDHALT}	CC	C	HALT mode current[7]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	10	18	mA
		P			T_A = 125 °C	—	17	28	
I_{DDSTOP}	CC	P	STOP mode current[8]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	350	900	µA
		D			T_A = 55 °C	—	750	—	
		D			T_A = 85 °C	—	2	—	mA
		D			T_A = 105 °C	—	4	—	
		P			T_A = 125 °C	—	7	14	
$I_{DDSTDBY2}$	CC	P	STANDBY2 mode current[9]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	30	100	µA
		D			T_A = 55 °C	—	75	—	
		D			T_A = 85 °C	—	180	—	
		D			T_A = 105 °C	—	315	—	
		P			T_A = 125 °C	—	560	1700	
$I_{DDSTDBY1}$	CC	T	STANDBY1 mode current[10]	Slow internal RC oscillator (128 kHz) running	T_A = 25 °C	—	20	60	µA
		D			T_A = 55 °C	—	45	—	
		D			T_A = 85 °C	—	100	—	
		D			T_A = 105 °C	—	165	—	
		D			T_A = 125 °C	—	280	900	

1 Except for I_{DDMAX}, all consumptions in this table apply to V_{DD_BV} only and do not include V_{DD_HV}.

2 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

3 Running consumption is given on voltage regulator supply (V_{DDREG}). I_{DDMAX} is composed of three components: $I_{DDMAX} = I_{DD}(V_{DD\_BV}) + I_{DD}(V_{DD\_HV}) + I_{DD}(V_{DD\_HV\_ADC})$. It does not include a fourth component linked to I/Os toggling which is **highly** dependent on the application. The given value is thought to be a **worst case value** with all peripherals running, and code fetched from code flash while modify operation on-going on data flash. Note that this value can be significantly reduced by the application: switch off not used peripherals (default), reduce peripheral frequency through internal prescaler, fetch from RAM most used functions, use low power mode when possible.

4 Higher current may be sunk by device during power-up and standby exit. Please refer to in-rush current in Table 21.

5 RUN current measured with typical application with accesses on both Flash and RAM.

6 Only for the “P” classification: Data and Code Flash in Normal Power. Code fetched from RAM: Serial IPs CAN and LIN in loop back mode, DSPI as Master, PLL as system clock (4 x Multiplier) peripherals on (eMIOS/CTU/ADC) and running at max frequency, periodic SW/WDG timer reset enabled.

7 Data Flash Power Down. Code Flash in Low Power. SIRC 128 kHz and FIRC 16 MHz on. 10 MHz XTAL clock. FlexCAN: instances: 0, 1, 2 ON (clocked but not reception or transmission), instances: 4, 5, 6 clocks gated. LINFlex: instances: 0, 1, 2 ON (clocked but not reception or transmission), instance: 3 to 9 clocks gated. eMIOS: instance: 0 ON (16 channels on PA[0]–PA[11] and PC[12]–PC[15]) with PWM 20 kHz, instance: 1 clock gated. DSPI: instance: 0 (clocked but no communication), instance: 1 to 5 clocks gated. RTC/API ON. PIT ON. STM ON. ADC1 OFF. ADC0 ON but no conversion except two analog watchdogs.

8 Only for the “P” classification: No clock, FIRC 16 MHz off, SIRC 128 kHz on, PLL off, HPvreg off, ULPVreg/LPVreg on. All possible peripherals off and clock gated. Flash in power down mode.

9 Only for the “P” classification: ULPreg on, HP/LPVreg off, 32 KB RAM on, device configured for minimum consumption, all possible modules switched off.

10 ULPreg on, HP/LPVreg off, 8 KB RAM on, device configured for minimum consumption, all possible modules switched off.

4.10 Flash memory electrical characteristics

4.10.1 Program/erase characteristics

Table 24 shows the program and erase characteristics.

Table 24. Program and erase specifications

Symbol		C	Parameter	Conditions	Value				Unit
					Min	Typ[1]	Initial max[2]	Max[3]	
$T_{dwprogram}$	CC	C	Double word (64 bits) program time[4]	Code Flash	—	18	50	500	µs
				Data Flash		22			
$T_{16Kpperase}$			16 KB block preprogram and erase time	Code Flash	—	200	500	5000	ms
				Data Flash		300			
$T_{32Kpperase}$			32 KB block preprogram and erase time	Code Flash	—	300	600	5000	ms
				Data Flash		400			
$T_{128Kpperase}$			128 KB block preprogram and erase time	Code Flash	—	600	1300	7500	ms
				Data Flash		800			
T_{eslat}		D	Erase Suspend Latency	—	—	—	30	30	µs
T_{ESRT}		C	Erase Suspend Request Rate	Code Flash	20	—	—	—	ms
				Data Flash	10	—	—	—	

1 Typical program and erase times assume nominal supply values and operation at 25 °C. All times are subject to change pending device characterization.

2 Initial factory condition: < 100 program/erase cycles, 25 °C, typical supply voltage.

3 The maximum program and erase times occur after the specified number of program/erase cycles. These maximum values are characterized but not guaranteed.

4 Actual hardware programming times. This does not include software overhead.

Table 25. Flash module life

Symbol		C	Parameter	Conditions	Value			Unit
					Min	Typ	Max	
P/E	CC	C	Number of program/erase cycles per block for 16 KB blocks over the operating temperature range (T_J)	—	100000	—	—	cycles
P/E	CC	C	Number of program/erase cycles per block for 32 KB blocks over the operating temperature range (T_J)	—	10000	100000	—	cycles
P/E	CC	C	Number of program/erase cycles per block for 128 KB blocks over the operating temperature range (T_J)	—	1000	100000	—	cycles
Retention	CC	C	Minimum data retention at 85 °C average ambient temperature[1]	Blocks with 0–1,000 P/E cycles	20	—	—	years
				Blocks with 1,001–10,000 P/E cycles	10	—	—	years
				Blocks with 10,001–100,000 P/E cycles	5	—	—	years

1 Ambient temperature averaged over duration of application, not to exceed recommended product operating temperature range.

ECC circuitry provides correction of single bit faults and is used to improve further automotive reliability results. Some units will experience single bit corrections throughout the life of the product with no impact to product reliability.

Table 26. Flash read access timing

Symbol		C	Parameter	Conditions[1]	Max	Unit
f_{READ}	CC	P	Maximum frequency for Flash reading	2 wait states	64	MHz
		C		1 wait state	40	
		C		0 wait states	20	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

Table 27 shows the FLASH_BIU settings versus frequency of operation. Refer to the device reference manual for definitions of these bit fields.

Table 27. FLASH_BIU settings vs. frequency of operation

Maximum frequency (MHz)	APC	RWSC	WWSC	DPFEN	IPFEN	PFLIM	BFEN
Up to and including 82 MHz	TBD	TBD	TBD	TBD	TBD	TBD	TBD
Up to and including 102 MHz	TBD	TBD	TBD	TBD	TBD	TBD	TBD
Up to and including 132 MHz	TBD	TBD	TBD	TBD	TBD	TBD	TBD
Default setting after reset	TBD	TBD	TBD	TBD	TBD	TBD	TBD

4.10.2 Flash power supply DC characteristics

Table 28 shows the power supply DC characteristics on external supply.

Table 28. Flash power supply DC electrical characteristics

Symbol		Parameter	Conditions[1]		Value			Unit
					Min	Typ	Max	
I_{CFREAD}	CC	Sum of the current consumption on V_{DDHV} and V_{DDBV} on read access	Flash module read f_{CPU} = 64 MHz[2]	Code Flash	—	—	33	mA
I_{DFREAD}				Data Flash	—	—	33	
I_{CFMOD}	CC	Sum of the current consumption on V_{DDHV} and V_{DDBV} on matrix modification (program/erase)	Program /Erase on-going while reading Flash registers f_{CPU} = 64 MHz[2]	Code Flash	—	—	52	mA
I_{DFMOD}				Data Flash	—	—	33	
I_{CFLPW}	CC	Sum of the current consumption on V_{DDHV} and V_{DDBV} during Flash low power mode	—	Code Flash	—	—	1.1	mA
I_{DFLPW}				Data Flash	—	—	900	µA
I_{CFPWD}	CC	Sum of the current consumption on V_{DDHV} and V_{DDBV} during Flash power down mode	—	Code Flash	—	—	150	µA
I_{DFPWD}				Data Flash	—	—	150	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 f_{CPU} 64 MHz can be achieved only at up to 105 °C

4.10.3 Start-up/Switch-off timings

Table 29. Start-up time/Switch-off time

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
$T_{FLARSTEXIT}$	CC	T	Delay for Flash module to exit reset mode	—	—	—	125	µs
$T_{FLALPEXIT}$	CC	T	Delay for Flash module to exit low-power mode	—	—	—	0.5	
$T_{FLAPDEXIT}$	CC	T	Delay for Flash module to exit power-down mode	—	—	—	30	
$T_{FLALPENTRY}$	CC	T	Delay for Flash module to enter low-power mode	—	—	—	0.5	
$T_{FLAPDENTRY}$	CC	T	Delay for Flash module to enter power-down mode	—	—	—	1.5	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

4.11 Electromagnetic compatibility (EMC) characteristics

Susceptibility tests are performed on a sample basis during product characterization.

4.11.1 Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user apply EMC software optimization and prequalification tests in relation with the EMC level requested for the application.

- Software recommendations – The software flowchart must include the management of runaway conditions such as:
  - Corrupted program counter
  - Unexpected reset
  - Critical data corruption (control registers...)
- Prequalification trials – Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the reset pin or the oscillator pins for 1 second.

  To complete these trials, ESD stress can be applied directly on the device. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring.

4.11.2 Electromagnetic interference (EMI)

The product is monitored in terms of emission based on a typical application. This emission test conforms to the IEC61967-1 standard, which specifies the general conditions for EMI measurements.

Table 30. EMI radiated emission measurement[1,2]

Symbol		C	Parameter	Conditions		Value			Unit
						Min	Typ	Max	
—	SR	—	Scan range	—		0.150		1000	MHz
f_{CPU}	SR	—	Operating frequency	—		—	64	—	MHz
V_{DD_LV}	SR	—	LV operating voltages	—		—	1.28	—	V
S_{EMI}	CC	T	Peak level	V_{DD} = 5 V, T_A = 25 °C, LQFP144 package Test conforming to IEC 61967-2, f_{OSC} = 8 MHz/f_{CPU} = 64 MHz	No PLL frequency modulation	—	—	18	dBµV
					± 2% PLL frequency modulation	—	—	14	dBµV

1 EMI testing and I/O port waveforms per IEC 61967-1, -2, -4

2 For information on conducted emission and susceptibility measurement (norm IEC 61967-4), please contact your local marketing representative.

4.11.3 Absolute maximum ratings (electrical sensitivity)

Based on two different tests (ESD and LU) using specific measurement methods, the product is stressed in order to determine its performance in terms of electrical sensitivity.

4.11.3.1 Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts×(n + 1) supply pin). This test conforms to the AEC-Q100-002/-003/-011 standard.

Table 31. ESD absolute maximum ratings[1,2]

Symbol	Ratings	Conditions	Class	Max value[3]	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (Human Body Model)	T_A = 25 °C conforming to AEC-Q100-002	H1C	2000	V
$V_{ESD(MM)}$	Electrostatic discharge voltage (Machine Model)	T_A = 25 °C conforming to AEC-Q100-003	M2	200	
$V_{ESD(CDM)}$	Electrostatic discharge voltage (Charged Device Model)	T_A = 25 °C conforming to AEC-Q100-011	C3A	500	
				750 (corners)	

1 All ESD testing is in conformity with CDF-AEC-Q100 Stress Test Qualification for Automotive Grade Integrated Circuits.

2 A device will be defined as a failure if after exposure to ESD pulses the device no longer meets the device specification requirements. Complete DC parametric and functional testing shall be performed per applicable device specification at room temperature followed by hot temperature, unless specified otherwise in the device specification.

3 Data based on characterization results, not tested in production

4.11.3.2 Static latch-up (LU)

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with the EIA/JESD 78 IC latch-up standard.

Table 32. Latch-up results

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T_A = 125 °C conforming to JESD 78	II level A

4.12 Fast external crystal oscillator (4 to 16 MHz) electrical characteristics

The device provides an oscillator/resonator driver. Figure 11 describes a simple model of the internal oscillator driver and provides an example of a connection for an oscillator or a resonator.

Table 33 provides the parameter description of 4 MHz to 16 MHz crystals used for the design simulations.

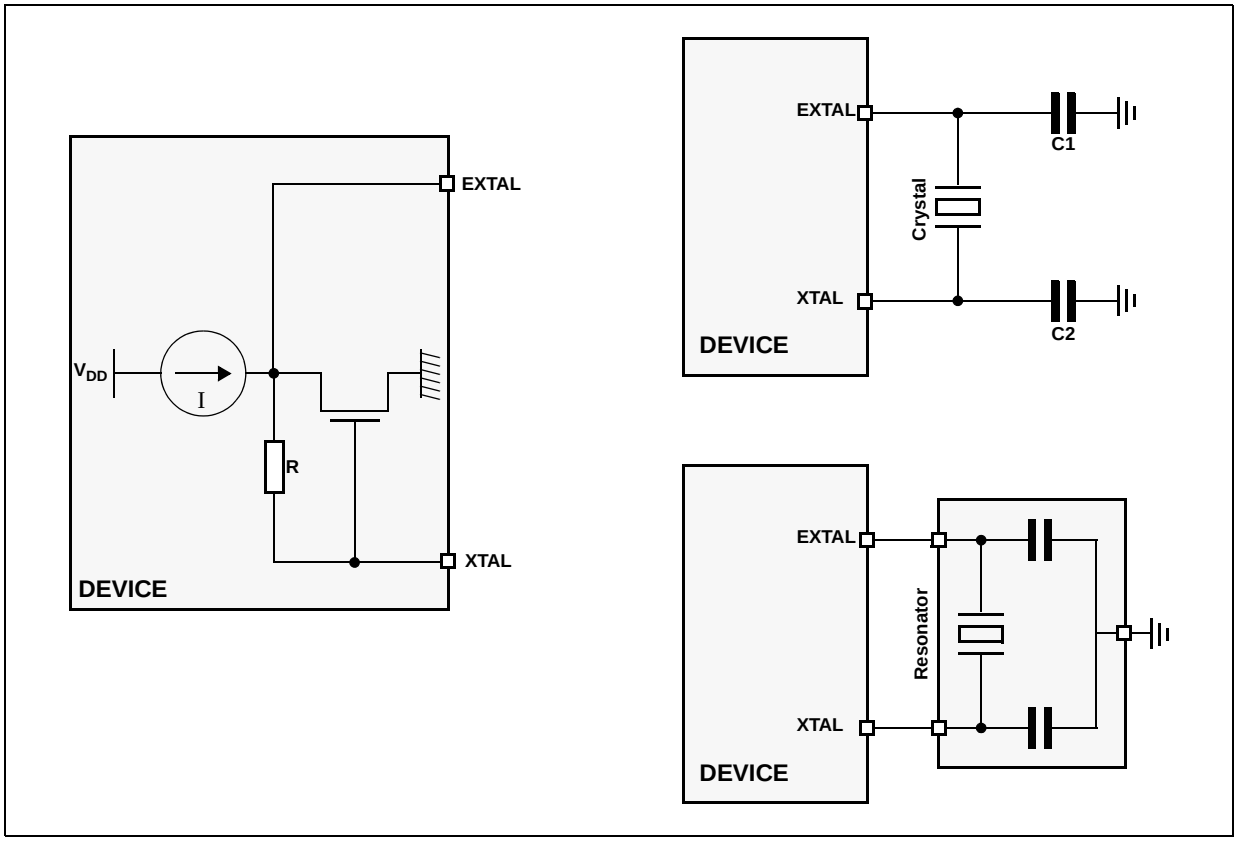


Figure 11. Crystal oscillator and resonator connection scheme

NOTE

XTAL/EXTAL must not be directly used to drive external circuits.

Table 33. Crystal description

Nominal frequency (MHz)	NDK crystal reference	Crystal equivalent series resistance ESR Ω	Crystal motional capacitance (C_m) fF	Crystal motional inductance (L_m) mH	Load on xtalin/xtalout C1 = C2 (pF)[1]	Shunt capacitance between xtalout and xtalin C0[2] (pF)
4	NX8045GB	300	2.68	591.0	21	2.93
8	NX5032GA	300	2.46	160.7	17	3.01
10		150	2.93	86.6	15	2.91
12		120	3.11	56.5	15	2.93
16		120	3.90	25.3	10	3.00

1 The values specified for C1 and C2 are the same as used in simulations. It should be ensured that the testing includes all the parasitics (from the board, probe, crystal, etc.) as the AC / transient behavior depends upon them.

2 The value of C0 specified here includes 2 pF additional capacitance for parasitics (to be seen with bond-pads, package, etc.).

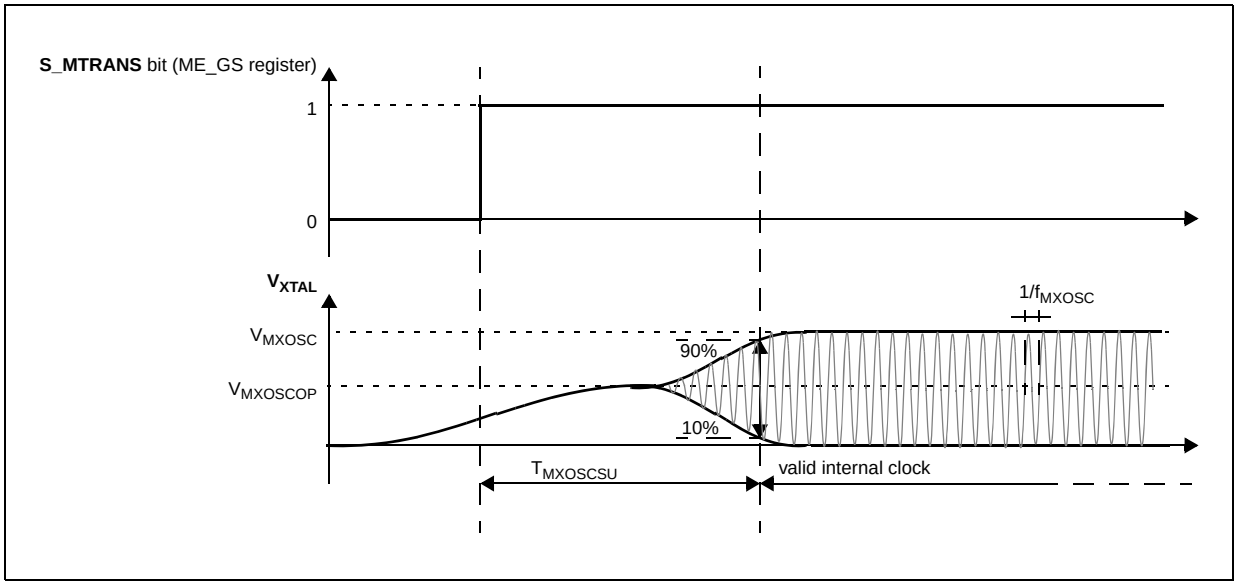


Figure 12. Fast external crystal oscillator (4 to 16 MHz) electrical characteristics

Table 34. Fast external crystal oscillator (4 to 16 MHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{FXOSC}	SR	—	Fast external crystal oscillator frequency	—	4.0	—	16.0	MHz

Table 34. Fast external crystal oscillator (4 to 16 MHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
g_{mFXOSC}	CC	C	Fast external crystal oscillator transconductance	V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 0	2.2	—	8.2	mA/V
	CC	P		V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 0	2.0	—	7.4	
	CC	C		V_{DD} = 3.3 V ± 10%, PAD3V5V = 1 OSCILLATOR_MARGIN = 1	2.7	—	9.7	
	CC	C		V_{DD} = 5.0 V ± 10%, PAD3V5V = 0 OSCILLATOR_MARGIN = 1	2.5	—	9.2	
V_{FXOSC}	CC	T	Oscillation amplitude at EXTAL	f_{OSC} = 4 MHz, OSCILLATOR_MARGIN = 0	1.3	—	—	V
				f_{OSC} = 16 MHz, OSCILLATOR_MARGIN = 1	1.3	—	—	
$V_{FXOSCOP}$	CC	P	Oscillation operating point	—	—	0.95		V
I_{FXOSC}[2]	CC	T	Fast external crystal oscillator consumption	—	—	2	3	mA
$T_{FXOSCSU}$	CC	T	Fast external crystal oscillator start-up time	f_{OSC} = 4 MHz, OSCILLATOR_MARGIN = 0	—	—	6	ms
				f_{OSC} = 16 MHz, OSCILLATOR_MARGIN = 1	—	—	1.8	
V_{IH}	SR	P	Input high level CMOS (Schmitt Trigger)	Oscillator bypass mode	$0.65V_{DD}$	—	V_{DD} + 0.4	V
V_{IL}	SR	P	Input low level CMOS (Schmitt Trigger)	Oscillator bypass mode	–0.4	—	$0.35V_{DD}$	V

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 Stated values take into account only analog module consumption but not the digital contributor (clock tree and enabled peripherals).

4.13 Slow external crystal oscillator (32 kHz) electrical characteristics

The device provides a low power oscillator/resonator driver.

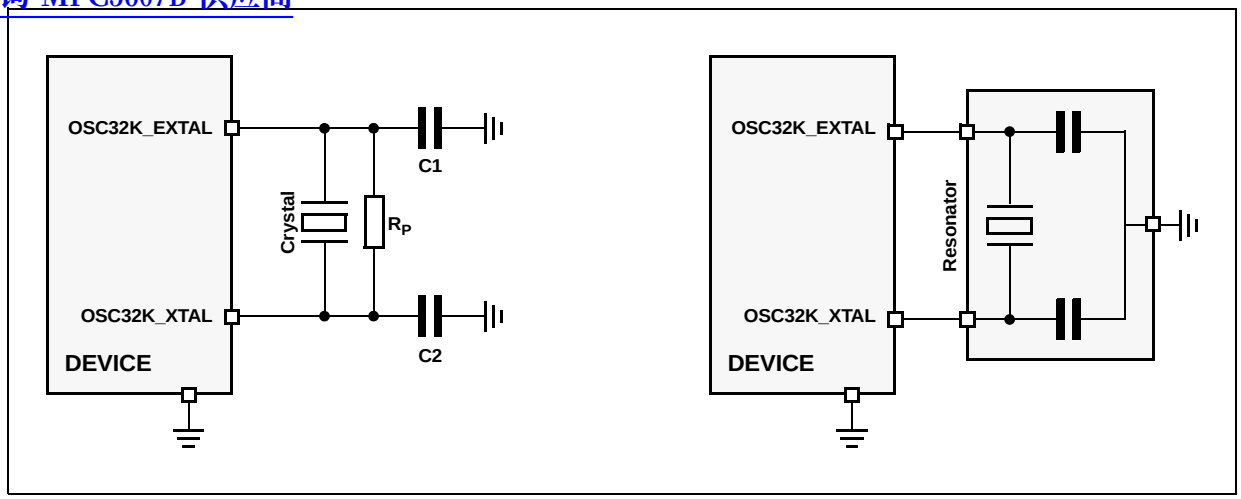


Figure 13. Crystal oscillator and resonator connection scheme

NOTE

OSC32K_XTAL/OSC32K_EXTAL must not be directly used to drive external circuits.

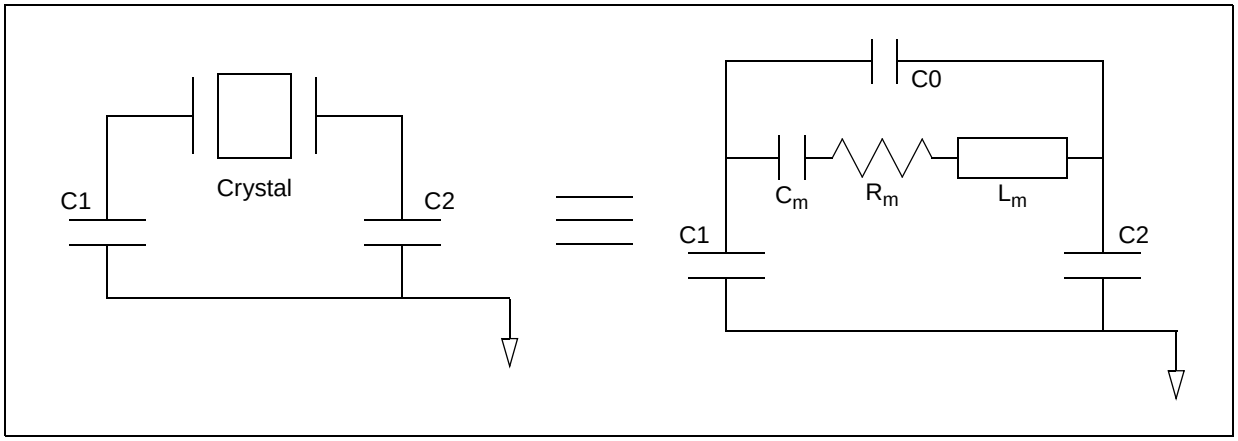


Figure 14. Equivalent circuit of a quartz crystal

Table 35. Crystal motional characteristics[1]

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
L_m	Motional inductance	—	—	11.796	—	KH
C_m	Motional capacitance	—	—	2	—	fF
C1/C2	Load capacitance at OSC32K_XTAL and OSC32K_EXTAL with respect to ground[2]	—	18	—	28	pF
R_m[3]	Motional resistance	AC coupled at C0 = 2.85 pF[4]	—	—	65	kΩ
		AC coupled at C0 = 4.9 pF[4]	—	—	50	
		AC coupled at C0 = 7.0 pF[4]	—	—	35	
		AC coupled at C0 = 9.0 pF[4]	—	—	30	

1 The crystal used is Epson Toyocom MC306.

2 This is the recommended range of load capacitance at OSC32K_XTAL and OSC32K_EXTAL with respect to ground. It includes all the parasitics due to board traces, crystal and package.

3 Maximum ESR (R_m) of the crystal is 50 kΩ

4 C0 Includes a parasitic capacitance of 2.0 pF between OSC32K_XTAL and OSC32K_EXTAL pins.

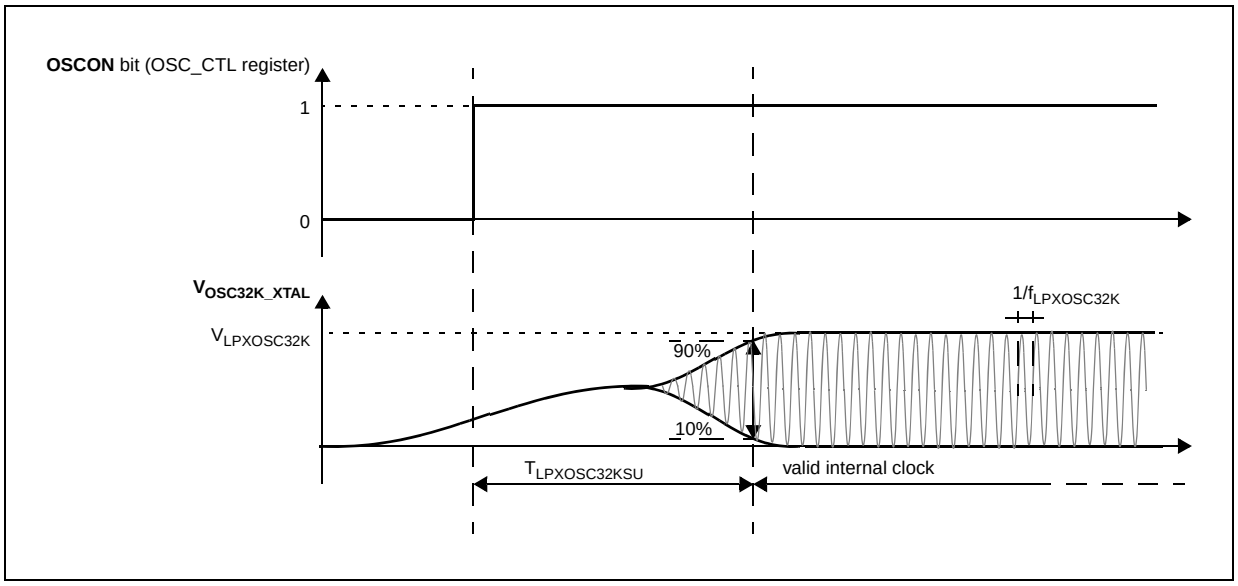


Figure 15. Slow external crystal oscillator (32 kHz) electrical characteristics

Table 36. Slow external crystal oscillator (32 kHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{SXOSC}	SR	—	Slow external crystal oscillator frequency	—	32	32.768	40	kHz
V_{SXOSC}	CC	T	Oscillation amplitude	—	—	2.1	—	V
$I_{SXOSCBIAS}$	CC	T	Oscillation bias current	—		2.5		µA
I_{SXOSC}	CC	T	Slow external crystal oscillator consumption	—	—	—	8	µA
$T_{SXOSCSU}$	CC	T	Slow external crystal oscillator start-up time	—	—	—	2[2]	s

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 Start-up time has been measured with EPSON TOYOCOM MC306 crystal. Variation may be seen with other crystal.

4.14 FMPLL electrical characteristics

The device provides a frequency modulated phase locked loop (FMPLL) module to generate a fast system clock from the main oscillator driver.

Table 37. FMPLL electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{PLLIN}	SR	—	FMPLL reference clock[2]	—	4	—	64	MHz
Δ_{PLLIN}	SR	—	FMPLL reference clock duty cycle[2]	—	40	—	60	%
f_{PLLOUT}	CC	P	FMPLL output clock frequency	—	16	—	64	MHz
f_{VCO}[3]	CC	P	VCO frequency without frequency modulation	—	256	—	512	MHz
		P	VCO frequency with frequency modulation	—	245.76	—	532.48	
f_{CPU}	SR	—	System clock frequency	—	—	—	64[4]	MHz
f_{FREE}	CC	P	Free-running frequency	—	20	—	150	MHz
t_{LOCK}	CC	P	FMPLL lock time	Stable oscillator (f_{PLLIN} = 16 MHz)		40	100	µs
Δt_{STJIT}	CC	—	FMPLL short term jitter[5]	f_{sys} maximum	–4	—	4	%
Δt_{LTJIT}	CC	—	FMPLL long term jitter	f_{PLLCLK} at 64 MHz, 4000 cycles	—	—	10	ns
I_{PLL}	CC	C	FMPLL consumption	T_A = 25 °C	—	—	4	mA

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 PLLIN clock retrieved directly from FXOSC clock. Input characteristics are granted when oscillator is used in functional mode. When bypass mode is used, oscillator input clock should verify f_{PLLIN} and Δ_{PLLIN}.

3 Frequency modulation is considered ± 4%.

4 f_{CPU} 64 MHz can be achieved only at up to 105 °C.

5 Short term jitter is measured on the clock rising edge at cycle n and n+4.

4.15 Fast internal RC oscillator (16 MHz) electrical characteristics

The device provides a 16 MHz main internal RC oscillator. This is used as the default clock at the power-up of the device.

Table 38. Fast internal RC oscillator (16 MHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{FIRC}	CC	P	Fast internal RC oscillator high frequency	T_A = 25 °C, trimmed	—	16	—	MHz
	SR	—		—	12		20	
$I_{FIRCRUN}$[2],	CC	T	Fast internal RC oscillator high frequency current in running mode	T_A = 25 °C, trimmed	—	—	200	µA
$I_{FIRCPWD}$	CC	D	Fast internal RC oscillator high frequency current in power down mode	T_A = 25 °C	—	—	10	µA

Table 38. Fast internal RC oscillator (16 MHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
$I_{FIRCSTOP}$	CC	T	Fast internal RC oscillator high frequency and system clock current in stop mode	T_A = 25 °C	sysclk = off	—	500	—	µA
					sysclk = 2 MHz	—	600	—	
					sysclk = 4 MHz	—	700	—	
					sysclk = 8 MHz	—	900	—	
					sysclk = 16 MHz	—	1250	—	
T_{FIRCSU}	CC	C	Fast internal RC oscillator start-up time	V_{DD} = 5.0 V ± 10%		—	1.1	2.0	µs
$\Delta_{FIRCPRE}$	CC	C	Fast internal RC oscillator precision after software trimming of f_{FIRC}	T_A = 25 °C		–1	—	1	%
$\Delta_{FIRCTRIM}$	CC	C	Fast internal RC oscillator trimming step	T_A = 25 °C		—	1.6		%
$\Delta_{FIRCVAR}$	CC	C	Fast internal RC oscillator variation over temperature and supply with respect to f_{FIRC} at T_A = 25 °C in high-frequency configuration	—		–5	—	5	%

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

4.16 Slow internal RC oscillator (128 kHz) electrical characteristics

The device provides a 128 kHz low power internal RC oscillator. This can be used as the reference clock for the RTC module.

Table 39. Slow internal RC oscillator (128 kHz) electrical characteristics

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
f_{SIRC}	CC	P	Slow internal RC oscillator low frequency	T_A = 25 °C, trimmed	—	128	—	kHz
	SR	—		—	100	—	150	
I_{SIRC}[2],	CC	C	Slow internal RC oscillator low frequency current	T_A = 25 °C, trimmed	—	—	5	µA
T_{SIRCSU}	CC	P	Slow internal RC oscillator start-up time	T_A = 25 °C, V_{DD} = 5.0 V ± 10%	—	8	12	µs
$\Delta_{SIRCPRE}$	CC	C	Slow internal RC oscillator precision after software trimming of f_{SIRC}	T_A = 25 °C	–2	—	2	%
$\Delta_{SIRCTRIM}$	CC	C	Slow internal RC oscillator trimming step	—	—	2.7	—	

Table 39. Slow internal RC oscillator (128 kHz) electrical characteristics (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
$\Delta_{SIRCVAR}$	CC	C	Slow internal RC oscillator variation in temperature and supply with respect to f_{SIRC} at T_A = 55 °C in high frequency configuration	High frequency configuration	–10	—	10	%

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 This does not include consumption linked to clock tree toggling and peripherals consumption when RC oscillator is ON.

4.17 ADC electrical characteristics

4.17.1 Introduction

The device provides two Successive Approximation Register (SAR) analog-to-digital converters (10-bit and 12-bit).

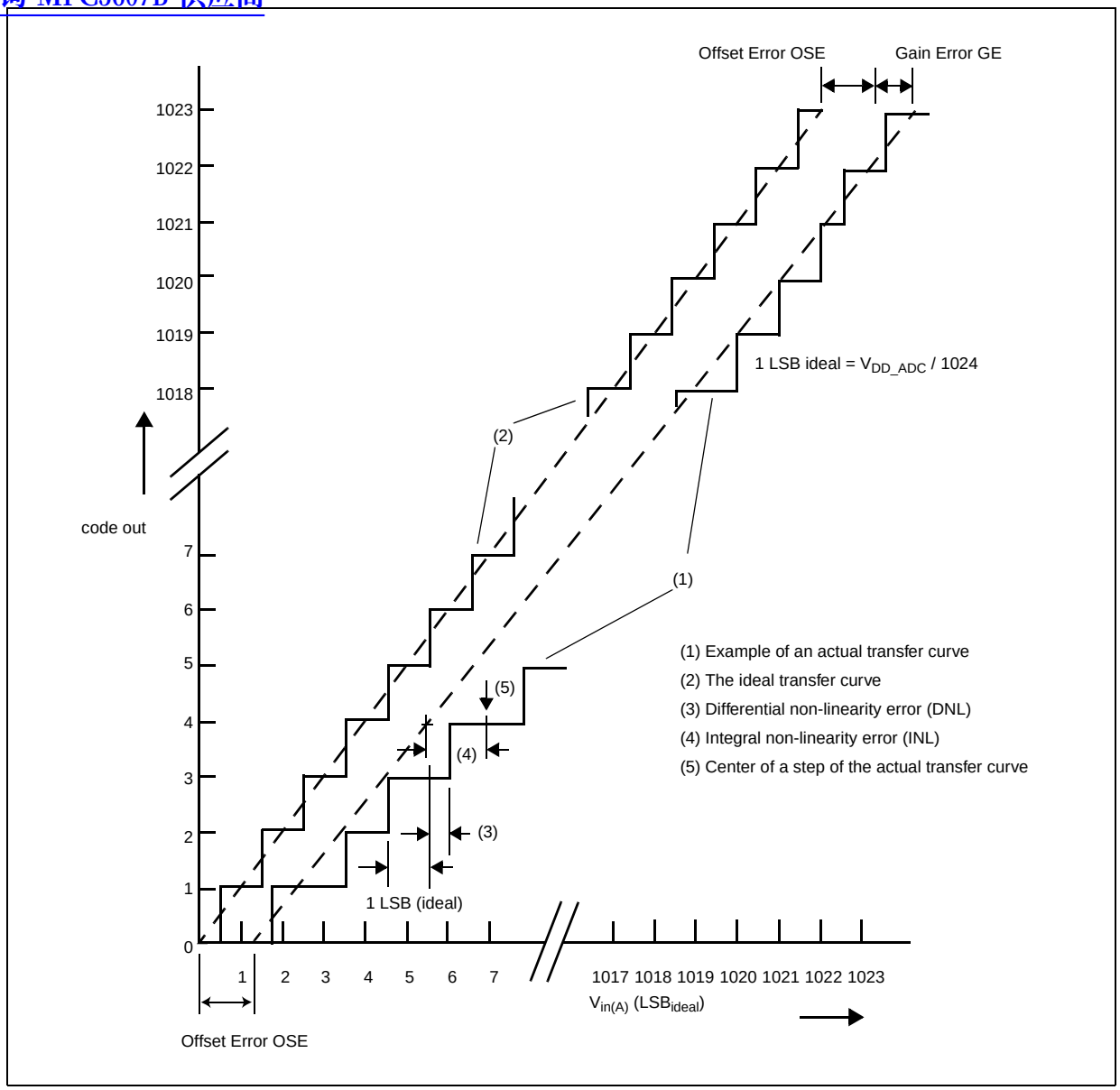


Figure 16. ADC_0 characteristic and error definitions

4.17.2 Input impedance and ADC accuracy

In the following analysis, the input circuit corresponding to the precise channels is considered.

To preserve the accuracy of the A/D converter, it is necessary that analog input pins have low AC impedance. Placing a capacitor with good high frequency characteristics at the input pin of the device can be effective: the capacitor should be as large as possible, ideally infinite. This capacitor contributes to attenuating the noise present on the input pin; furthermore, it sources charge during the sampling phase, when the analog signal source is a high-impedance source.

A real filter can typically be obtained by using a series resistance with a capacitor on the input pin (simple RC filter). The RC filtering may be limited according to the value of source impedance of the transducer or circuit supplying the analog signal to be measured. The filter at the input pins must be designed taking into account the dynamic characteristics of the input signal (bandwidth) and the equivalent input impedance of the ADC itself.

In fact a current sink contributor is represented by the charge sharing effects with the sampling capacitance: C_S being substantially a switched capacitance, with a frequency equal to the conversion rate of the ADC, it can be seen as a resistive path to ground. For instance, assuming a conversion rate of 1 MHz, with C_S equal to 3 pF, a resistance of 330 kΩ is obtained (R_{EQ} = 1 / (fc*C_S), where fc represents the conversion rate at the considered channel). To minimize the error induced by the voltage partitioning between this resistance (sampled voltage on C_S) and the sum of $R_S + R_F + R_L + R_{SW} + R_{AD}$, the external circuit must be designed to respect the Equation 4:

Eqn. 4

$$V_A \bullet \frac{R_S + R_F + R_L + R_{SW} + R_{AD}}{R_{EQ}} < \frac{1}{2} LSB$$

Equation 4 generates a constraint for external network design, in particular on resistive path. Internal switch resistances (R_{SW} and R_{AD}) can be neglected with respect to external resistances.

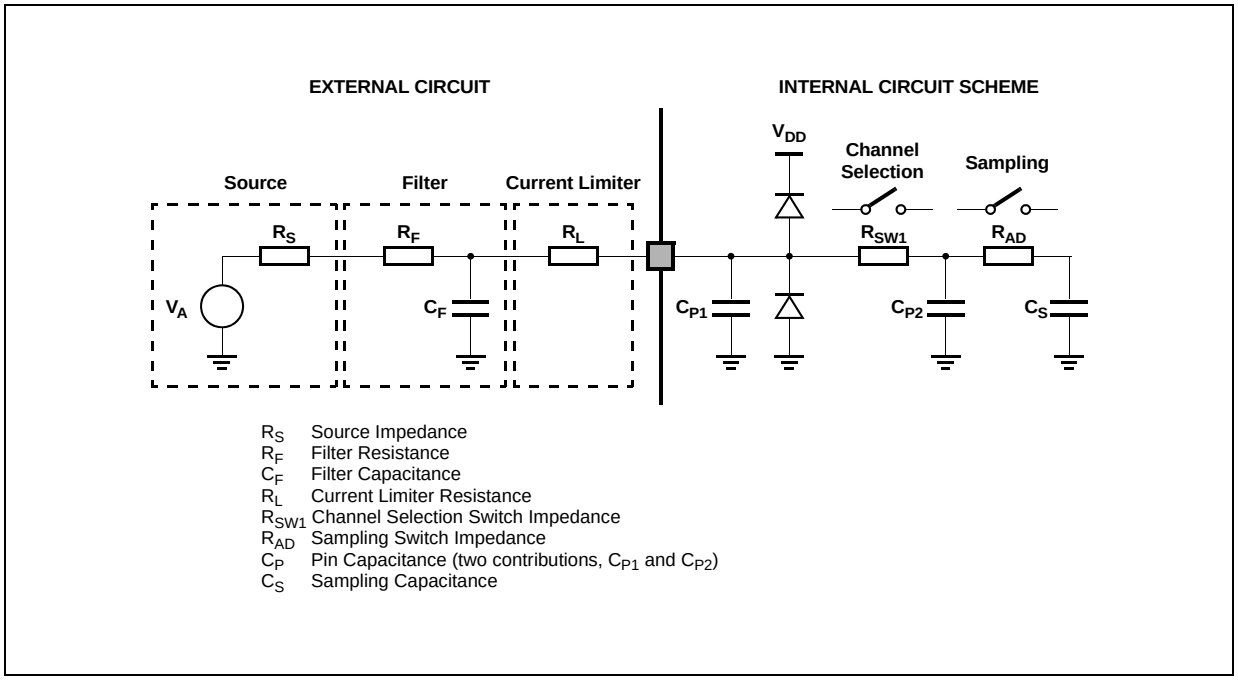


Figure 17. Input equivalent circuit (precise channels)

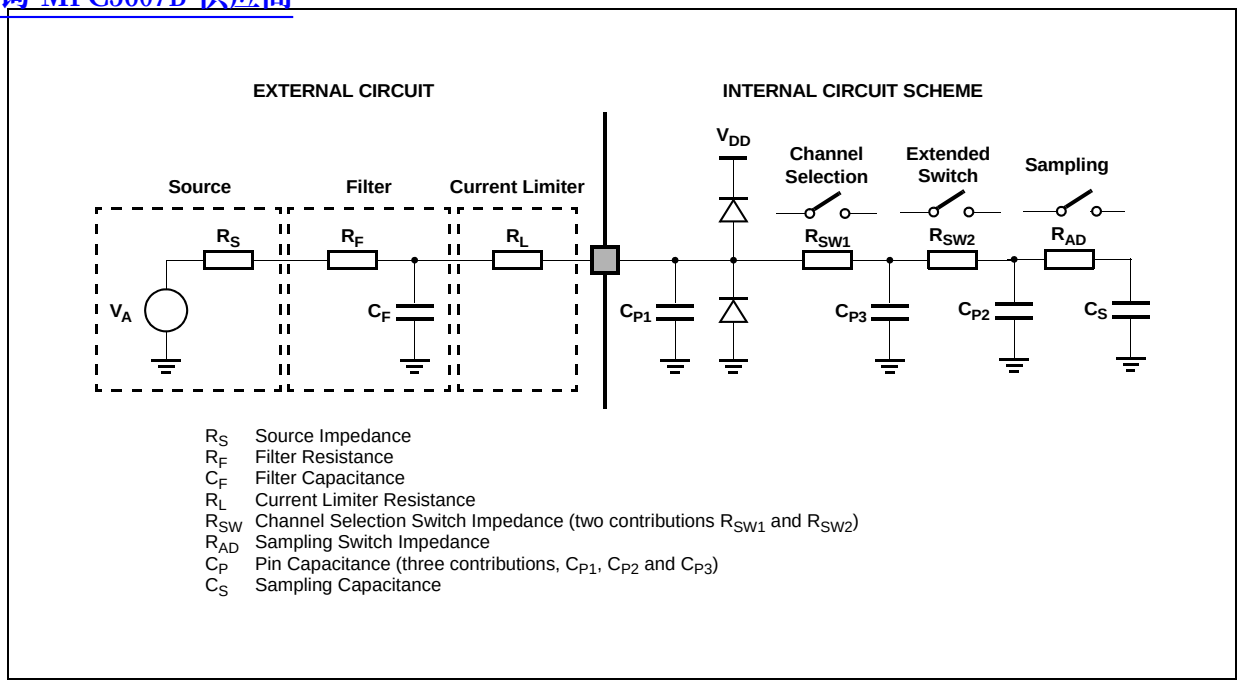


Figure 18. Input equivalent circuit (extended channels)

A second aspect involving the capacitance network shall be considered. Assuming the three capacitances C_F, C_{P1} and C_{P2} are initially charged at the source voltage V_A (refer to the equivalent circuit reported in Figure 17): A charge sharing phenomenon is installed when the sampling phase is started (A/D switch close).

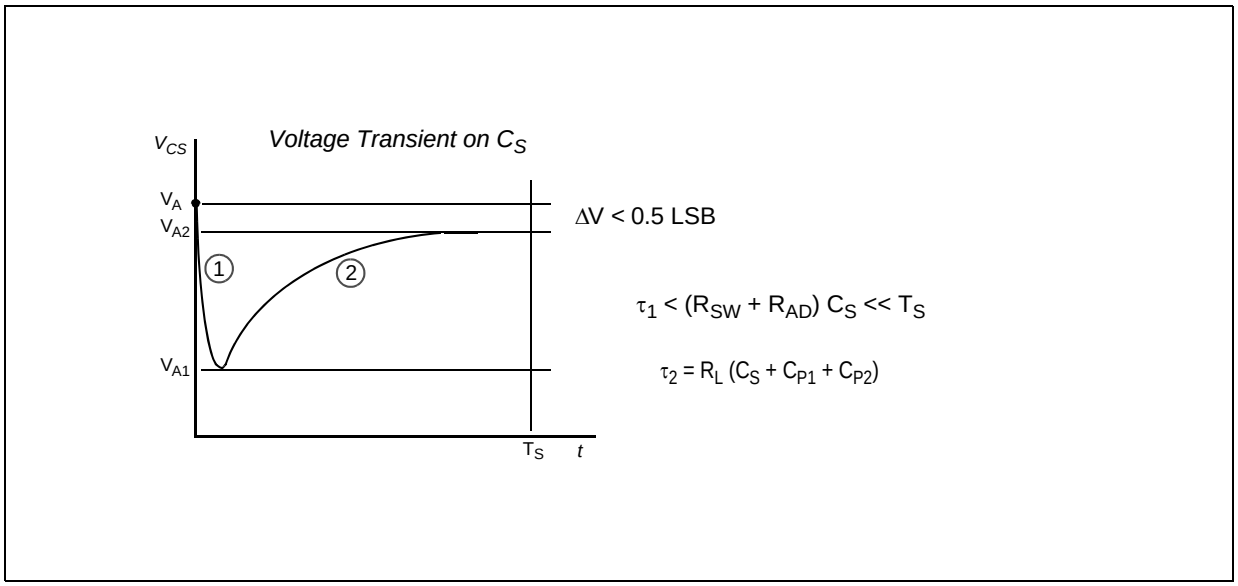


Figure 19. Transient behavior during sampling phase

In particular two different transient periods can be distinguished:

1. A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Eqn. 5

$$\tau_1 = (R_{SW} + R_{AD}) \bullet \frac{C_P \bullet C_S}{C_P + C_S}$$

Equation 5 can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time T_S is always much longer than the internal time constant:

Eqn. 6

$$\tau_1 < (R_{SW} + R_{AD}) \bullet C_S \ll T_S$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S, determining a new value of the voltage V_{A1} on the capacitance according to Equation 7:

Eqn. 7

$$V_{A1} \bullet (C_S + C_{P1} + C_{P2}) = V_A \bullet (C_{P1} + C_{P2})$$

2. A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L: again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Eqn. 8

$$\tau_2 < R_L \bullet (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time T_S, a constraints on R_L sizing is obtained:

Eqn. 9

$$10 \bullet \tau_2 = 10 \bullet R_L \bullet (C_S + C_{P1} + C_{P2}) < T_S$$

Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1}, C_{P2} and C_S, then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1}. Equation 10 must be respected (charge balance assuming now C_S already charged at V_{A1}):

Eqn. 10

$$V_{A2} \bullet (C_S + C_{P1} + C_{P2} + C_F) = V_A \bullet C_F + V_{A1} \bullet (C_{P1} + C_{P2} + C_S)$$

The two transients above are not influenced by the voltage source that, due to the presence of the $R_F C_F$ filter, is not able to provide the extra charge to compensate the voltage drop on C_S with respect to the ideal source V_A; the time constant $R_F C_F$ of the filter is very high with respect to the sampling time (T_S). The filter is typically designed to act as antialiasing.

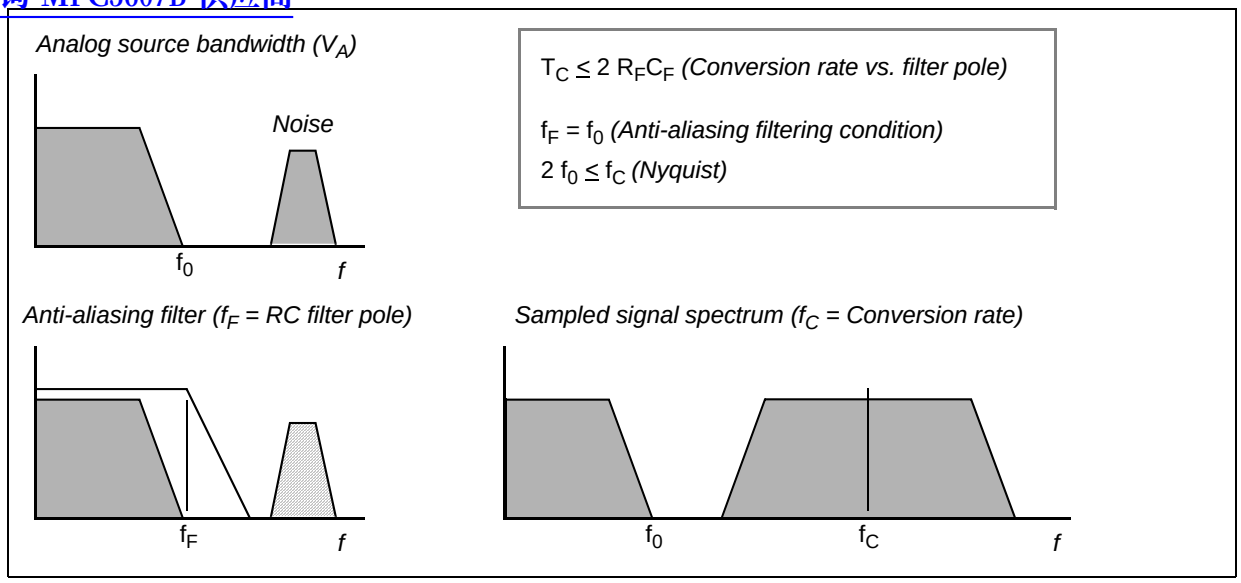


Figure 20. Spectral representation of input signal

Calling f_0 the bandwidth of the source signal (and as a consequence the cut-off frequency of the antialiasing filter, f_F), according to the Nyquist theorem the conversion rate f_C must be at least $2f_0$; it means that the constant time of the filter is greater than or at least equal to twice the conversion period (T_C). Again the conversion period T_C is longer than the sampling time T_S, which is just a portion of it, even when fixed channel continuous conversion mode is selected (fastest conversion rate at a specific channel): in conclusion it is evident that the time constant of the filter $R_F C_F$ is definitively much higher than the sampling time T_S, so the charge level on C_S cannot be modified by the analog signal source during the time in which the sampling switch is closed.

The considerations above lead to impose new constraints on the external circuit, to reduce the accuracy error due to the voltage drop on C_S; from the two charge balance equations above, it is simple to derive Equation 11 between the ideal and real sampled voltage on C_S:

Eqn. 11

$$\frac{V_A}{V_{A2}} = \frac{C_{P1} + C_{P2} + C_F}{C_{P1} + C_{P2} + C_F + C_S}$$

From this formula, in the worst case (when V_A is maximum, that is for instance 5 V), assuming to accept a maximum error of half a count, a constraint is evident on C_F value:

ADC_0 (10-bit) ***Eqn. 12***

$$C_F > 2048 \bullet C_S$$

ADC_1 (12-bit) ***Eqn. 13***

$$C_F > 8192 \bullet C_S$$

4.17.3 ADC electrical characteristics

Table 40. ADC input leakage current

Symbol		C	Parameter	Conditions		Value			Unit
						Min	Typ	Max	
I_{LKG}	CC	C	Input leakage current	T_A = –40 °C	No current injection on adjacent pin	—	1	—	nA
		C		T_A = 25 °C		—	1	—	
		C		T_A = 105 °C		—	8	200	
		P		T_A = 125 °C		—	45	400	

Table 41. ADC_0 conversion characteristics (10-bit ADC_0)

Symbol		C		Parameter	Conditions[1]	Value			Unit
						Min	Typ	Max	
V_{SS_ADC0}	SR	—		Voltage on VSS_HV_ADC0 (ADC_0 reference) pin with respect to ground (V_{SS})[2]	—	–0.1	—	0.1	V
V_{DD_ADC0}	SR	—		Voltage on VDD_HV_ADC pin (ADC reference) with respect to ground (V_{SS})	—	V_{DD} – 0.1	—	V_{DD} + 0.1	V
V_{AINx}	SR	—		Analog input voltage[3]	—	V_{SS_ADC0} – 0.1	—	V_{DD_ADC0} + 0.1	V
$I_{ADC0pwd}$	SR	—		ADC_0 consumption in power down mode	—	—	—	50	µA
$I_{ADC0run}$	SR	—		ADC_0 consumption in running mode	—	—	—	40	mA
f_{ADC0}	SR	—		ADC_0 analog frequency	—	6	—	32 + 4%	MHz
Δ_{ADC0_SYS}	SR	—		ADC_0 digital clock duty cycle (ipg_clk)	ADCLKSEL = 1[4]	45	—	55	%
t_{ADC0_PU}	SR	—		ADC_0 power up delay	—	—	—	1.5	µs
t_{ADC0_S}	CC	T		Sample time[5]	f_{ADC} = 32 MHz, ADC0_conf_sample_input = 17	0.5	—		µs
					f_{ADC} = 6 MHz, INPSAMP = 255	—	—	42	
t_{ADC0_C}	CC	P		Conversion time[6]	f_{ADC} = 32 MHz, ADC_conf_comp = 2	0.625	—	—	µs
C_S	CC	D		ADC_0 input sampling capacitance	—	—	—	3	pF
C_{P1}	CC	D		ADC_0 input pin capacitance 1	—	—	—	3	pF
C_{P2}	CC	D		ADC_0 input pin capacitance 2	—	—	—	1	pF
C_{P3}	CC	D		ADC_0 input pin capacitance 3	—	—	—	1	pF

Table 41. ADC_0 conversion characteristics (continued)(10-bit ADC_0)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
R_{SW1}	CC	D	Internal resistance of analog source	—		—	—	3	kΩ
R_{SW2}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
R_{AD}	CC	D	Internal resistance of analog source	—		—	—	2	kΩ
I_{INJ}	SR	—	Input current Injection	Current injection on one ADC_0 input, different from the converted one	V_{DD} = 3.3 V ± 10%	–5	—	5	mA
					V_{DD} = 5.0 V ± 10%	–5	—	5	
| INL |	CC	T	Absolute value for integral nonlinearity	No overload		—	0.5	1.5	LSB
| DNL |	CC	T	Absolute differential nonlinearity	No overload		—	0.5	1.0	LSB
| OFS |	CC	T	Absolute offset error	—		—	0.5	—	LSB
| GNE |	CC	T	Absolute gain error	—		—	0.6	—	LSB
TUEP	CC	P	Total unadjusted error[7] for precise channels, input only pins	Without current injection		–2	0.6	2	LSB
		T		With current injection		–3	—	3	
TUEX	CC	T	Total unadjusted error[7] for extended channel	Without current injection		–3	1	3	LSB
		T		With current injection		–4		4	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified.

2 Analog and digital V_{SS} **must** be common (to be tied together externally).

3 V_{AINx} may exceed V_{SS_ADC0} and V_{DD_ADC0} limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0x3FF.

4 Duty cycle is ensured by using system clock without prescaling. When ADCLKSEL = 0, the duty cycle is ensured by internal divider by 2.

5 During the sample time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within t_{ADC0_S}. After the end of the sample time t_{ADC0_S}, changes of the analog input voltage have no effect on the conversion result. Values for the sample clock t_{ADC0_S} depend on programming.

6 This parameter does not include the sample time t_{ADC0_S}, but only the time for determining the digital result and the time to load the result's register with the conversion result.

7 Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

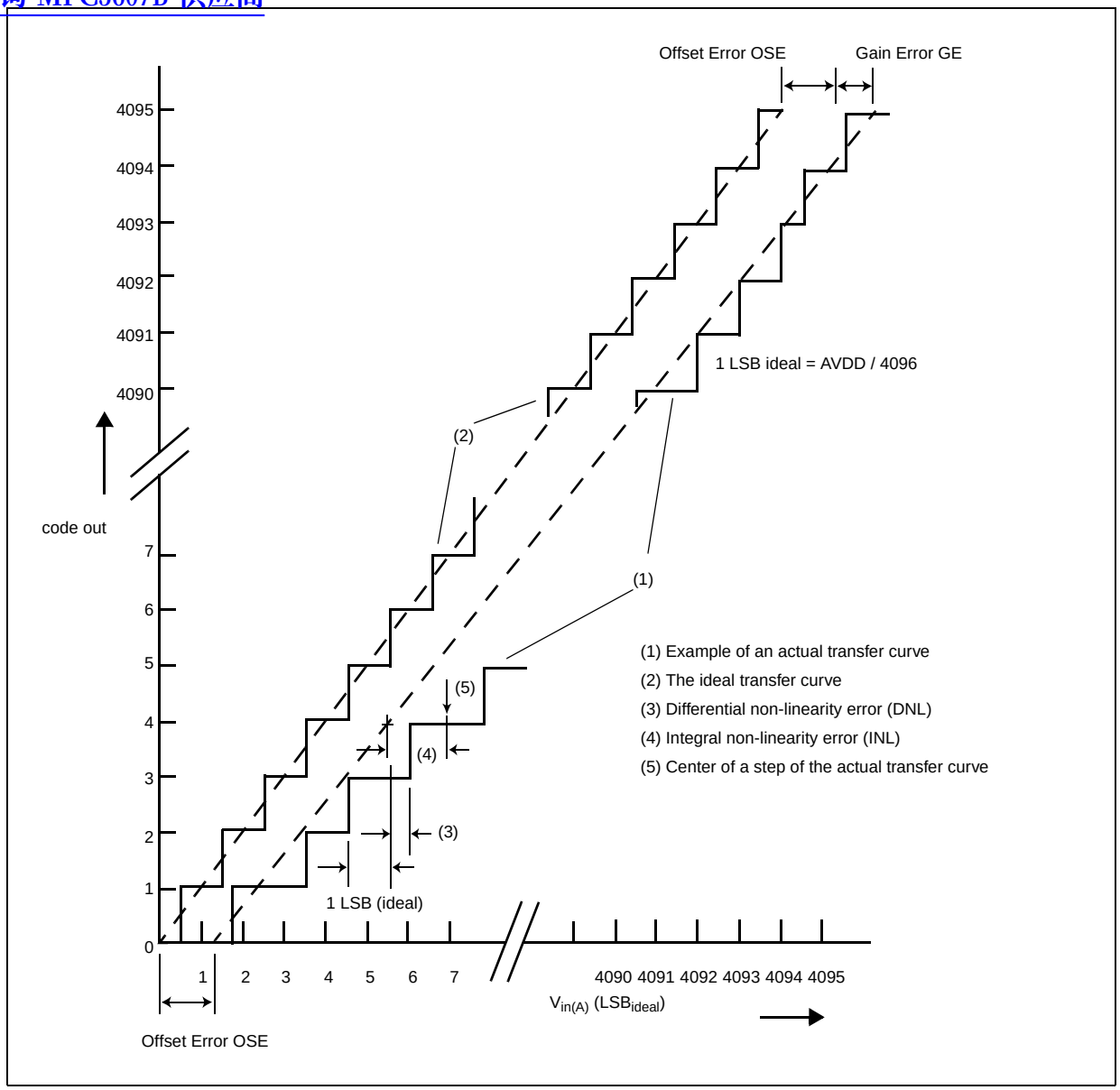


Figure 21. ADC_1 characteristic and error definitions

Table 42. ADC_1 conversion characteristics (12-bit ADC_1)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{SS_ADC1}	SR	—	Voltage on VSS_HV_ADC1 (ADC_1 reference) pin with respect to ground (V_{SS})[2]	—	–0.1	—	0.1	V
V_{DD_ADC1}	SR	—	Voltage on VDD_HV_ADC1 pin (ADC_1 reference) with respect to ground (V_{SS})	—	V_{DD} – 0.1	—	V_{DD} + 0.1	V

Table 42. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol		C	Parameter	Conditions[1]	Value			Unit
					Min	Typ	Max	
V_{AINx}	SR	—	Analog input voltage[3]	—	V_{SS_ADC1} – 0.1	—	V_{DD_ADC1} + 0.1	V
$I_{ADC1pwd}$	SR	—	ADC_1 consumption in power down mode	—	—	—	50	µA
$I_{ADC1run}$	SR	—	ADC_1 consumption in running mode	—	—	—	6	mA
f_{ADC1}	SR	—	ADC_1 analog frequency	V_{DD} = 3.3 V	3.33	—	20 + 4%	MHz
				V_{DD} = 5 V	3.33	—	32 + 4%	
t_{ADC1_PU}	SR	—	ADC_1 power up delay	—	—	—	1.5	µs
t_{ADC1_S}	CC	T	Sample time[4] VDD = 3.3 V	f_{ADC1} = 20 MHz, ADC1_conf_sample_input = 12	600	—	—	ns
			Sample time[4] VDD = 5.0 V	f_{ADC1}= 32 MHz, ADC1_conf_sample_input = 17	500	—	—	
			Sample time[4] VDD = 3.3 V	f_{ADC1}= 3.33 MHz, ADC1_conf_sample_input = 255	—	—	76.2	µs
			Sample time[4] VDD = 5.0 V	f_{ADC1}= 3.33 MHz, ADC1_conf_sample_input = 255	—	—	76.2	
t_{ADC1_C}	CC	P	Conversion time[5] VDD = 3 .3 V	f_{ADC1} = 20MHz, ADC1_conf_comp = 0	2.4	—	—	µs
			Conversion time[5] VDD = 5.0 V	$f_{ADC\ 1}$ = 32 MHz, ADC1_conf_comp = 0	1.5	—	—	µs
			Conversion time[5] VDD = 3.3 V	$f_{ADC\ 1}$ = 13.33 MHz, ADC1_conf_comp = 0	—	—	3.6	µs
			Conversion time[5] VDD = 5.0 V	f_{ADC1} = 13.33 MHz, ADC1_conf_comp = 0	—	—	3.6	µs
Δ_{ADC1_SYS}	SR	—	ADC_1 digital clock duty cycle	ADCLKSEL = 1[6]	45	—	55	%
C_S	CC	D	ADC_1 input sampling capacitance	—	—	—	5	pF
C_{P1}	CC	D	ADC_1 input pin capacitance 1	—	—	—	3	pF
C_{P2}	CC	D	ADC_1 input pin capacitance 2	—	—	—	1	pF
C_{P3}	CC	D	ADC_1 input pin capacitance 3	—	—	—	1.5	pF
R_{SW1}	CC	D	Internal resistance of analog source	—	—	—	1	kΩ
R_{SW2}	CC	D	Internal resistance of analog source	—	—	—	2	kΩ
R_{AD}	CC	D	Internal resistance of analog source	—	—	—	0.3	kΩ

Table 42. ADC_1 conversion characteristics (12-bit ADC_1) (continued)

Symbol		C	Parameter	Conditions[1]		Value			Unit
						Min	Typ	Max	
I_{INJ}	SR	—	Input current Injection	Current injection on one ADC_1 input, different from the converted one	V_{DD} = 3.3 V ± 10%	–5	—	5	mA
					V_{DD} = 5.0 V ± 10%	–5	—	5	
INLP	CC	T	Absolute Integral non-linearity-Precise channels	No overload		—	1	3	LSB
INLX	CC	T	Absolute Integral non-linearity-Extended channels	No overload		—	1.5	5	LSB
DNL	CC	T	Absolute Differential non-linearity	No overload		—	0.5	1	LSB
OFS	CC	T	Absolute Offset error	—		—	2	—	LSB
GNE	CC	T	Absolute Gain error	—		—	2	—	LSB
TUEP[7]	CC	P	Total Unadjusted Error for precise channels, input only pins	Without current injection		–6	—	6	LSB
		T		With current injection		–8	—	8	
TUEX[7]	CC	T	Total Unadjusted Error for extended channel	Without current injection		–10	—	10	LSB
		T		With current injection		–12	—	12	

1 V_{DD} = 3.3 V ± 10% / 5.0 V ± 10%, T_A = –40 to 125 °C, unless otherwise specified

2 Analog and digital V_{SS} **must** be common (to be tied together externally).

3 V_{AINx} may exceed V_{SS_ADC1} and V_{DD_ADC1} limits, remaining on absolute maximum ratings, but the results of the conversion will be clamped respectively to 0x000 or 0xFFF.

4 During the sample time the input capacitance C_S can be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach its final voltage level within t_{ADC1_S}. After the end of the sample time t_{ADC1_S}, changes of the analog input voltage have no effect on the conversion result. Values for the sample clock t_{ADC1_S} depend on programming.

5 This parameter does not include the sample time t_{ADC1_S}, but only the time for determining the digital result and the time to load the result's register with the conversion result.

6 Duty cycle is ensured by using system clock without prescaling. When ADCLKSEL = 0, the duty cycle is ensured by internal divider by 2.

7 Total Unadjusted Error: The maximum error that occurs without adjusting Offset and Gain errors. This error is a combination of Offset, Gain and Integral Linearity errors.

4.18 On-chip peripherals

4.18.1 Current consumption

Table 43. On-chip peripherals current consumption[1]

Symbol		C	Parameter	Conditions		Value (Typ)	Unit
$I_{DD\_BV(CAN)}$	CC	T	CAN (FlexCAN) supply current on V_{DD_BV}	Bitrate = 500 Kbps	Total (static + dynamic) consumption: • FlexCAN in loop-back mode • XTAL at 8 MHz used as CAN engine clock source • Message sending period is 580 µs	8 * f_{periph} + 85	µA
				Bitrate = 125 Kbps		8 * f_{periph} + 27	
$I_{DD\_BV(eMIOS)}$	CC	T	eMIOS supply current on V_{DD_BV}	Static consumption: • eMIOS channel OFF • Global prescaler enabled		29 * f_{periph}	
				Dynamic consumption: • It does not change varying the frequency (0.003 mA)		3	
$I_{DD\_BV(SCI)}$	CC	T	SCI (LINFlex) supply current on V_{DD_BV}	Total (static + dynamic) consumption: • LIN mode • Baudrate: 20 Kbps		5 * f_{periph} + 31	
$I_{DD\_BV(SPI)}$	CC	T	SPI (DSPI) supply current on V_{DD_BV}	Ballast static consumption (only clocked)		1	
				Ballast dynamic consumption (continuous communication): • Baudrate: 2 Mbit • Trasmission every 8 µs • Frame: 16 bits		16 * f_{periph}	
I_{DD_BV} (ADC_0/ADC_1)	CC	T	ADC_0/ADC_1 supply current on V_{DD_BV}	V_{DD} = 5.5 V	Ballast static consumption (no conversion)	41 * f_{periph}	µA
				V_{DD} = 5.5 V	Ballast dynamic consumption (continuous conversion)	46 * f_{periph}	
$I_{DD\_HV\_ADC0}$	CC	T	ADC_0 supply current on $V_{DD\_HV\_ADC0}$	V_{DD} = 5.5 V	Analog static consumption (no conversion)	200	
				V_{DD} = 5.5 V	Analog dynamic consumption (continuous conversion)	3	mA

Table 43. On-chip peripherals current consumption[1] (continued)

Symbol		C	Parameter	Conditions		Value	Unit
						Typ	
$I_{DD\_HV\_ADC1}$	CC	T	ADC_1 supply current on $V_{DD\_HV\_ADC1}$	V_{DD} = 5.5 V	Analog static consumption (no conversion)	300 * f_{periph}	µA
				V_{DD} = 5.5 V	Analog dynamic consumption (continuous conversion)	4	mA
$I_{DD\_HV(FLASH)}$	CC	T	CFlash + DFlash supply current on V_{DD_HV}	V_{DD} = 5.5 V	—	12	mA
$I_{DD\_BV(PLL)}$	CC	T	PLL supply current on V_{DD_BV}	V_{DD} = 5.5 V	—	2.5	mA

1 Operating conditions: T_A = 25 °C, f_{periph} = 8 MHz to 64 MHz

4.18.2 DSPI characteristics

Table 44. DSPI characteristics[1]

No.	Symbol		C	Parameter		DSPI0/DSPI1/DSPI5/DSPI6			DSPI2/DSPI4			Unit
						Min	Typ	Max	Min	Typ	Max	
1	t_{SCK}	SR	D	SCK cycle time	Master mode (MTFE = 0)	125	—	—	333	—	—	ns
			D		Slave mode (MTFE = 0)	125	—	—	333	—	—	
			D		Master mode (MTFE = 1)	83	—	—	125	—	—	
			D		Slave mode (MTFE = 1)	83	—	—	125	—	—	
—	f_{DSPI}	SR	D	DSPI digital controller frequency		—	—	f_{CPU}	—	—	f_{CPU}	MHz
—	Δt_{CSC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1->0	Master mode	—	—	130[2]	—	—	15[3]	ns
—	Δt_{ASC}	CC	D	Internal delay between pad associated to SCK and pad associated to CSn in master mode for CSn1->1	Master mode	—	—	130[3]	—	—	130[3]	ns
2	t_{CSCext}[4]	SR	D	CS to SCK delay	Slave mode	32	—	—	32	—	—	ns
3	t_{ASCext}[5]	SR	D	After SCK delay	Slave mode	$1/f_{DSPI} + 5$	—	—	$1/f_{DSPI} + 5$	—	—	ns
4	t_{SDC}	CC	D	SCK duty cycle	Master mode	—	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	ns
		SR	D		Slave mode	$t_{SCK}/2$	—	—	$t_{SCK}/2$	—	—	
5	t_A	SR	D	Slave access time	Slave mode	—	—	$1/f_{DSPI} + 70$	—	—	$1/f_{DSPI} + 130$	ns
6	t_{DI}	SR	D	Slave SOUT disable time	Slave mode	7	—	—	7	—	—	ns
9	t_{SUI}	SR	D	Data setup time for inputs	Master mode	43	—	—	145	—	—	ns
					Slave mode	5	—	—	5	—	—	
10	t_{HI}	SR	D	Data hold time for inputs	Master mode	0	—	—	0	—	—	ns
					Slave mode	2[6]	—	—	2[6]	—	—	

Table 44. DSPI characteristics[1] (continued)

No.	Symbol		C	Parameter		DSPI0/DSPI1/DSPI5/DSPI6			DSPI2/DSPI4			Unit
						Min	Typ	Max	Min	Typ	Max	
11	t_{SUO}[7]	CC	D	Data valid after SCK edge	Master mode	—	—	32	—	—	50	ns
					Slave mode	—	—	52	—	—	160	
12	t_{HO}[7]	CC	D	Data hold time for outputs	Master mode	0	—	—	0	—	—	ns
					Slave mode	8	—	—	13	—	—	

1 Operating conditions: C_{out} = 10 to 50 pF, $Slew_{IN}$ = 3.5 to 15 ns

2 Maximum value is reached when CSn pad is configured as SLOW pad while SCK pad is configured as MEDIUM. A positive value means that SCK starts before CSn is asserted. DSPI2 has only SLOW SCK available.

3 Maximum value is reached when CSn pad is configured as MEDIUM pad while SCK pad is configured as SLOW. A positive value means that CSn is deasserted before SCK. DSPI0 and DSPI1 have only MEDIUM SCK available.

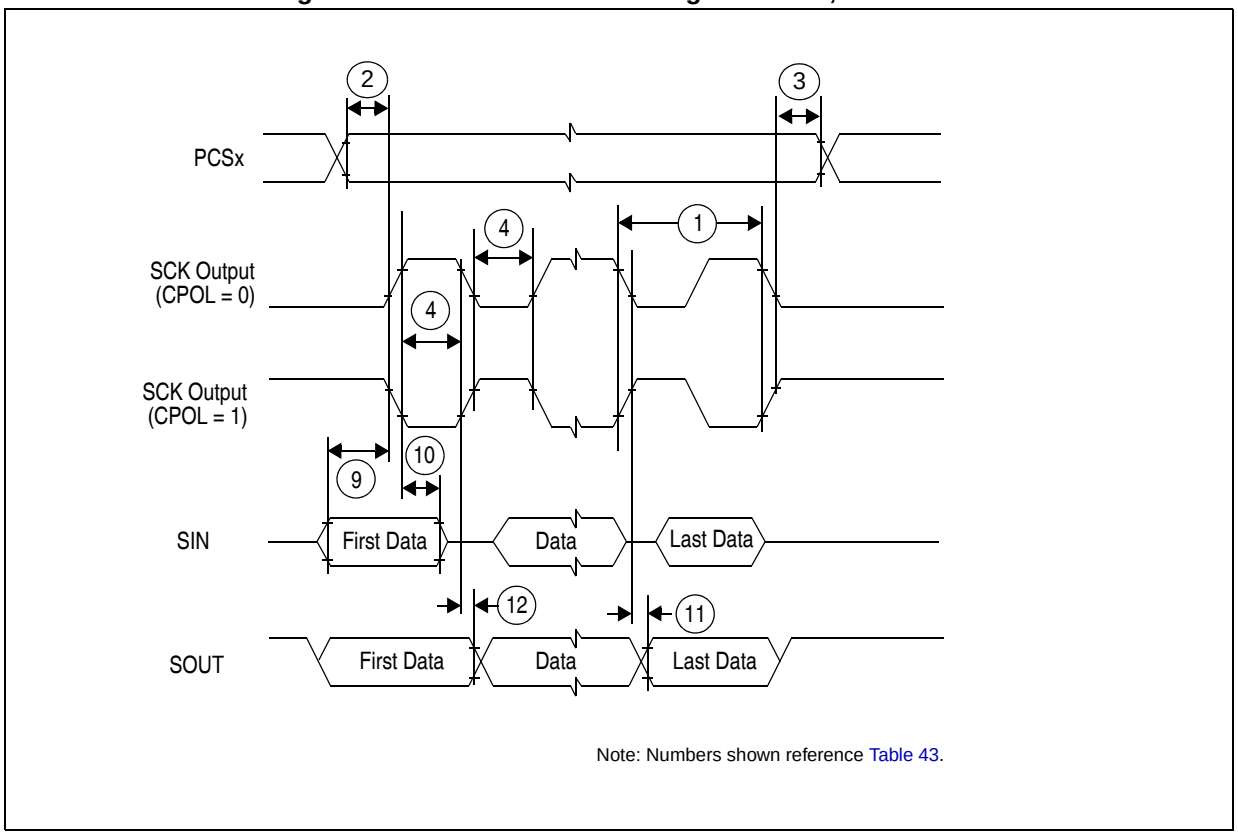
4 The t_{CSC} delay value is configurable through a register. When configuring t_{CSC} (using PCSSCK and CSSCK fields in DSPI_CTARx registers), delay between internal CS and internal SCK must be higher than Δt_{CSC} to ensure positive t_{CSCext}.

5 The t_{ASC} delay value is configurable through a register. When configuring t_{ASC} (using PASC and ASC fields in DSPI_CTARx registers), delay between internal CS and internal SCK must be higher than Δt_{ASC} to ensure positive t_{ASCext}.

6 This delay value corresponds to SMPL_PT = 00b which is bit field 9 and 8 of DSPI_MCR register.

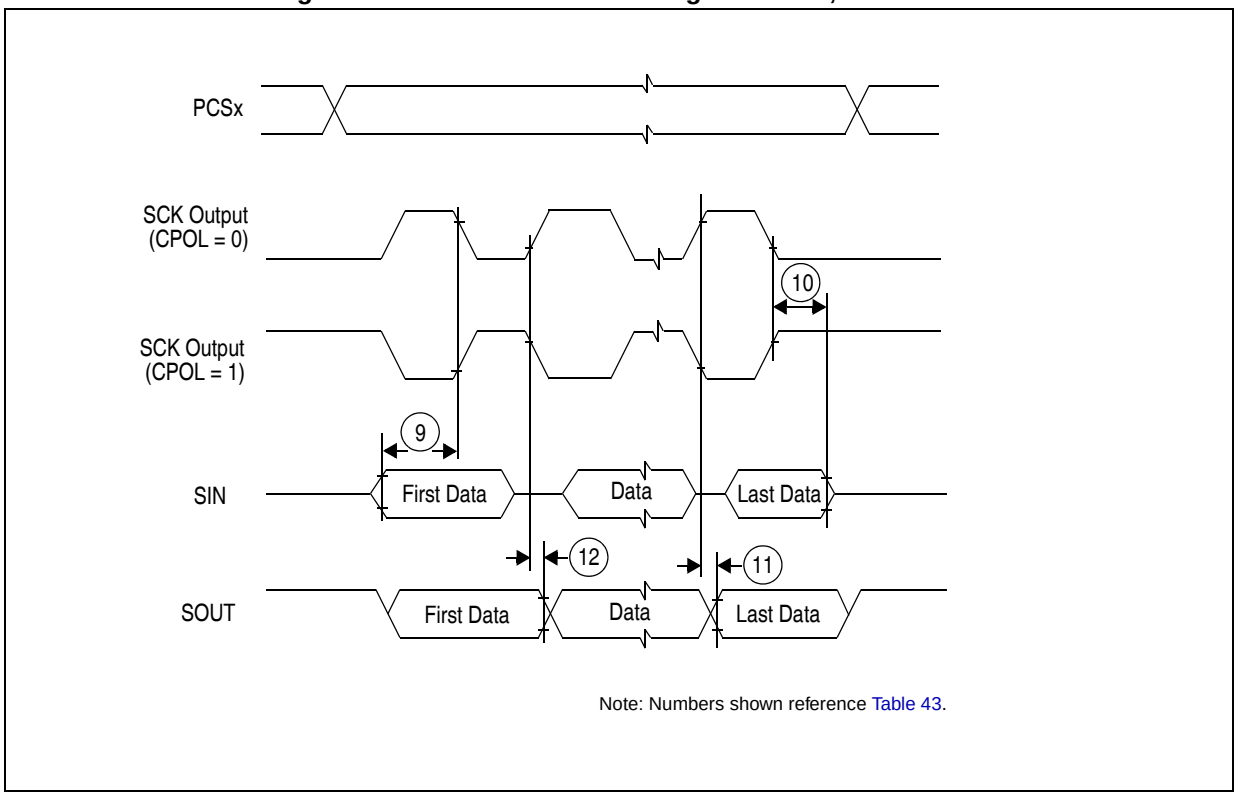
7 SCK and SOUT are configured as MEDIUM pad.

Figure 22. DSPI classic SPI timing — master, CPHA = 0



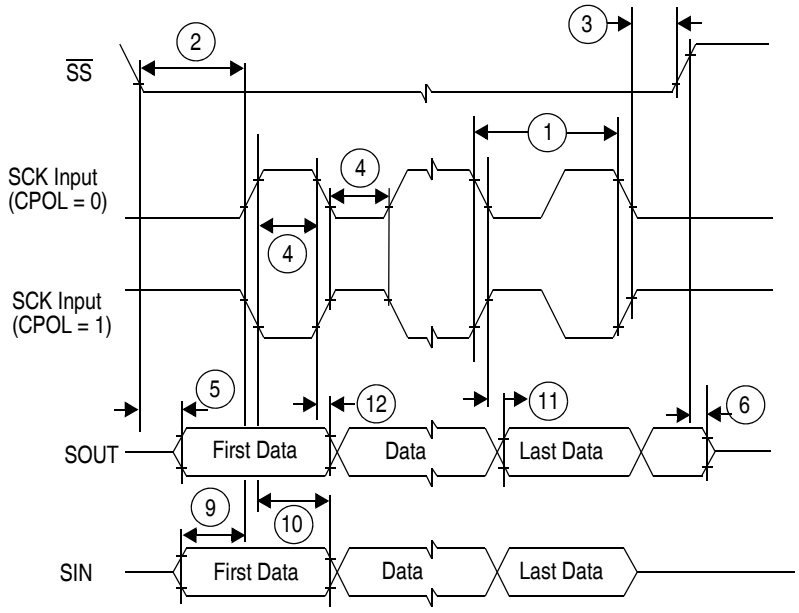
Note: Numbers shown reference Table 43.

Figure 23. DSPI classic SPI timing — master, CPHA = 1



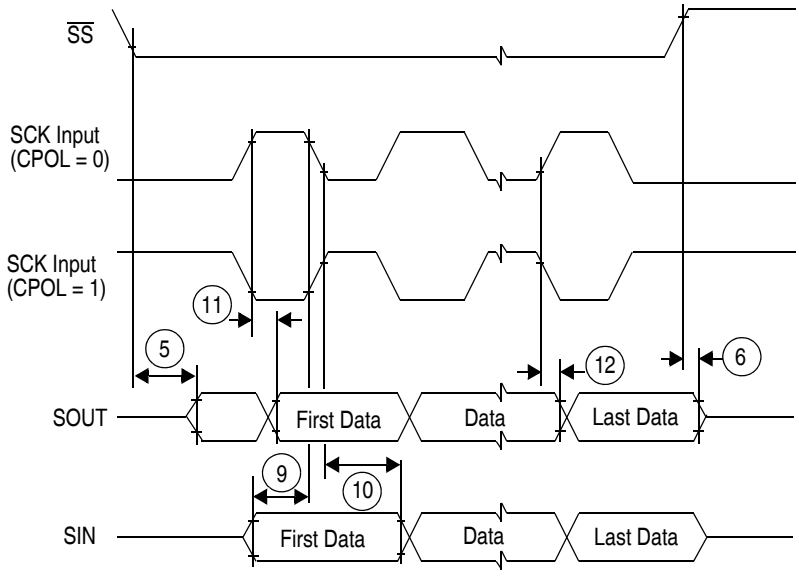
Note: Numbers shown reference Table 43.

Figure 24. DSPI classic SPI timing — slave, CPHA = 0



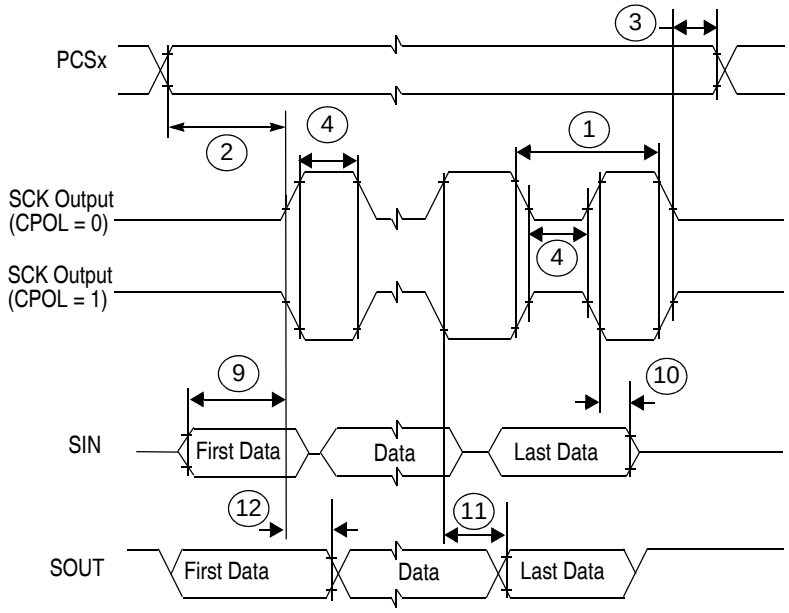
Note: Numbers shown reference Table 43.

Figure 25. DSPI classic SPI timing — slave, CPHA = 1



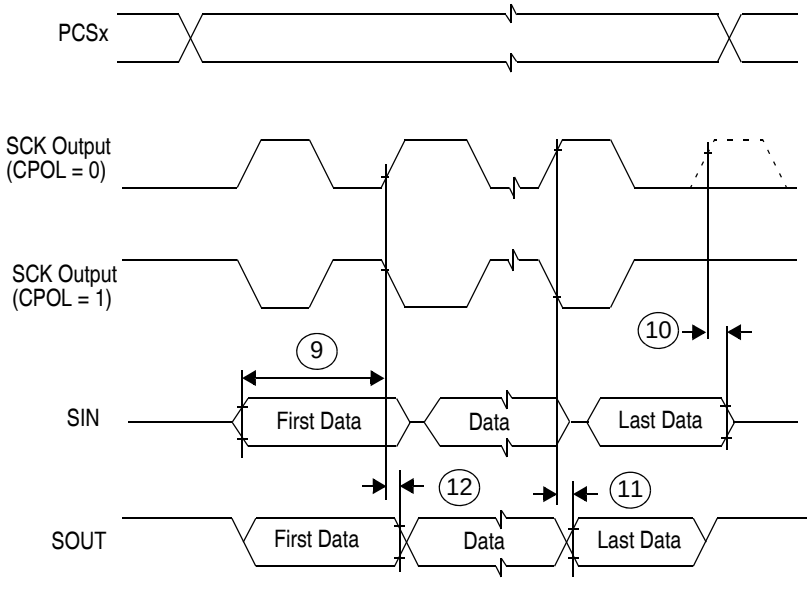
Note: Numbers shown reference Table 43.

Figure 26. DSPI modified transfer format timing — master, CPHA = 0



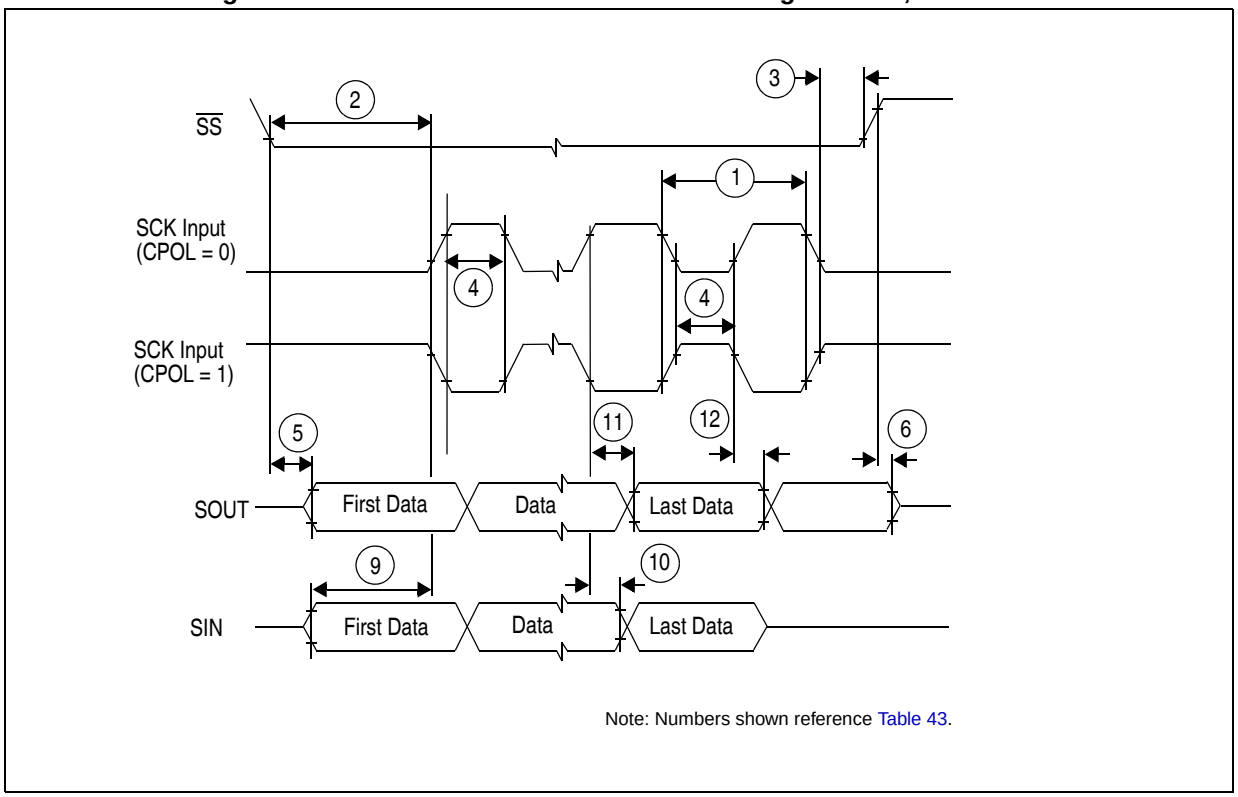
Note: Numbers shown reference Table 43.

Figure 27. DSPI modified transfer format timing — master, CPHA = 1



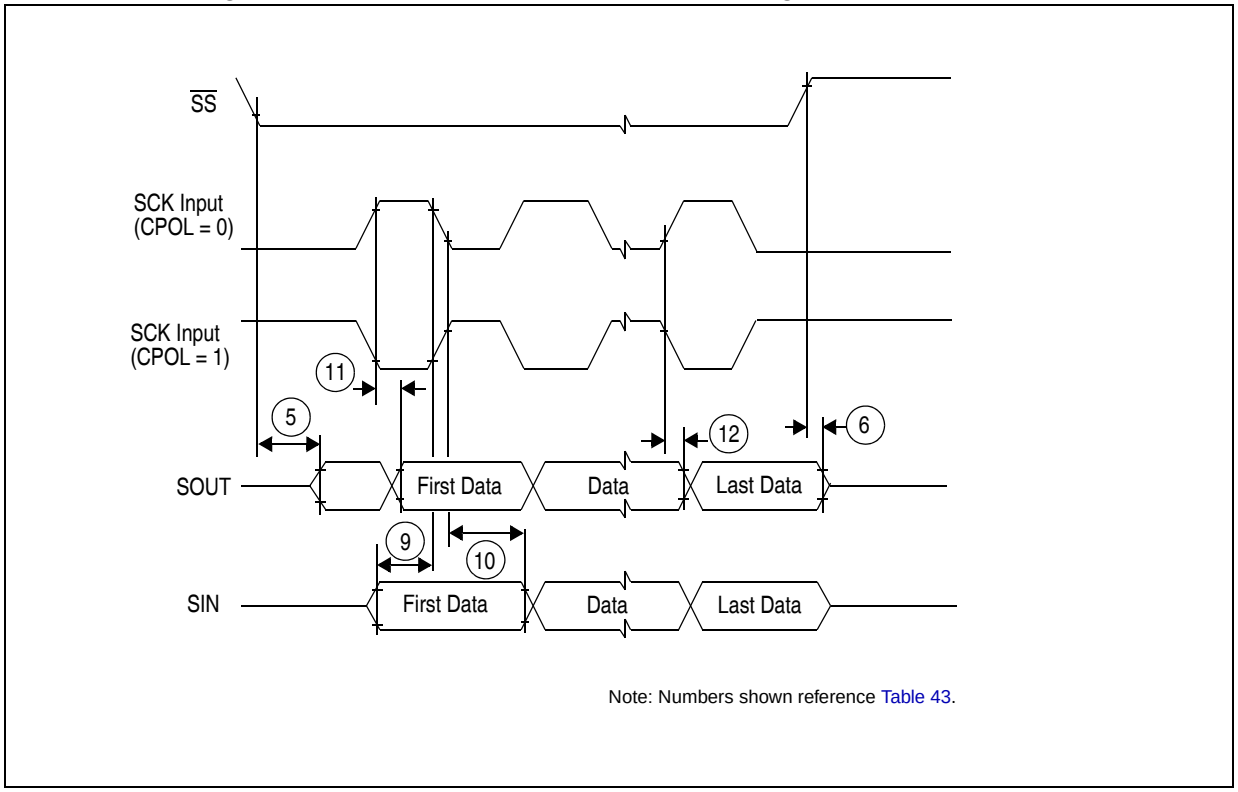
Note: Numbers shown reference Table 43.

Figure 28. DSPI modified transfer format timing — slave, CPHA = 0

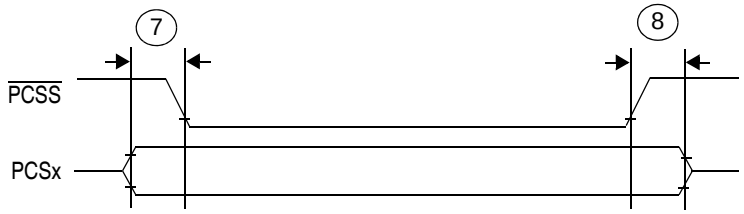


Note: Numbers shown reference Table 43.

Figure 29. DSPI modified transfer format timing — slave, CPHA = 1



Note: Numbers shown reference Table 43.



Note: Numbers shown reference Table 43.

Figure 30. DSPI PCS strobe ($\overline{\text{PCSS}}$) timing

4.18.3 Nexus characteristics

Table 45. Nexus characteristics

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t_{TCYC}	CC	D	TCK cycle time	64	—	—	ns
2	t_{MCYC}	CC	D	MCKO cycle time	32	—	—	ns
3	t_{MDOV}	CC	D	MCKO low to MDO data valid	—	—	8	ns
4	t_{MSEOV}	CC	D	MCKO low to MSEO_b data valid	—	—	8	ns
5	t_{EVTOV}	CC	D	MCKO low to EVTO data valid	—	—	8	ns
6	t_{NTDIS}	CC	D	TDI data setup time	15	—	—	ns
	t_{NTMSS}	CC	D	TMS data setup time	15	—	—	ns
7	t_{NTDIH}	CC	D	TDI data hold time	5	—	—	ns
	t_{NTMSH}	CC	D	TMS data hold time	5	—	—	ns
8	t_{TDOV}	CC	D	TCK low to TDO data valid	35	—	—	ns
9	t_{TDOI}	CC	D	TCK low to TDO data invalid	6	—	—	ns

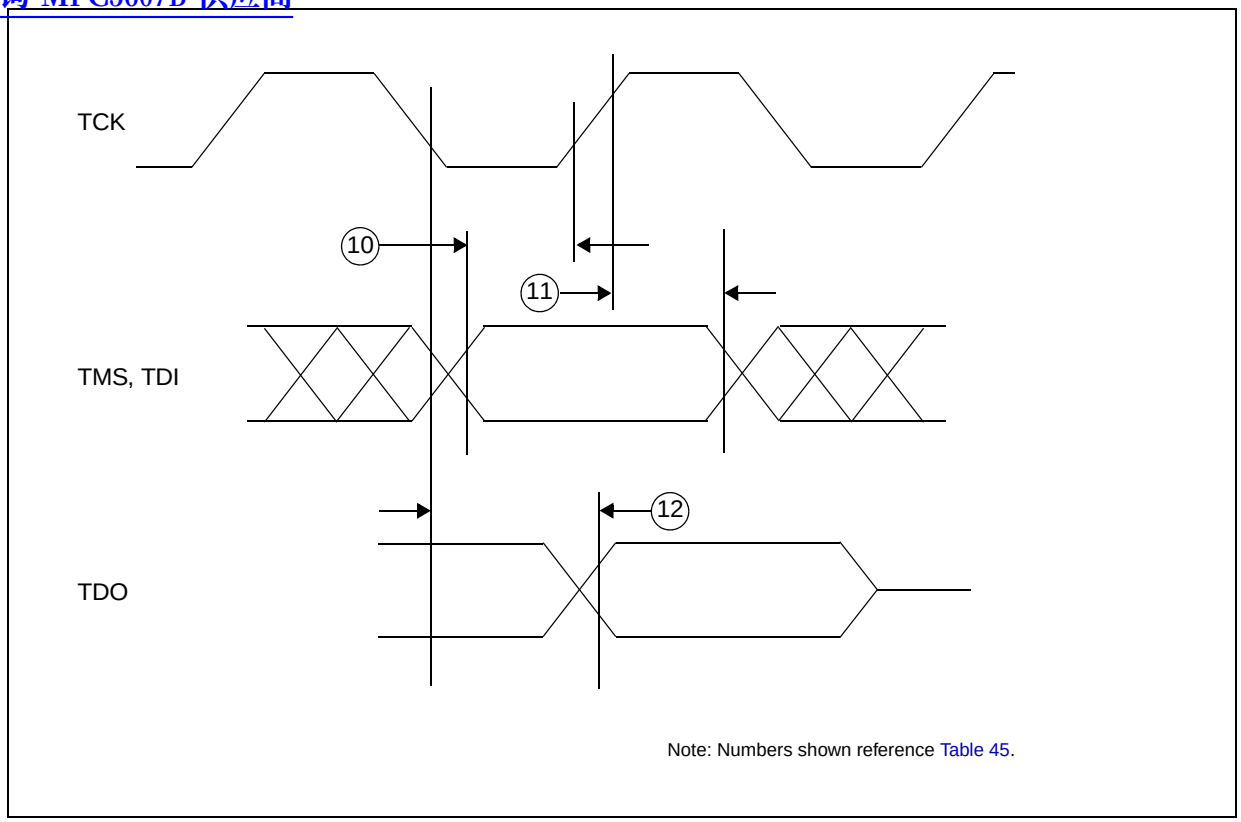


Figure 31. Nexus TDI, TMS, TDO timing

4.18.4 JTAG characteristics

Table 46. JTAG characteristics

No.	Symbol		C	Parameter	Value			Unit
					Min	Typ	Max	
1	t_{JCYC}	CC	D	TCK cycle time	64	—	—	ns
2	t_{TDIS}	CC	D	TDI setup time	15	—	—	ns
3	t_{TDIH}	CC	D	TDI hold time	5	—	—	ns
4	t_{TMSS}	CC	D	TMS setup time	15	—	—	ns
5	t_{TMSH}	CC	D	TMS hold time	5	—	—	ns
6	t_{TDOV}	CC	D	TCK low to TDO valid	—	—	33	ns
7	t_{TDOI}	CC	D	TCK low to TDO invalid	6	—	—	ns

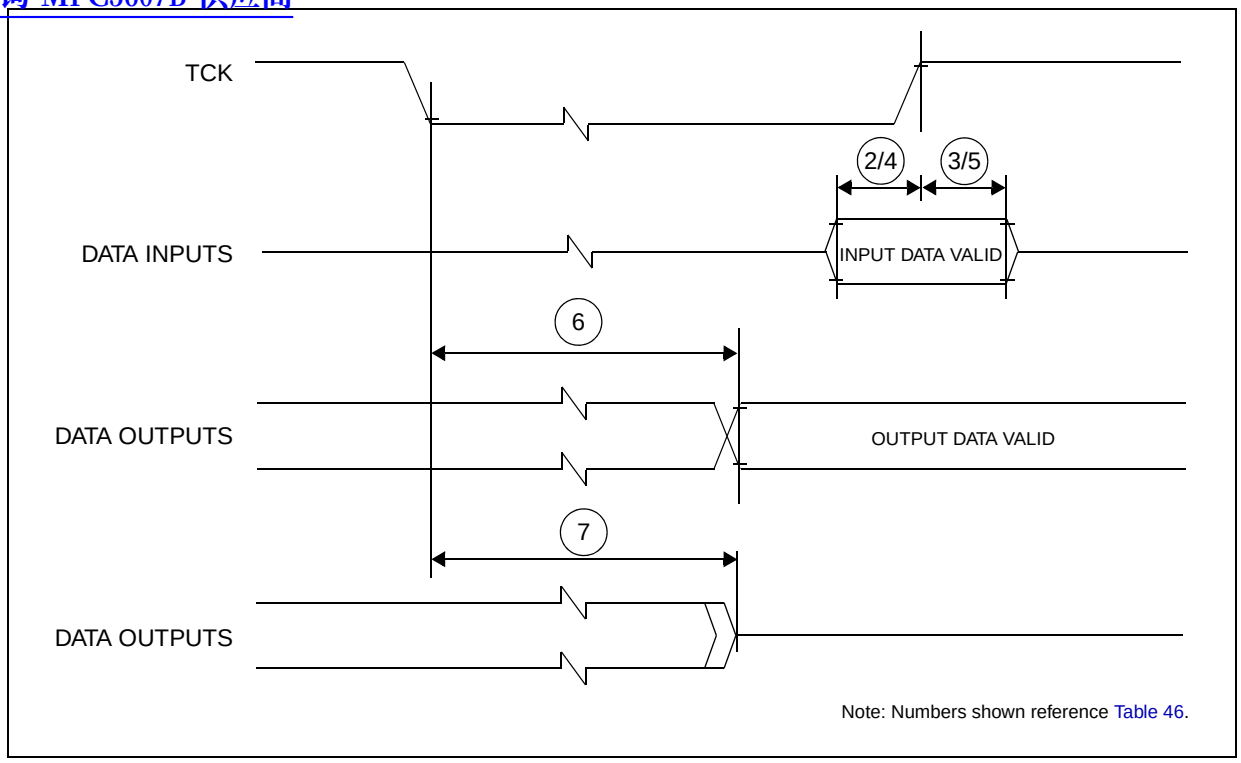


Figure 32. Timing diagram — JTAG boundary scan

5 Package characteristics

5.1 Package mechanical data

5.1.1 176 LQFP

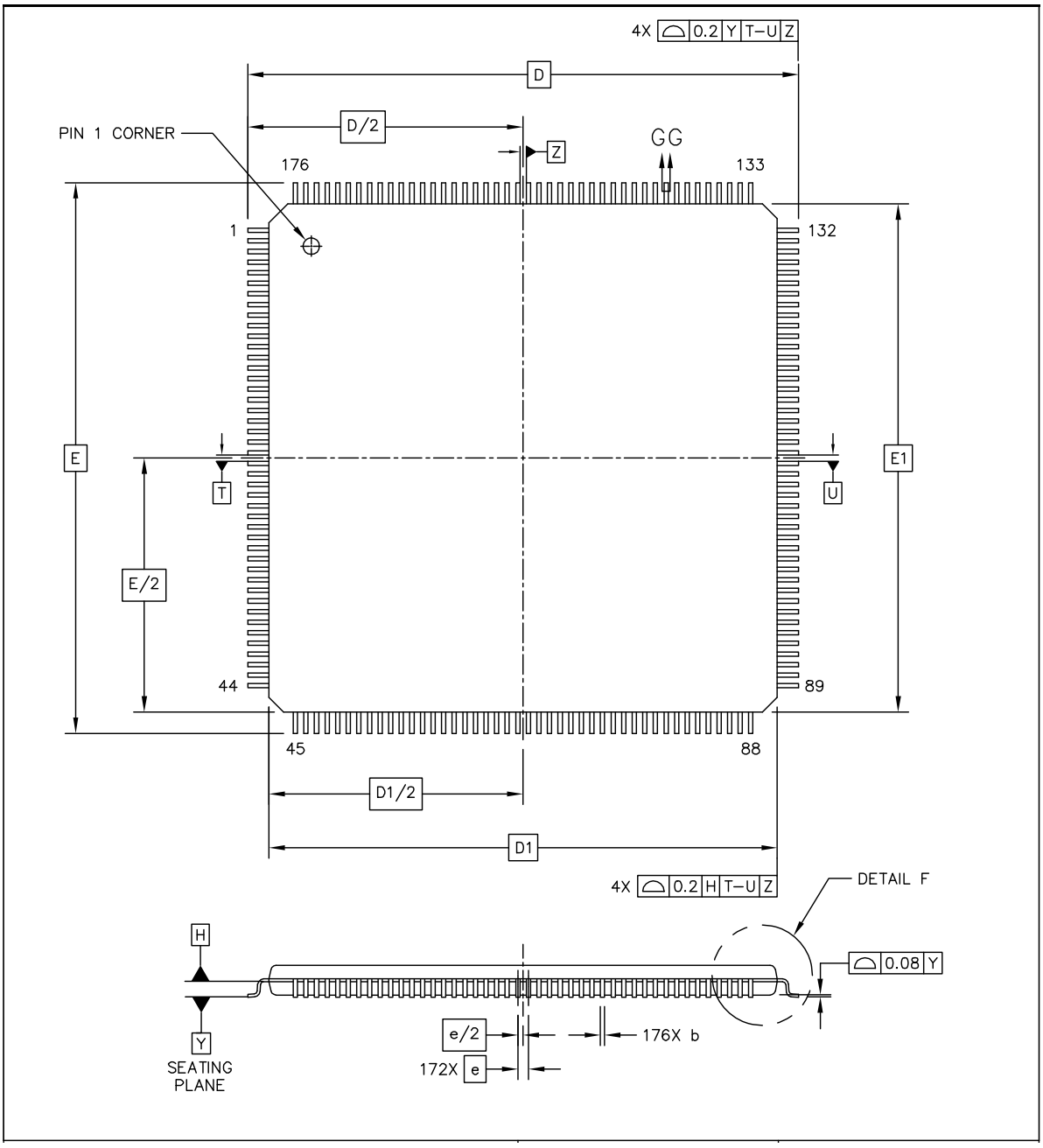


Figure 33. 176 LQFP package mechanical drawing (Part 1 of 3)

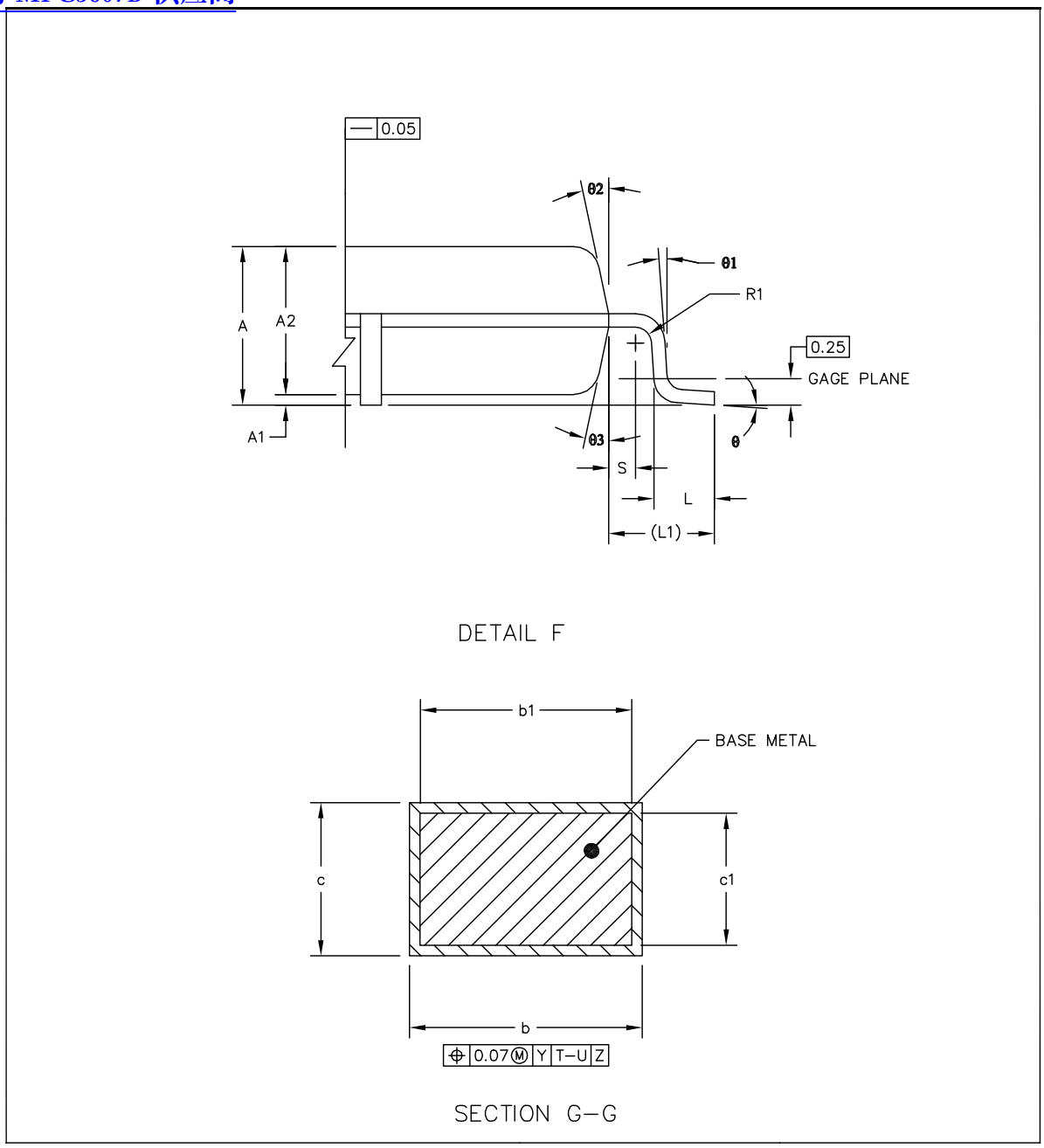


Figure 34. 176 LQFP package mechanical drawing (Part 2 of 3)

NOTES:

1. DIMENSIONS D1 AND E1 DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25MM PER SIDE. DIMENSIONS D1 AND E1 DO INCLUDE MOLD MISMATCH AND ARE DETERMINED AT DATUM PLANE DATUM H.

2. DIMENSION b DOES NOT INCLUDE DAMBAR PROTRUSION. DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED THE MAXIMUM b DIMENSION BY MORE THEN 0.08MM. DAMBAR CAN NOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM BETWEEN PROTRUSION AND AN ADJACENT LEAD IS 0.07MM FOR 0.4MM AND 0.5MM PITCH PACKAGES.

DIM	MIN	NOM	MAX
A	---		1.6
A1	0.05		0.15
A2	1.35	1.4	1.45
b	0.17	0.22	0.27
b1	0.17	0.2	0.23
c	0.09		0.2
c1	0.09		0.16
D		26 BSC	
D1		24 BSC	
e		0.5 BSC	
E		26 BSC	
E1		24 BSC	
L	0.45	0.6	0.75

DIM	MIN	NOM	MAX
L1		1 REF	
R1	0.08		---
R2	0.08		0.2
S		0.2 REF	
θ	0°	3.5°	7°
θ1	0°		---
θ2	11°	12°	13°
θ3	11°	12°	13°

DIM	MIN	NOM	MAX

UNIT	DIMENSION AND TOLERANCES	REFERANCE DOCUMENT
MM	ASME Y14.5M	64-06-280-1392

Figure 35. 176 LQFP package mechanical drawing (Part 3 of 3)

5.1.2 144 LQFP

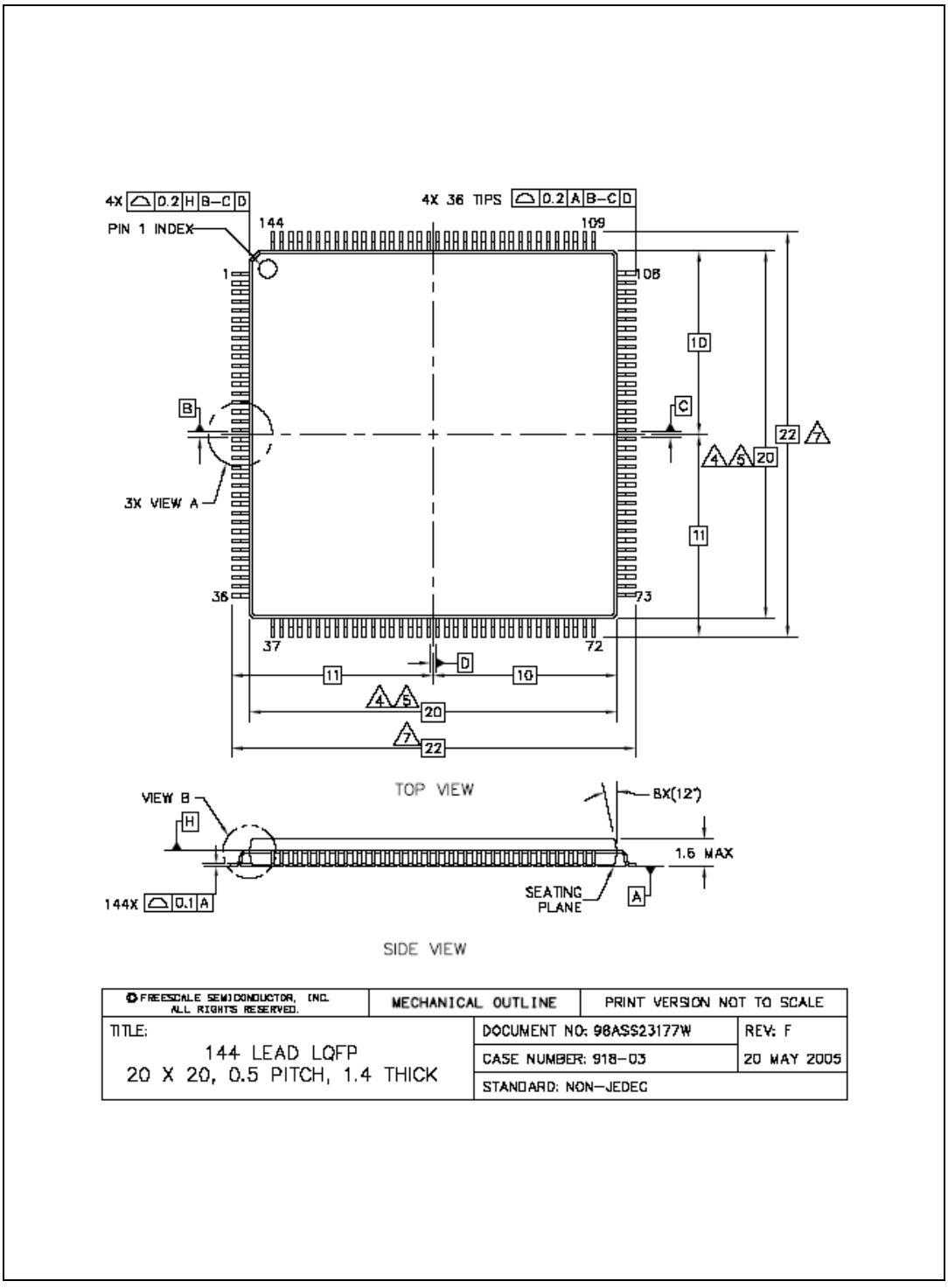


Figure 36. 144 LQFP package mechanical drawing (Part 1 of 2)

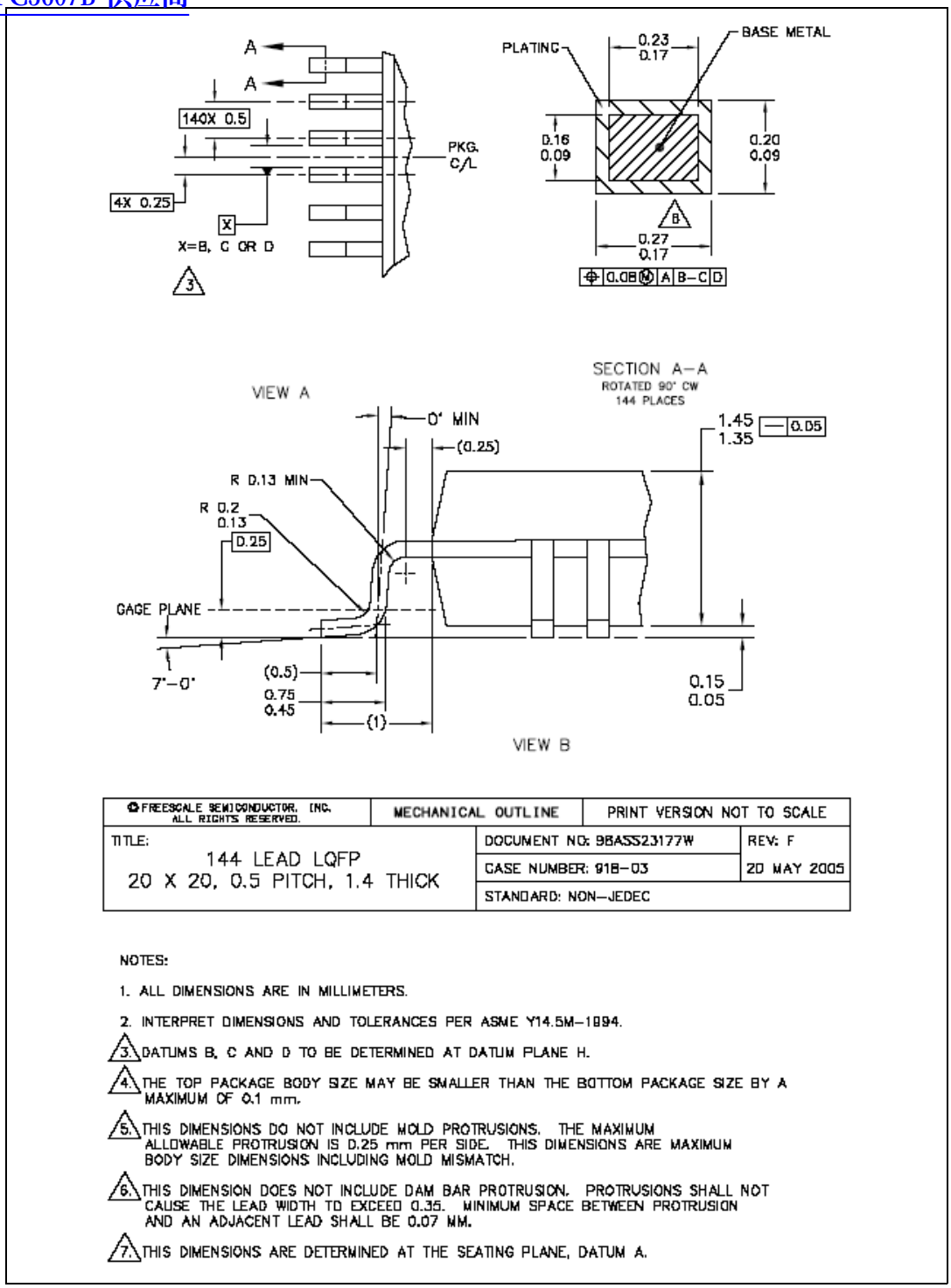


Figure 37. 144 LQFP package mechanical drawing (Part 2 of 2)

5.1.3 100 LQFP

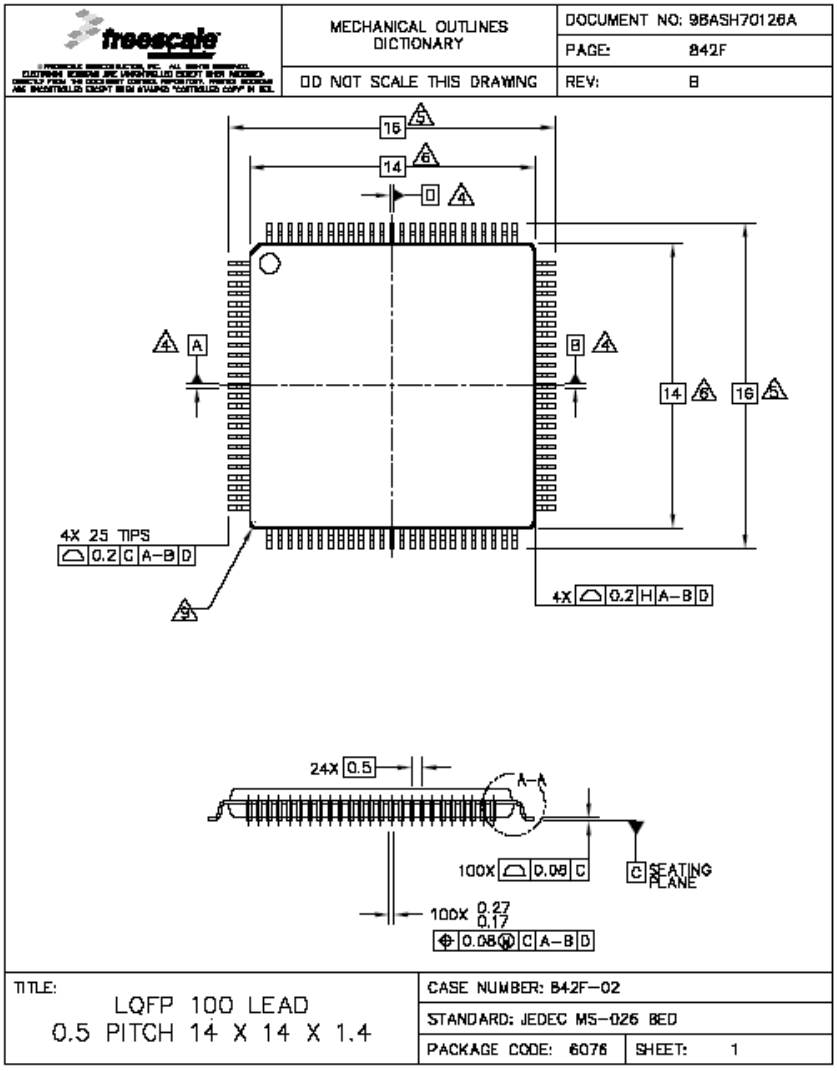


Figure 38. 100 LQFP package mechanical drawing (Part 1 of 3)

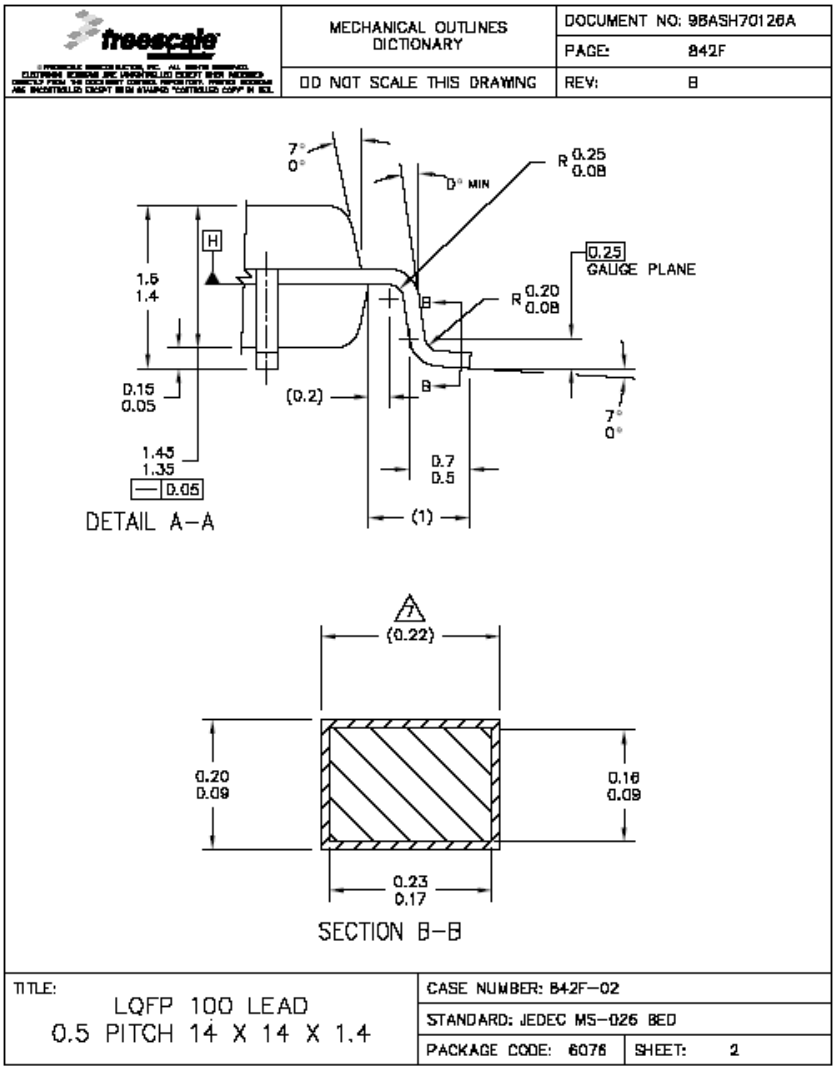


Figure 39. 100 LQFP package mechanical drawing (Part 2 of 3)

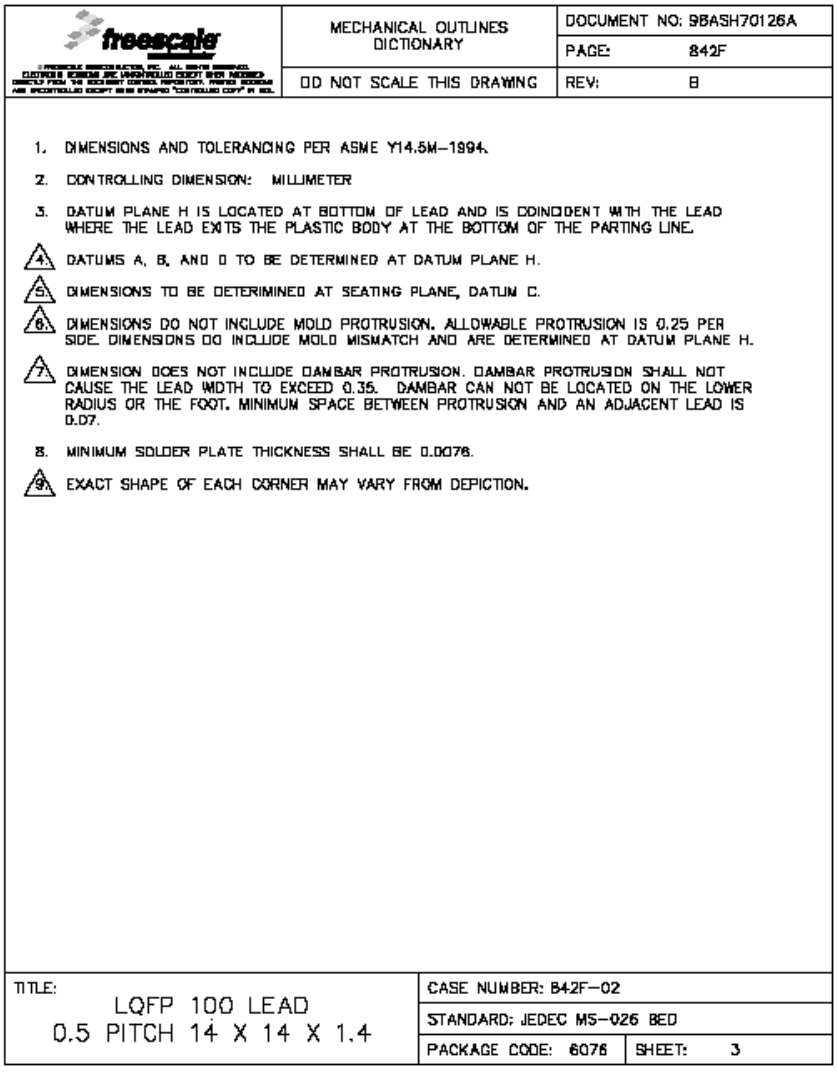


Figure 40. 100 LQFP package mechanical drawing (Part 3 of 3)

5.1.4 208 MAPBGA

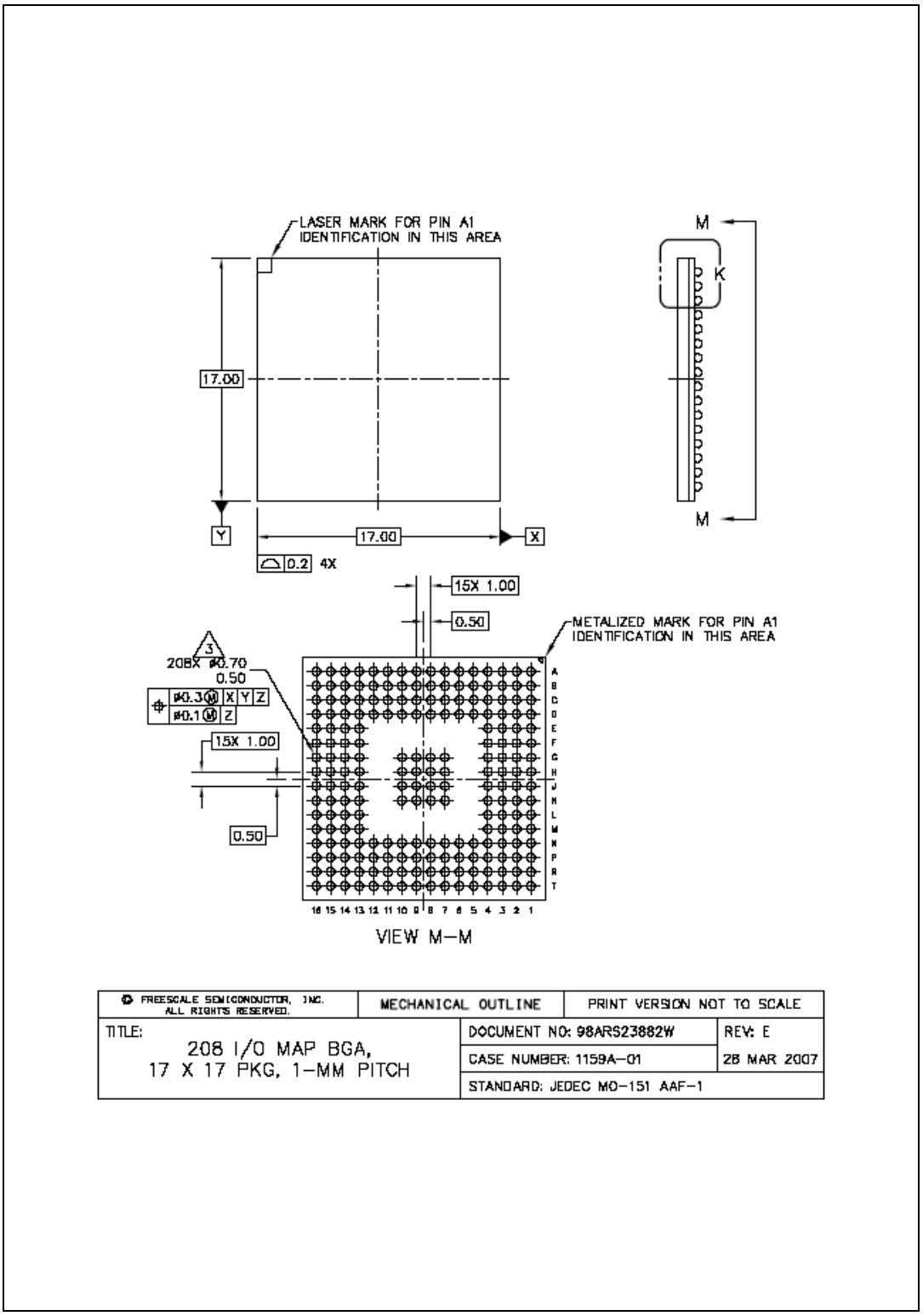
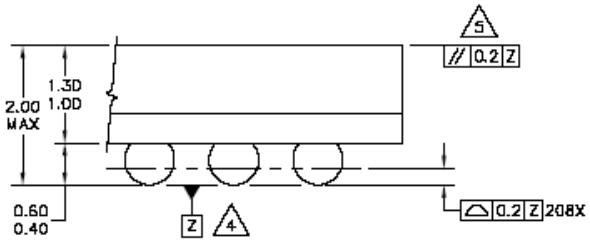


Figure 41. 208 MAPBGA package mechanical drawing (Part 1 of 2)



DETAIL K
(ROTATED 90° CLOCKWISE)

NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. INTERPRET DIMENSIONS AND TOLERANCES PER ASME Y14.5M–1994.
3. DIMENSION b IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER, PARALLEL TO DATUM PLANE Z.
4. DATUM Z (SEATING PLANE) IS DEFINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.
6. PACKAGE CODE SUMMARY:
   MAP BGA: 5253
   MAP BGA PGE DIE: 5371

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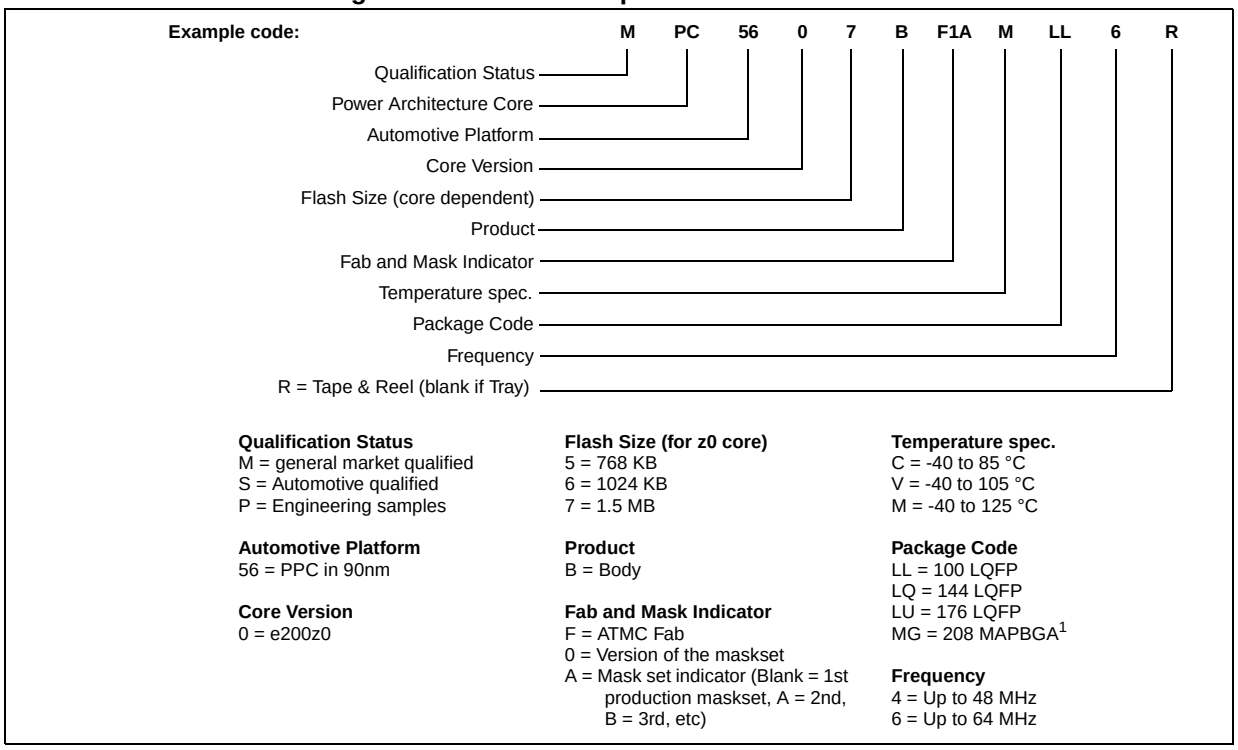
TITLE: 208 I/O MAP BGA, 17 X 17 PKG, 1–MM PITCH	DOCUMENT NO: 98ARS23882W	REV: E
	CASE NUMBER: 1159A–01	28 MAR 2007
	STANDARD: JEDEC MO–151 AAF–1	

Figure 42. 208 MAPBGA package mechanical drawing (Part 2 of 2)

6 Ordering information

Figure 43. Commercial product code structure



1 208 MAPBGA is available only as development package for Nexus2+.

Appendix A Abbreviations

Table 47 lists abbreviations used but not defined elsewhere in this document.

Table 47. Abbreviations

Abbreviation	Meaning
CMOS	Complementary metal oxide semiconductor
CPHA	Clock phase
CPOL	Clock polarity
CS	Peripheral chip select
EVTO	Event out
MCKO	Message clock out
MDO	Message data out
MSEO	Message start/end out
MTFE	Modified timing format enable
SCK	Serial communications clock
SOUT	Serial data out
TBD	To be defined
TCK	Test clock input
TDI	Test data input
TDO	Test data output
TMS	Test mode select

7 Revision history

Table 48 summarizes revisions to this document.

Table 48. Revision history

Revision	Date	Substantive changes
1	12-Jan-2009	Initial release
2	09 Nov-2009	Updated Features Replaced 27 IRQs in place of 23 ADC features External Ballast resistor support conditions Updated device summary-added 208 BGA details Updated block diagram to include WKUP Updated block diagram to include 5 ch ADC 12 -bit Updated Block summary table Updated LQFP 144, 176 and 100 pinouts. Applied new naming convention for ADC signals as ADCx_P[x] and ADCx_S[x] Section 1, “General description Updated device comparison table Updated block diagram-aligned with 512k Updated block summary-aligned with 512k Section 2, “Package pinouts Updated 100,144,176,208 packages according to cut2.0 changes Added Section 3.5.1, “External ballast resistor recommendations Added NVUSRO [WATCHDOG_EN] field description Updated Absolute maximum ratings Updated LQFP thermal characteristics Updated I/O supply segments Updated Voltage regulator capacitance connection Updated Low voltage monitor electrical characteristics Updated Low voltage power domain electrical characteristics Updated DC electrical characteristics Updated Program/Erase specifications Updated Conversion characteristics (10 bit ADC) Updated FMPLL electrical characteristics Updated Fast RC oscillator electrical characteristics-aligned with MPC5604B Updated On-chip peripherals current consumption Updated ADC characteristics and error definitions diagram Updated ADC conversion characteristics (10 bit and 12 bit) Added ADC characteristics and error definitions diagram for 12 bit ADC
3	25 Jan-2010	Updated Features Updated block diagram to connect peripherals to pad I/O Updated block summary to include ADC 12-bit Updated 144, 176 and 100 pinouts to adjust format issues Table 26 Flash module life-retention value changed from 1-5 to 5 yrs Minor editing changes

Table 48. Revision history (continued)

Revision	Date	Substantive changes
4	24 Aug 2010	Editorial changes and improvements. Updated “Features“ section Table 1: updated footnote concerning 208 MAPBGA In the block diagram: • Added “5ch 12-bit ADC“ block. • Updated Legend. • Added “Interrupt request with wakeup functionality” as an input to the WKPU block. Figure 2: removed alternate functions Figure 3: removed alternate functions Figure 4: removed alternate functions Table 3: added contents concerning the following blocks: CMU, eDMA, ECSM, MC_ME, MC_PCU, NMI, SSCM, SWT and WKPU Added Section 3.2, “Pin muxing Section 4, “Electrical characteristics: removed “Caution” note Section 4.2, “NVUSRO register: removed “NVUSRO[WATCHDOG_EN] field description“ section Table 7: V_{IN}: removed min value in “relative to V_{DD}” row Table 8 • $T_{A\ C\text{-}Grade\ Part}$, $T_{J\ C\text{-}Grade\ Part}$, $T_{A\ V\text{-}Grade\ Part}$, $T_{J\ V\text{-}Grade\ Part}$, $T_{A\ M\text{-}Grade\ Part}$, $T_{J\ M\text{-}Grade\ Part}$: added new rows • TV_{DD}: contents merged into one row • V_{DD_BV}: changed min value in “relative to V_{DD}” row Section 4.5, “Thermal characteristics • Section 4.5.1, “External ballast resistor recommendations: added new paragraph about power supply • Table 10: added $R_{\theta JB}$ and $R_{\theta JC}$ rows • Removed “208 MAPBGA thermal characteristics” table Table 11: rewrote parameter description of W_{FI} and W_{NFI} Section 4.6.5, “I/O pad current specification • Removed I_{DYNSEG} information • Updated “I/O supply segments” table Table 18: removed I_{DYNSEG} row Added Table 19 Table 21 • Updated all values • Removed $I_{VREGREF}$ and $I_{VREDLVD12}$ rows • Added the footnote “The duration of the in-rush current depends on the capacitance placed on LV pins. BV decaps must be sized accordingly. Refer to IMREG value for minimum amount of current to be provided in cc.” to the I_{DD_BV} specification. Table 22 • Updated V_{PORH} min/max value • Updated $V_{LVDLVCORL}$ min value Updated Table 23 Table 24 • $T_{dwprogram}$: added initial max value • Inserted T_{eslat} row Table 25: removed the “To be confirmed” footnote In the “Crystal oscillator and resonator connection scheme” figure, removed R_P Table 36 • Removed g_{mSXOSC} row • $I_{SXOSCBIAS}$: added min/typ/max value

Table 48. Revision history (continued)

Revision	Date	Substantive changes
4 (cont.)	24 Aug 2010 (cont.)	Table 37: • Added f_{VCO} row • Added Δt_{STJIT} row Table 38 • $I_{FIRCPWD}$: removed row for T_A = 55 °C • Updated T_{FIRCSU} row Table 41: Added two rows: $I_{ADC0pwd}$ and $I_{ADC0run}$ Table 42 • Added two rows: $I_{ADC1pwd}$ and $I_{ADC1run}$ • Updated values of f_{ADC_1} and t_{ADC1_PU} • Updated t_{ADC1_C} row Updated Table 43 Updated Table 44 Updated Figure 43 Section 6, “Ordering information: deleted “Orderable part number summary“ table
5	27 Aug 2010	Removed “Preliminary—Subject to Change Without Notice” marking. This data sheet contains specifications based on characterization data.

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